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M66335FP

Facsimile Image Data Processor

REJ03F0276-0200 Rev.2.00 Jun 16, 2008

Description

The M66335 is a facsimile image processing controller to turn into binary signals analog signals which have been output through photo-electric conversion by the image sensor.

The image processing functions includes peak value detection, uniformity correction, resolution change, MTF compensation, γ correction, detection of background/character levels, error diffusion, separation of image zones, and designation of regions.

This controller contains not only the analog processing circuit, the A/D converter of a 7-bit flash type and image processing memory, but also the image sensor and the interface circuit to the CODEC (Coder and Decoder). Therefore, this LSI alone is capable of image processing.

Features

- High speed scan (Max 2 ms/line, Typ 5 ms/line)
- Compatibility with up to the B4 (8 pixels/mm, 16 pixels/mm) image sensor
- Generation of control signals for the image sensor (CCD, CIS)

For CCD: SH, CK1, CK2, RS

For the contact sensor (CIS): SH, CK1, CK2

• Built-in analog processing circuit (equivalent to the M64291)

Sample and hold circuit

Gain control circuit

Black level clamping circuit

Reference internal power supply for the A/D converter

- Built-in A/D converter of a 7-bit flash type
- Built-in image processing memories

Uniformity correction memory, Line memory, Error memory, γ correction memory

• External output interface for converted binary data

Serial output (\rightarrow M66330)

DMA output

• External output interface for multivalued data

DMA transfer of data compensated for uniformity

• Various image processing functions

Uniformity correction

Resolution change from 50% to 200% (by the 1% step)

MTF compensation (2-dimensional processing, capable of correction for each character/photo)

γ correction (capable of correction for each character/photo)

Detection of background/character levels

Change to pseudo-halftone

- Error diffusion (64 tone steps through 6-bit processing)
- Organized dither (64 tone steps through the 8×8 matrix)

Image zone separation (2-dimensional processing)

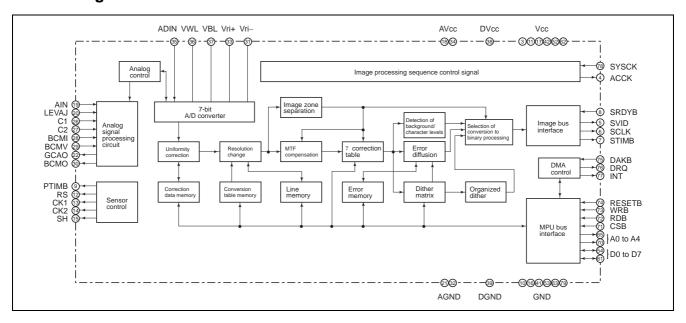
5 V single power supply

Application

Facsimile, word processor and image scanner



Block Diagram



Pin Arrangement

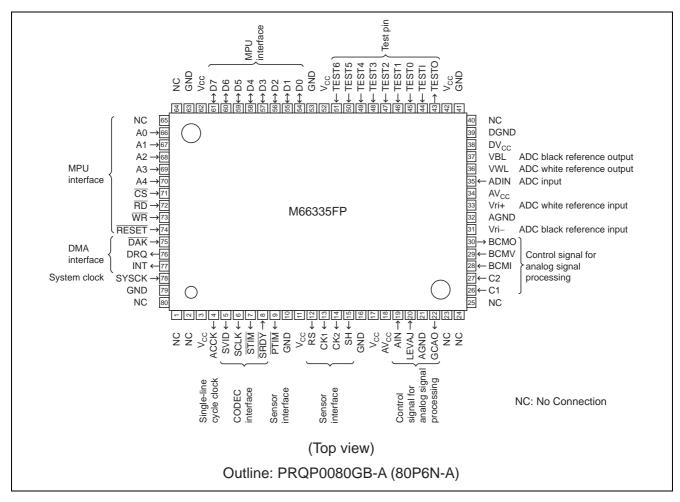


Table 1 Image Processing Functions

Image Processing Functions	Specifications	Remarks
Reading range	• A4, B4	
Resolution	8 pixels/mm, 16 pixels/mm (for the horizontal scanning direction)	
Reading speed	Typ: 5 ms/line; Max: 2 ms/line	Controlled through the system clock
Uniformity correction	White correction, black correctionCorrection range: 50%	Correction memory is built-in Readable from/writable in MPU
γ correction	Logarithmic correction	 γ correction memory is built-in Capable of correction for each character/photo
MTF compensation	Laplacian filter circuit through 2- dimensional processing	Correction memory is built-in Capable of correction for each character/photo
Simple conversion to binary	Floating slice system through the detection circuit for background/character levels	
Pseudo-halftone	 Error diffusion: 6-bit processing (for 64 tone steps) Organized dither: 8 × 8 matrix (for 64 tone steps) 	Error buffer memory is built-in 64 W × 6 bits dither memory is built-in
Image zone separation	2-dimensional processing through luminance difference	
Image reduction	Range of the reduction rate: 50% to 100% (by the 1% step)	Capable of outputting the average line of a dropped line and the subsequent line instead of both lines
Image enlargement	Range of the enlargement rate: 100% to 200% (by the 1% step)	Capable of outputting the average line of a repeated line and the subsequent line instead of the repeated line
Image sensor control signal	CIS image sensor (clock duty: 75%) CCD image sensor	
Analog processing	The sample/hold circuit, gain control amplifier, black level clamping circuit, and 7-bit A/D converter are built-in.	

Pin Description

Item	Pin Name	Input/Output	Function
Sensor	SH	Output	Outputs the shift pulse signal to transfer electric charges from the
interface			sensor's photoconductor component to its transferring component for
			CCD and the start signal to start the sensor reading circuit for CIS.
	CK1	Output	Outputs the clock pulse signal to sequentially transfer out signaling
			electric charges from the sensor's transferring component for CCD and
			the clock pulse signal for the shift register of the sensor reading circuit for
			CIS.
	CK2	Output	Reversed-phase pulses of CK1
	RS	Output	Outputs the reset pulse to return the voltage at the floating capacitor of
			the CCD sensor to the initial one.
	PTIM	Output	Outputs the pulse motor control signal for the reading roller.
CODEC	SRDY	Input	Transfer start ready signal for data from CODEC
interface	STIM	Output	Defines the data transfer section to CODEC
	SCLK	Output	Clock signal to transfer image data to CODEC
	SVID	Output	Outputs image data in serial to CODEC
DMA	DRQ	Output	DMA request signal to the external DMA controller to output in parallel
interface			image data through the MPU bus
	DAK	Input	DMA acknowledge signal from the external DMA controller in response to
		1	the above DRQ signal
	INT	Output	Single-line termination interrupt
Clock	SYSCK	Input	System clock input pin
	ACCK	Output	Single-line cycle clock
MPU	RESET	Input	Input of the system reset. The cycle counter, register, F/F, and latch are
interface			reset.
intoriaco	CS	Input	Chip select signal for MPU to access the M66335
	RD	Input	Control signal for MPU to read data from the M66335
	WR	Input	Control signal for MPU to write data to the M66335
	A0 to A4	Input	Address signal to access various registers inside the M66335
	D0 to D7	Input/Output	8-bit two way buffer
Others		input/Output	•
Others	V _{CC}	 	Positive power supply pin
	GND	 -	GND pin
	TESTI, 0 to 6	Input	Test input pin. Hold this at "L".
	TESTO	Output	Test output pin. Set this open.
Power	AV _{CC}	<u> </u>	Analog power supply pin (rated supply voltage: 5 V)
supply	DV _{CC}	_	Digital power supply pin (rated supply voltage: 5 V)
GND	AGND	_	Analog ground pin
	DGND	_	Digital ground pin
Sensor	AIN	Input	Pin to input analog signals output from CCD or CIS (Signals from CCD
signal			are input through capacity coupling and those from CIS, with no
input part			clamping levels, are input directly.)
Gain	C1, C2	Input	Pin to control the frequency characteristic of the gain control circuit
control	LEVAJ	Input	Pin to control the DC level of output signals of the gain control circuit.
circuit			The output voltage, V _{GCAO} , is obtained by the following equation:
			$V_{GCAO} = V_{LEVAJ} + G_V \times V_{IN}$
			where,
			V _{LEVAJ} : voltage at LEVAJ
			V _{IN} : input signal
			G _V : gain of the gain control circuit
			V _{IN} is the signal element corresponding to the signal level clamped
			through the input clamping circuit for CCD • CIS3 input or to the GND
			level for CIS1 • CIS2 input.
			level for Olo 1 - Oloz Input.



Pin Description (cont.)

Item	Pin Name	Input/Output	Function
Black level clamping	BCMI	Input	Signal input pin to the black clamping circuit. Use this with capacity coupling with the GCAO pin.
circuit	BCMV	Input	Pin to set the black level clamping voltage. Sets the black level of signals output from the BCMO pin for CCD signal processing.
	ВСМО	Output	Signal output pin of the black level clamping circuit
A/D converter	Vri+	Input	Output of the circuit to generate the A/D full-scale point reference voltage (3.8 V). Connected with VWL through the buffer inside the IC. To change the A/D reference voltage range, input a DC voltage from this pin.
	Vri–	Input	Output of the circuit to generate the A/D zero point reference voltage (1.8 V). Connected with VBL through the buffer inside the IC. To change the A/D reference voltage range, input a DC voltage from this pin.
	ADIN	Input	Signal input pin to the A/D converting circuit. Use this by connecting with the BCMO pin for CCD or with the GCAO pin for CIS. Input signals in the voltage range (1.8 V to 3.8 V) set through VWL and VBL.
	VWL	Output	Output of the circuit generating the A/D full-scale reference voltage (3.8 V). Connected inside the IC with the A/D converter.
	VBL	Output	Output of the circuit generating the A/D zero point reference voltage (1.8 V). Connected inside the IC with the A/D converter.

Absolute Maximum Ratings

 $(Ta = -20 \text{ to } +75^{\circ}\text{C}, \text{ unless otherwise noted})$

Item	Symbol	Ratings	Unit
Supply voltage	V _{CC}	-0.3 to +6.5	V
Input voltage	V _I	-0.3 to V _{CC} + 0.3	V
Output voltage	Vo	0 to V _{CC}	V
Analog supply voltage	AV _{CC}	$V_{CC} - 0.3$ to $V_{CC} + 0.3$	V
Supply voltage	DV _{CC}	$V_{CC} - 0.3$ to $V_{CC} + 0.3$	V
Reference voltage (white)	V _{WL}	-0.3 to AV _{CC} + 0.3	V
Reference voltage (black)	V_{BL}	-0.3 to AV _{CC} + 0.3	V
Analog input voltage	V _{AIN}	-0.3 to AV _{CC} + 0.3	V
Storage temperature	Tstg	−55 to +150	°C

Recommended Operational Conditions

Item	Symbol	Min	Тур	Max	Unit
Supply voltage (for the digital system component)	V _{CC}	4.75	5.0	5.25	V
GND voltage	GND	_	0.0	_	V
Input voltage	VI	0	_	V _{CC}	V
Analog supply voltage	AV _{CC}	4.75	5.0	5.25	V
Analog GND voltage	A _{GND}	_	0.0	_	V
Supply voltage (for the digital system component)	DV _{CC}	4.75	5.0	5.25	V
GND voltage	D _{GND}	_	0.0	_	V
Input range: $V_{WL} \le AV_{CC}$; $V_{BL} \ge A_{GND}$	V _{AIN}	1.8	2.0	2.2	Vp-p
Operating temperature	Topr	-20	_	+75	°C

Note: Connect the analog system component and the digital system component separately to power supply on the evaluation board for noise prevention.



Electrical Characteristics

(Ta = -20 to +75°C, V_{CC} = 5 V \pm 5%, unless otherwise noted)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
"H" input voltage	V_{IH}	2.0	_	_	V	
"L" input voltage	V_{IL}	_	_	0.8	V	
Positive direction input threshold	VT+	_	_	2.4	V	
Negative direction input threshold	VT-	0.6	-	_	V	
Hysteresis value	V _H	_	0.2	_	V	
"H" output voltage	V _{OH}	V _{CC} - 0.8	-	_	V	$I_{OH} = -12 \text{ mA}$
"L" output voltage	V _{OL}	_	-	0.55	V	I _{OL} = 12 mA
"H" output voltage	V _{OH}	V _{CC} - 0.8	-	_	V	$I_{OH} = -4 \text{ mA}$
"L" output voltage	V _{OL}	_	-	0.55	V	$I_{OL} = 4 \text{ mA}$
"H" input current	I _{IH}			1.0	mA	$V_{CC} = 5.25 \text{ V}$ $V_{I} = 5.25 \text{ V}$
"L" input current	I _{IL}	_	_	-1.0	mA	$V_{CC} = 5.25 \text{ V}$ $V_{I} = 0 \text{ V}$
"H" input current in the off state	l _{OZH}	_	_	5.0	mA	$V_{CC} = 5.25 \text{ V}$ $V_{O} = 5.25 \text{ V}$
"L" input current in the off state	lozL	_		-5.0	mA	$V_{CC} = 5.25 \text{ V}$ $V_{O} = 0 \text{ V}$
Analog input current	I _{AIN}	_	_	1.0	mA	
Reference resistance	R _L	_	120	_	Ω	
Differential non-linear error	Ed	_	±1.0	_	LSB	
Static current dissipation (during standby)	I _{ccs}	_	21	35	mA	$V_{CC} = 5.25 \text{ V}$ $V_{I} = V_{CC}, \text{ GND}$

Timing Conditions

(Ta = -20 to +75°C, V_{CC} = 5 V \pm 5%, unless otherwise noted)

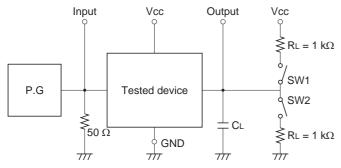
Item		Symbol	Min	Тур	Max	Unit
System clock cycle		t _{c (SYS)}	50	_	_	ns
System clock "H" pulse wi	dth	t _{w+ (SYS)}	25	_	_	ns
System clock "L" pulse wid	dth	t _{w- (SYS)}	25	_	_	ns
System clock rise time		t _{r (SYS)}	_	_	20	ns
System clock fall time		t _{f (SYS)}	_	_	20	ns
Read pulse width		t _{w (RD)}	100	_	_	ns
Set-up time before read	CS	t _{su (CS-RD)}	20	_	_	ns
Set-up time before read	A0 to A4	t _{su (A-RD)}	20	_	_	ns
Set-up time before read	DAK	t _{su (DAK-RD)}	20	_	_	ns
Hold time after read	CS	t _{h (RD-CS)}	10	_	_	ns
Hold time after read	A0 to A4	t _{h (RD-A)}	10	_	_	ns
Hold time after read	DAK	t _{h (RD-DAK)}	10	_	_	ns
Write pulse width		t _{w (WR)}	100	_	_	ns
Set-up time before write	CS	t _{su (CS-WR)}	20	_	_	ns
Set-up time before write	A0 to A4	t _{su (A-WR)}	20	_	_	ns
Set-up time before write	D0 to D7	t _{su (D-WR)}	50	_	_	ns
Hold time after write	CS	t _{h (WR-CS)}	20	_	_	ns
Hold time after write	A0 to A4	t _{h (WR-A)}	10	_	_	ns
Hold time after write	D0 to D7	t _{h (WR-D)}	0	_	_	ns
Hold time after STIM	SRDY	t _{h (STIM} -SRDY)	0	_	_	ns

Switching Characteristics

(Ta = -20 to +75°C, $V_{CC} = 5$ V \pm 5%, unless otherwise noted)

Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Enable time for data output	t _{PZL (RD-D)}	_	_	75	ns	C _L = 150 pF
after read	t _{PZH (RD-D)}				ns	
Disable time for data output	t _{PLZ} (RD-D)	10	_	50	ns	
after read	t _{PHZ (RD-D)}				ns	
Propagation time of DRO output after read	t _{PHL} (RD-DRO)	_	_	50	ns	C _L = 50 pF

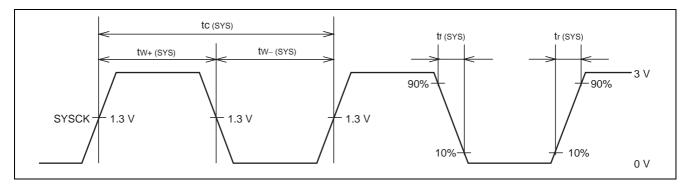
Test Circuit



Item	SW1	SW2
t _{PLH} , t _{PHL}	Open	Open
t _{PLZ}	Closed	Open
t _{PHZ}	Closed	Closed
t _{PZL}	Closed	Open
t _{PZH}	Open	Closed

- (1) Characteristics (10% to 90%) of the pulse generator (PG): t_r = 3 ns; t_f = 3 ns
- (2) Capacitance C_L (= 150 pF) includes the stray capacitance of connections and input capacitance of the probe.

System Clock



(1) Operation Mode

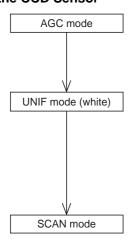
The M66335 has three basic operations.

- Peak value detection: Adjusting the peak value of analog signals output from the analog circuit to the white reference voltage (VWL) of the A/D converter built in the M66335.
- Generation of data for uniformity correction: Generating data on a white reference original sheet for uniformity correction by the sensor unit and writing them to the memory for correction built-in the M66335.
- Read: Reading original sheets, performing image processing of the read image data, and outputting in serial or parallel the indicated converted binary data.

The M66335 is capable of performing the DMA transfer of multivalued data (6-bit data = D7 to D2, D1 = D0 = 0) after correction about uniformities.

These three basic operations are performed in the following mode sequences for the CCD sensor and CIS sensor. The sensor is set through the register 00 (SENS).

For the CCD Sensor



The peak value of the 16 line cycle is detected by setting the AGC command in the register 00 at "H".

To escape this mode, set the AGC command at "L" after a 20 line cycle (or a cycle of 16 lines or more) passed since the start.

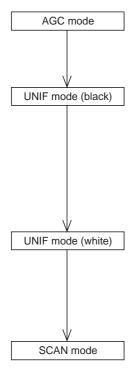
This operation mode is started by setting the UNIF command in the register 00 at "H" after setting UMODE: "H" (white correction) in the register 00 and UNIFM: "L" (only white correction) in the register 01.

Starting by the UNIF command also makes the system generate data for non-uniformity correction for white correction (for the 8 line cycle).

To escape this mode, set the UNIF command at "L" after a 10 line cycle (or a cycle of 8 lines or more) passed since the start.

The read operation mode is started by setting the SCAN command in the register 00 at "H". To escape this mode, set the SCAN command at "L".

For the CIS Sensor



The peak value of the 16 line cycle is detected by setting the AGC command in the register 00 at "H".

To escape this mode, set the AGC command at "L" after a 20 line cycle (or a cycle of 16 lines or more) passed since the start.

When this operation mode is started by the UNIF command after setting UMODE: "L" (black correction) in the register 00 and UNIFM: "H" (black and white correction) in the register 01, the system also generates black data for non-uniformity correction for black correction (for the 8 line cycle).

To escape this mode, set the UNIF command at "L" after a 10 line cycle (or a cycle of 8 lines or more) passed since the start.

In the case of only white correction, the setting is not necessary. Follow the instruction below.

When this operation mode is started by the UNIF command in the register 00 after setting UMODE: "H" (white correction) in the register 00 and UNIFM: "L" (only white correction) in the register 01, the system also generates white data for non-uniformity correction for white correction (for the 8 line cycle).

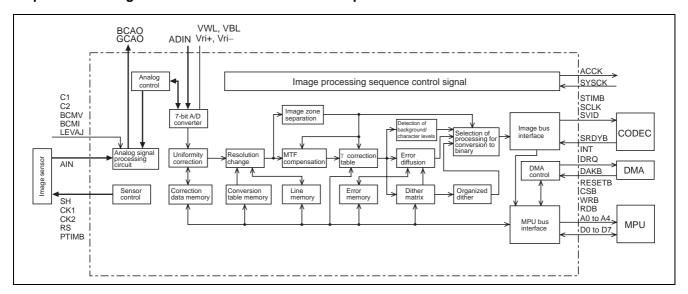
To escape this mode, set the UNIF command at "L" after a 10 line cycle (or a cycle of 8 lines or more) passed since the start.

The reading operation is started by setting the SCAN command in the register 00 at "H". To escape this mode, set the SCAN mode at "L".

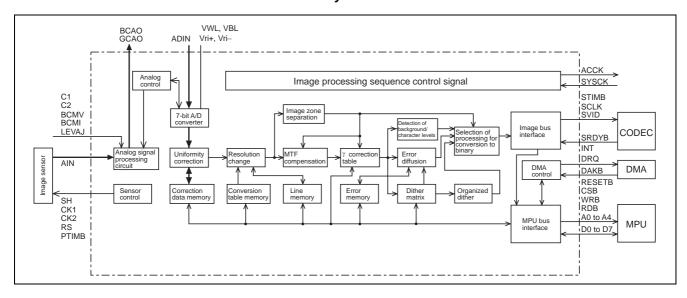
The signal operations and data flow in each basic operation are shown in the page 9 and 10, and the flowchart is in the page 26 and 27.



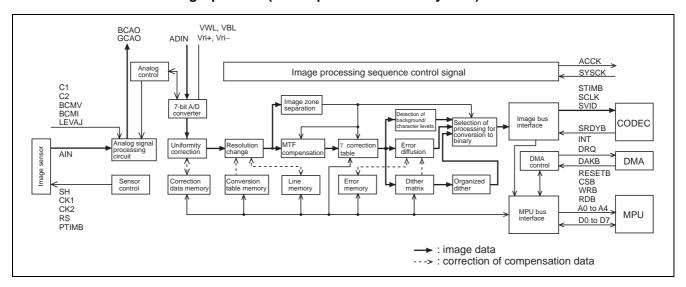
Operations of Signals in the Peak Value Detection Operation



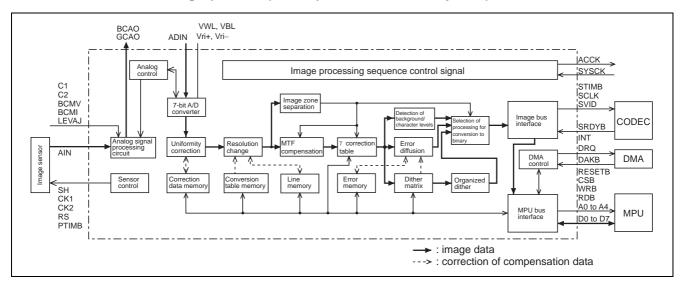
Flow of Data in the Creation of Data for Uniformity Correction



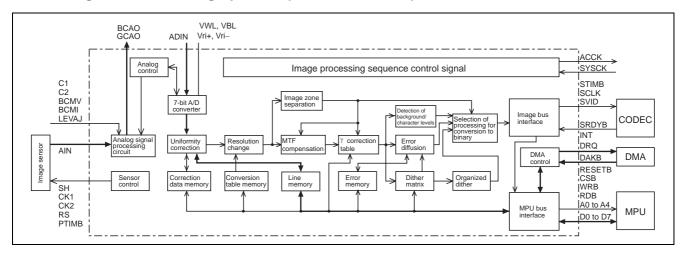
Flow of Data in the Reading Operation (for Output in Serial: Binary Data)



Flow of Data in the Reading Operation (for Output in Parallel: Binary Data)



Flow of Signals in the Reading Operation (for Multivated Data)



(2) Line Cycle and Reading Sequence

The relationship between the line cycle and the reading sequence of the M66335 is shown in figure 1.

The relationship between the CODEC interface operations and the reading sequence is shown in figure 2 and that between the DMA interface operations and the reading sequence is shown in figure 3.

• Single-line cycle (1/ACCK)

Defines the processing time per line of the M66335.

The single-line cycle is decided by the line cycle counter value registers 03 and 04 (PRE_DATA), and the pixel transfer clock.

The pixel transfer clock is 1/16 of SYSCK.

- 1 line cycle (1/ACCK) [NS]
- = line cycle counter value × pixel transfer clock cycle [NS]
- = (PRE_DATA + 1) × pixel transfer clock cycle [NS]
- $= (PRE_DATA + 1) \times 16/SYSCK [NS]$

After loading the PRE_DATA value, the line cycle counter generates the addresses of the following gate signals while counting down with the pixel transfer clock.

Sensor start pulse (SH)

Image sensor start pulse. The point of the start pulse is decided by the uniformity correction range (UNIFG) and the value of the register 05.

[ST_PL]

The ST_PL value must be set according to the following formulas for each image sensor type.

CCD: ST_PL = dummy pixels of the sensor + 2

CIS: ST PL = 2

• Uniformity correction range (UNIFG)

Defines the range where uniformity correction is performed. This range corresponds to the width of the sensor (B4 to A4).

For the relationship between the sensor width and the uniformity correction range, see table 2.

• AGC range (AGCG)

Defines the range where peak value detection is performed. This range corresponds to the sensor width (B4 to A4). Auto gain control is performed for the whole width of the sensor (solid line) in the AGC mode and for the narrower width (dashed line) than the sensor width in the SCAN mode.

For the relationship between the sensor width and the AGC range, see table 2.

• Original sheet reading width

Defines the reading width for original sheets.

For original sheet widths narrower than the sensor width, the reading range (dashed line) is set, using the sensor center as the base center point. Therefore, the points for the original sheet should be based on the sensor center. For the relationship between the sensor width and the original sheet reading width, see table 3.

• Pulse motor control signal (PTIM)

Generates control signals for the pulse motor for the reading roller.



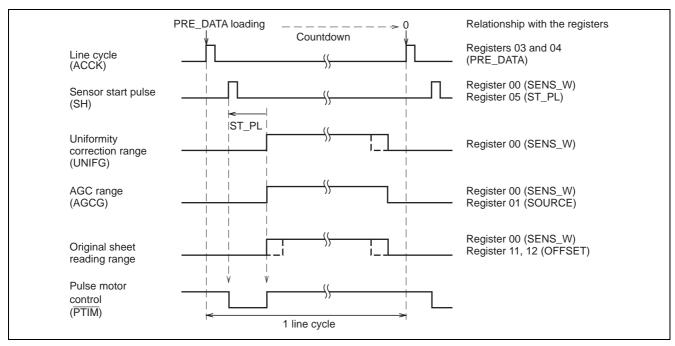


Figure 1 Line Cycle and the Reading Sequence

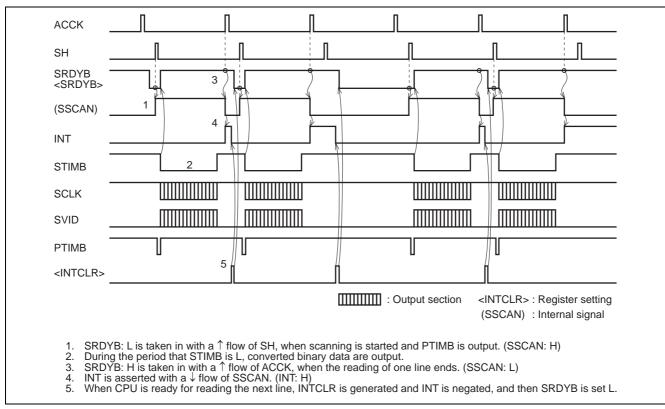


Figure 2 CODEC Interface Operations and the Reading Sequence (Binary Data Output: Serial Output)

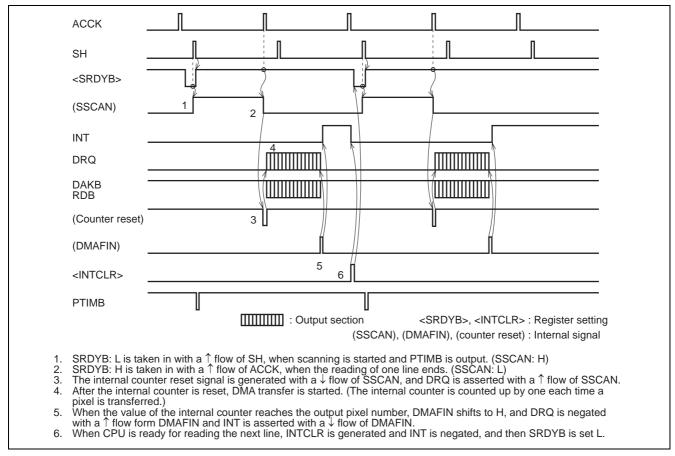


Figure 3 DMA Interface Operations and the Reading Sequence (Multivated Data Output)

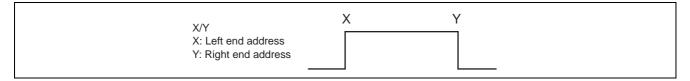
Table 2 Gate Signal Ranges for the Sensor Widths

	Sensor Width			
Gate Signal		Resolution	B4	A4
Uniformity correction ra	ange (UNIFG)	200 dpi	2103/55	1943/215
		400 dpi	4207/111	3887/431
AGC range (AGCG)	AGC mode	200 dpi	2103/55	1943/215
		400 dpi	4207/111	3887/431
	SCAN mode	200 dpi	2018/130	1584/564
		400 dpi	4037/261	3169/1129

Table 3 Original Sheet Reading Widths According to the Original Sheet Widths for the Sensor Widths

	Sensor Width			
Original Sheet Width		Resolution	B4	A4
B4		200 dpi	2102/54	_
		400 dpi	4206/110	
A4		200 dpi	2102/54	1942/214
		400 dpi	4206/110	3886/430

When original sheets narrower than the sensor width, cut out the original sheet width with the registers 11 to 14. (OFFSET, OUTLENGTH): (Region designation function)



(3) Image Processing Function

The M66335 converts image signals input from the image sensor into binary data. This includes the simple conversion of characters and the change of images with various densities into pseudo-half-tone.

Before the conversion, distortions and characteristic degradations which signals from the image sensor almost always have must be corrected or compensated.

Image zone separation must also be performed to realize optimal conversion-to-binary of the image for the possible shortest transmission time.

Functions required for image processing are as follows.

- Peak value detection
- Uniformity correction
- Resolution change (enlargement, reduction and averaging)
- MTF compensation
- γ correction
- Background/character level detection (simple conversion to binary)
- Change to pseudo-halftone
 - Organized dither
 - Error diffusion
- Image zone separation
- · Designation of regions

Peak Value Detection

Because the A/D converter of the M66335 uses the input dynamic range at 2 Vp-p, the reference voltages (V_{WL} , V_{BL}) corresponding to the peak value are fixed. The peak value of analog signals output from the analog processing circuit must be detected before those signals are input to the A/D converter in order to adjust the analog signal peak value to the full-scale value of the converter.

The peak value detection is performed by reading white data from the sensor in the AGC mode selected from its three modes (AGC, UNIF and SCAN) of the M66335.

As shown in figure 4, preprocessing of peak value detection to increase the gain at the gain control is performed for a 8 line cycle and gain control processing to decrease the gain when the A/D converter over-flows is performed for another 8 line cycle after the start command (register 00: AGC) in the AGC mode.

As a result, the gain changes as shown in figure 5.

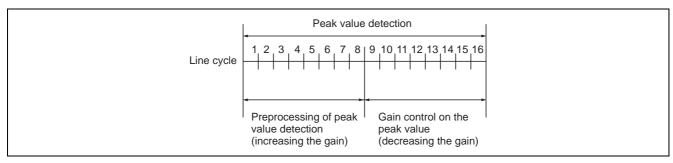


Figure 4 Peak Value Detection

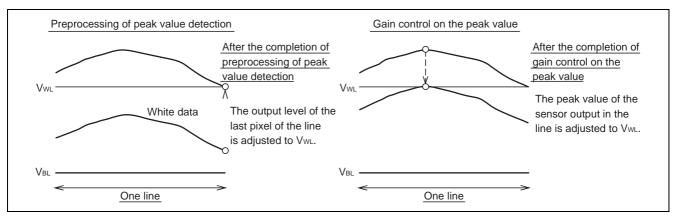


Figure 5 Changes of the Gain in Peak Value Detection

Uniformity Correction

Uniformity correction is to correct shading distortion due to less light at each end of the light source and faded light around the lens, or high frequency distortion due to characteristic variations pixel by pixel in the image sensor.

As shown in figure 7, the M66335 makes blocks each of two pixels, creates a set of uniformity correction data for each block, and write them to the built-in correction memory (SRAM: $1024 \text{ word} \times 6 \text{ bits}$) in the UNIF mode selected from its three modes (AGC, UNIF and SCAN).

The correction data created each for two pixels are read from the built-in correction memory to correct the input image data consecutively in the SCAN mode. With the register 01 (UNIFS) set at "1", the uniformity is not implemented.

With the register 02 (RES) set at "1", uniformity correction is performed on a block for 4 pixels.

For uniformity correction, white correction or the combination of black correction and white correction can be selected according to the types of image sensors as shown in table 4.

This is set in the register 00 (SENS, UMODE) and register 01 (UNIFM).

To perform both black correction and white correction, the black correction must be done first.

The M66335 implements the correction in the correction range of 50% as shown in figure 7. If a set of white correction data is beyond the correction range of 50%, the correction are not exactly performed as shown in figure 7. Therefore, ensure that input signals are within the range.

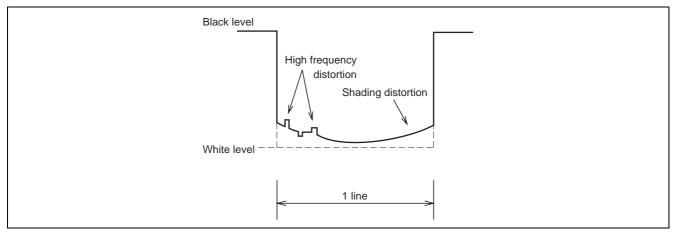


Figure 6 Waveform of White Data Output from the Image Sensor

Table 4 Uniformity Correction due to the Image Sensor

		Register					
Image Sensor	Correction	Type of the Sensor Register 00 (SENS)	Creation of Uniformity Correction Data Register 00 (UMODE)	Selection of Correction Mode Register 01 (UNIFM)			
CCD	White correction	0	1	0			
CIS	White correction	1	1	0			
	Black correction White correction	1	Period of black correction: 0 Period of white correction: 1	1			

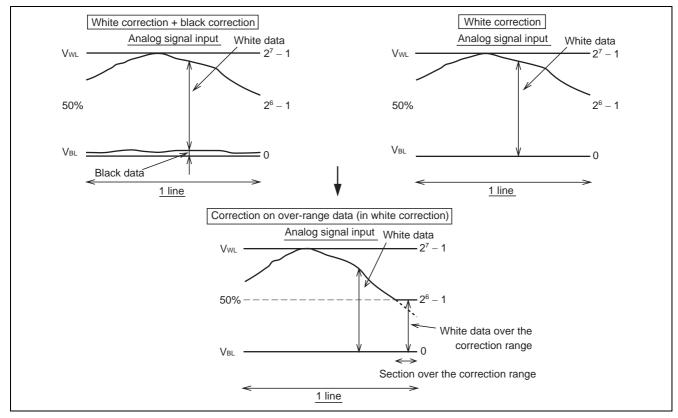


Figure 7 Uniformity Correction

Resolution Change

Resolution change is controlled through H/W in the horizontal scanning direction and through S/W in the vertical scanning direction.

The sequence for resolution change is shown in figure 8.

Horizontal Scanning Direction

The scaling factor is written from the register 15 (CNV_D) to the built-in resolution change memory (100 W \times 1 bit) bit by 100 operations.

MSSEL of the register 6 must be set at "0" (which specifies the horizontal scanning direction) before the scaling factor is written in the memory.

The procedure to specify CNV_D is as follows.

In the Case of Reduction

Data written in the resolution change memory have the following meaning.

"0": 1 pixel is output.

"1": No pixel is output.

(Example of reduction to 75%)

75 0's and 25 1's are written in the memory. The intervals of 1's should be as equal as possible to obtain the image with better quality.

In the Case of Enlargement

Data written in the resolution change memory have the following meaning.

"0": 1 pixel is output.

"1": 2 pixels are output.

(Example of enlargement to 150%)

50 0's and 50 1's are written in the memory. The intervals of 1's should be as equal as possible to obtain the image with better quality as in the reduction.



Vertical Scanning Direction

Processing of lines to implement the scaling factor in the vertical scanning direction is decided for each line through the register.

MSSEL of the register 6 must be set at "1" (which specifies the vertical scanning direction), and either "0" or "1" written in the register 15 (CNV_D) before the processing of each line.

The timing for this setting is in the period between the first transition of the INT signal (synchronized with that of ACCK) and that of the SH signal (the start of taking the SRDY signal in).

The procedure to specify CNV_D is as follows.

In the Case of Reduction

CNV_D indicates the current line read.

"0": 1 line of data are output.

"1": No line of data are output.

In the Case of Enlargement

CNV_D indicates the next line read.

"0": 1 line of data are output with PTIM generated (paper driven).

"1": 1 line of data are output with PTIM generated (paper not driven).

(Paper not driven: the same line is read again.)

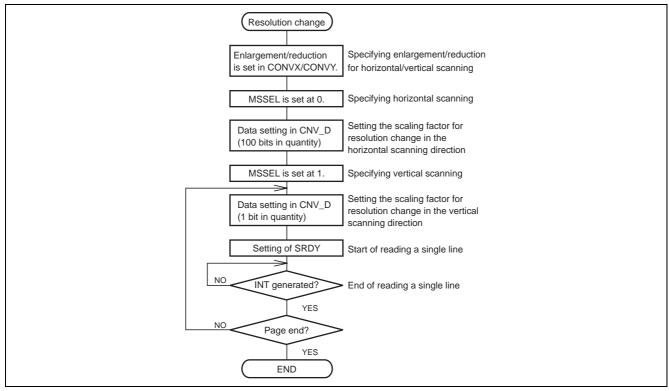
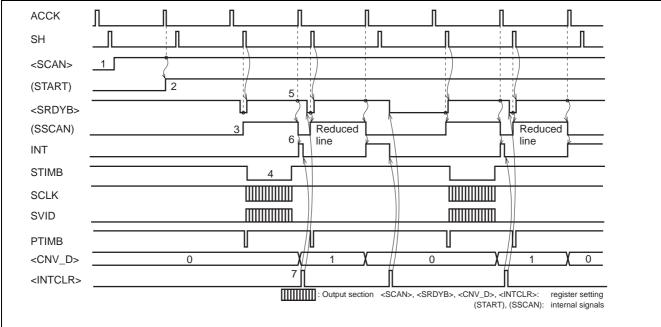


Figure 8 Sequence of Resolution Change Setting

Use the PTIMB signal as control signals for the pulse motor for the reading roller. The sequence for reduction is shown in figure 9 and that for enlargement in figure 10.





- 1. At the initial setting, the enlargement/reduction setting (CNV_D) in horizontal scanning is implemented. Then, after the system is switched into the setting mode for enlargement/reduction in vertical scanning, the first line is set.
- 2. With a ↓ flow of ACCK, the SCAN command is taken in, when the system comes into the standby mode for SRDYB. (START: H)
- 3. With a ↑ flow of SH, SRDYB: L is taken in, when scanning starts and PTIMB is output. (SSCAN: H)
- 4. During the period that STIMB is at L, converted binary data are output while the data for reduced lines are not output because STIMB for them are at H.
- 5. With a ↑ flow of ACCK, SRDYB: H is taken in, when the reading of the single line is completed. (SSCAN: L)
- 6. With a ↓ flow of SSCAN, INT is asserted. (INT: H)
- 7. With CPU ready for reading the next line, the enlargement/reduction setting (CNV_D) in vertical scanning is implemented; INTCLR is generated; INT is negated; and then SRDYB is set at L.

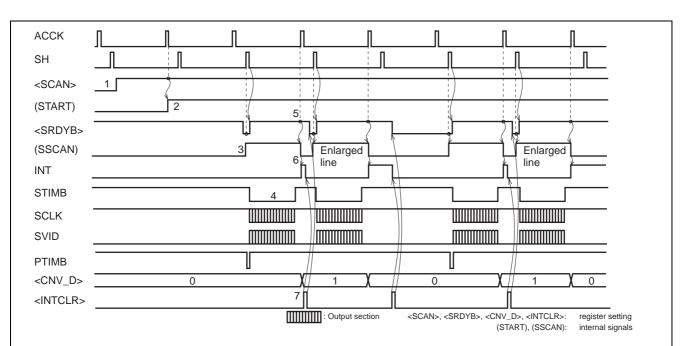


Figure 9 Reduction Processing Sequence

- 1. At the initial setting, the enlargement/reduction setting (CNV_D) in horizontal scanning is implemented. Then, after the system is switched into the setting mode for enlargement/reduction in vertical scanning, the first line is set.
- 2. With a ↓ flow of ACCK, the SCAN command is taken in, when the system comes into the standby mode for SRDYB. (START: H)

 3. With a ↓ flow of SH, SRDYB: L is taken in, when scanning starts and PTIMB is output while it is not output for enlarged lines. (SSCAN: H)
- 4. During the period that STIMB is at L, converted binary data are output.
- 5. With a ↑ flow of ACCK, SRDYB: H is taken in, when the reading of the single line is completed. (SSCAN: L)
- 6. With a ↓ flow of SSCAN, INT is asserted. (INT: H)
- 7. With CPU ready for reading the next line, the enlargement/reduction setting (CNV_D) in vertical scanning is implemented; INTCLR is generated; INT is negated; and then SRDYB is set at L.

Figure 10 Enlargement Processing Sequence

MTF Compensation

As shown in figure 11, image data of characters or pictures photoelectrically converted by the sensor unit show degradation in resolution.

MTF compensation function of the M66335 restores the resolution of those data and expands the apparent dynamic range by strengthening the high-pass frequency constituent with the Laplacian filter.

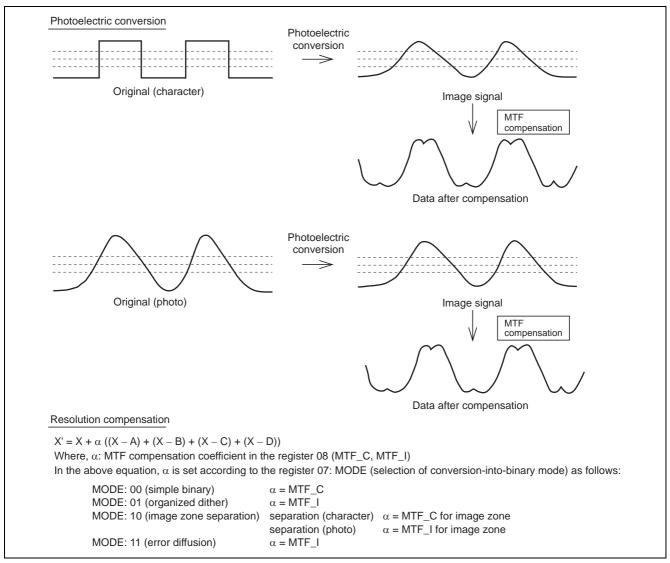


Figure 11 MTF Compensation

γ Correction

 γ correction according to the sensitivity characteristics (logarithmic characteristics) of human eyes is implemented to approximate the image data to natural images.

To do this, the M66335 writes the γ correction table to the built-in SRAM and read the corrected values corresponding to read image data values from the SRAM.

 $\gamma = 0.45$ is considered to be the optimal for γ correction for thermal head printers. Figure 12 shows a characteristics example at $\gamma = 0.45$.

 γ correction processing is set through the register 06: GAMMA as follows.

GAMMA: $00 \quad \gamma = 1$

GAMMA: 01 γ = conversion table value

GAMMA: 10 $\gamma = 1$ for image zone separation (character)

 γ = conversion table value for image zone separation (photo)

GAMMA: 11 γ = conversion table value for image zone separation (character)

 $\gamma = 1$ for image zone separation (photo)

For the procedures of inputting/outputting of data, refer to the section on writing to/reading from the γ correction memory.

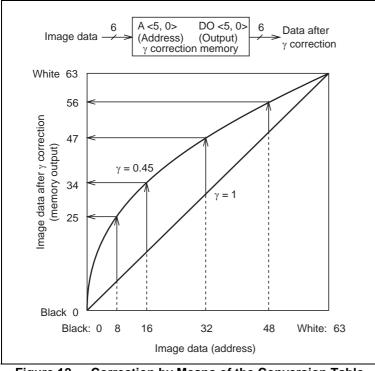
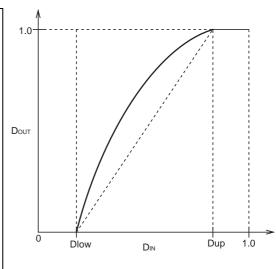


Figure 12 γ Correction by Means of the Conversion Table



IF (DIN < Dlow) DOUT = 0

 $\begin{aligned} \text{IF (Dlow} & \leq \text{Din} < \text{Dup)} \\ \text{Dout} & = \left(\frac{\text{Din} - \text{Dlow}}{\text{Dup} - \text{Dlow}}\right)^{\gamma} \end{aligned}$

IF (Dup \leq DiN)
DOUT = 1.0

Background/Character Level Detection

The M66335 uses not the fixed threshold system but the floating threshold system, where the optimal threshold for simple conversion-to-binary of objective pixels are continually generated by constantly detecting background/character levels.

Accordingly, the threshold value proper for image data is generated without processing the data.

The threshold value is used for the areas to be converted to binary when simple conversion-to-binary or image zone separation is selected as the mode of conversion to binary in reading data.

: register 07 (MODE)

· Background level counter

When image data greater (lighter in light) than the current value are input, this counter counts up to approximate to the data.

When image data smaller (darker in light) than the current value are input, this counter counts down to approximate to the data.

- Setting of the rate of count-up/count-down following data input: register 0C (MAX_UP, MAX_DOWN)
- Setting of the lowest limit for background levels: register 0E (LL_MAX)

• Character level counter

When image data greater (lighter in light) than the current value are input, this counter counts up to approximate to the data.

When image data smaller (darker in light) than the current value are input, this counter counts down to approximate to the data.

- Setting of the rate of count-down following data input: register 0C (MIN_UP)
- Setting of the highest limit for character levels: register 0D (UL_MIN)

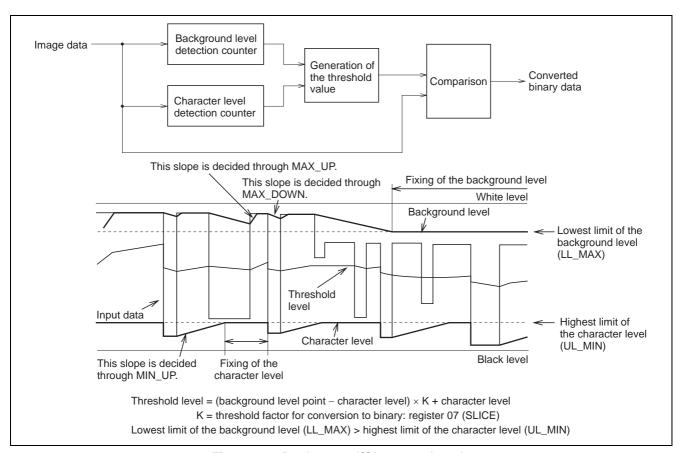


Figure 13 Background/Character Levels

Error Diffusion

The error diffusion, which is a conditional determination method, locally diffuses density errors between the original image and the result to obtain the best approximation. This generates images with good compatibility of gradation and resolution.

This is operated by selecting the error diffusion in conversion-into-binary mode selection.

: register 07 (MODE)

In error diffusion, dithers as well as density errors are added to image data. The dithers are data as commonly used for the dither matrix.

: register 08 (ERROR)

γ correction must be performed in the error diffusion.

· Organized dither

The M66335 has built-in SRAM with a configuration of 64 words \times 6 bits for organized dither memory. In the initial setting, write the threshold value proper for the preferred dither pattern to the dither memory after setting the dither matrix size.

: register 07 (DITH)

: register 10 (DITH_D)

For the procedure of inputting/outputting data, refer to the section on writing to/reading from the dither memory.

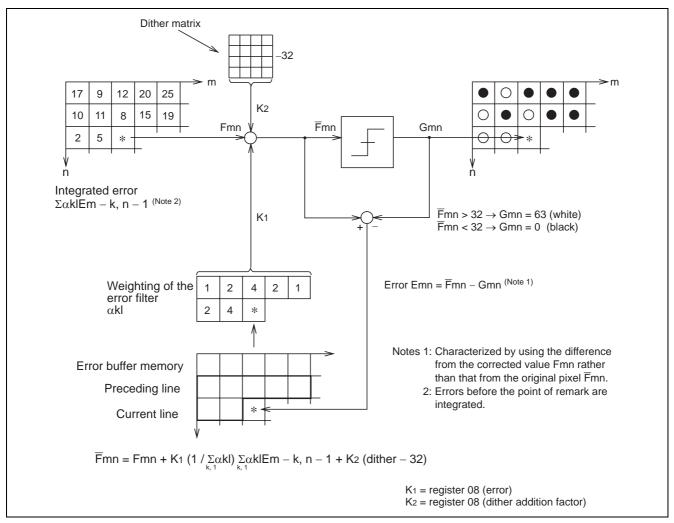


Figure 14 Error Diffusion Method

Image Zone Separation

To make data conversion fit for each image zone, a black and white image is separated into the zones to be converted to binary and the gradation zones. The binary zone is processed through simple conversion to binary and the gradation zone through the error diffusion.

: register 08 to 0E

In the black and white image, each window of the gradation zone (photo) does not have a large difference of luminance in it. With this characteristic of the gradation zone, it is distinguished from the conversion-into-binary zone through the following method. Lmax: maximum illumination in window Lmin: minimum illumination in window Determining inequality 1: Lmax - Lmin > A (because the zone to be converted to binary has a large difference in luminance in it.): register 09 Difference (SEPA_A) Lmin > B (for the wholly white area): register 0A Minimum (SEPA_B) Determining inequality 2: Determining inequality 3: Lmax < C (for the wholly black area): register 0B Maximum (SEPA_C) If the window satisfies determination inequalities 1, 2 or 3, simple conversion to binary is applied. If the window does not satisfy any of determination inequalities 1, 2 and 3, change to pseudo-halftone is applied. Difference Minimum Maximum White level = 63Lmax Input data В Black level = 0 Lmax Lmax Lmin

Figure 15 Image Zone Separation

Region Designation Function

The sensor width is fixed for A4 and B4.

The region designation function is to output only the data for a region defined and designated in terms of output data after resolution change (or after uniformity correction for multivalued data).

Registers 11 to 14 (OFFSET, OUTLENGTH)

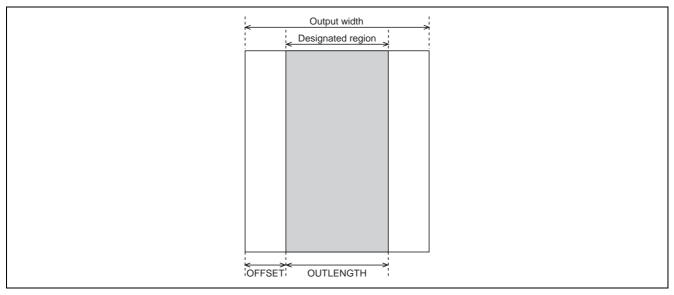
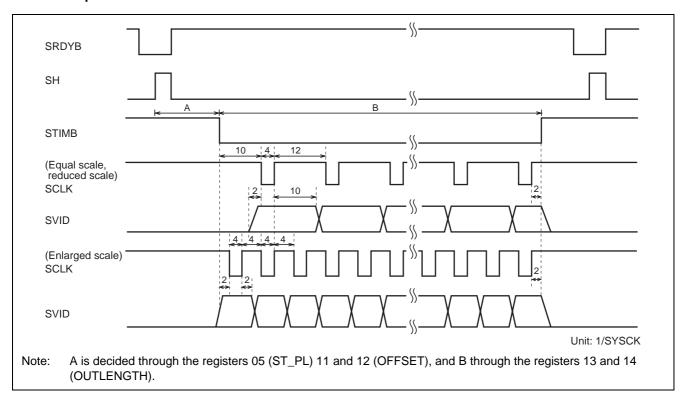


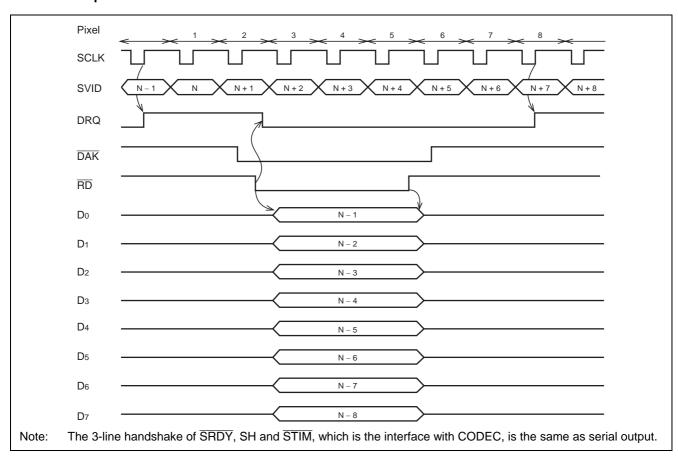
Figure 16 Cut-out Function

(4) CODEC Interface (Binary Data Output)

Serial Output

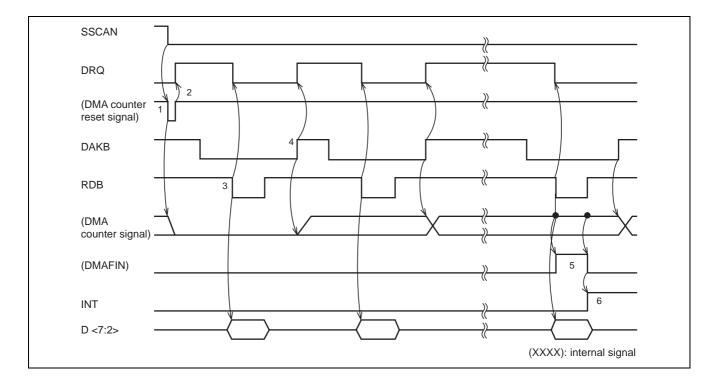


Parallel Output



(5) DMA Interface (Multivalued Output)

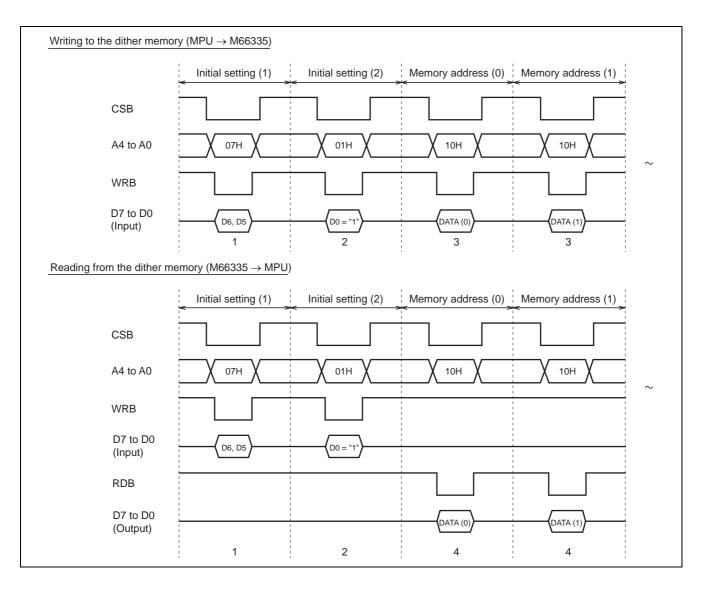
The DMA transfer of data after non-uniformity correction can be performed by setting P_O) of the register 01: at "1" (existence of DMA output) and M_B of that register at "1" (multivalue). With this setting, neither enlargement, nor reduction, nor 400 dpi of resolution can be set.



- 1. On completion of reading one line, with a \downarrow flow of SSCAN, the reset signal is entered in the DMA counter.
- 2. With a ↑ flow of the reset signal, DRQ shifts to "H", when the DMA transfer becomes ready.
- 3. With DAKB at "L" and a \downarrow flow of RDB, DRQ shifts to "L", when multivalued data are output to D <7:2> during the period that RDB is at "L".
- 4. With a ↑ flow of DAKB, the DMA counter counts up and DRQ shifts to "H", when the DMA transfer becomes ready again.
- 5. The cycle of the above 3 and 4 is repeated until the DMA counter counts up to reach the number of output pixels set in the registers 13 and 14 OUTLENGTH subtracted by one. By that repetitive operation, DMAFIN shifts to "H" to terminate the DMA transfer when it reaches the set number.
- 6. With a \downarrow flow of DMAFIN, INT shifts to "H", when CPU has an interrupt.
- 7. Reading is resumed from the next line by negating the INT signal through the register 17 (INTCLR).

(6) Writing to/Reading from the Dither Memory, γ Correction Memory, Uniformity **Correction Memory, and Resolution Change Memory**

The sequences of writing a dither pattern to and reading it from SRAM with a configuration of 64 words × 6 bits which is built in the M66335 for organized dither are shown below.



- 1. D6 and D5 (DITH) of the register 07 are set to define the dither matrix size.
- 2. D0 (CNTRST) of the register 01 is set at "1" to reset the address counter of the dither memory.
- 3. DITH D is selected in the register 10, and DATA (0) of the MPU bus (D5 to D0) is written in the memory. The address counter of the dither memory is incremented at the edge of the first transition of \overline{WR} . (For writing)
- 4. DITH_D is selected in the register 10, and DATA (0) of the dither memory is read into the MPU bus (D5 to D0). The address counter of the dither memory is incremented at the edge of the first transition of \overline{RD} . (For reading)

RENESAS

Dither Matrix Addresses

A0	A1	A2	А3		
A4	A5	A6	A7		
A8	A9	A10	A11		
A12	A13	A14	A15		

4		4	Mat	riv
4	х	4	wat	гіх

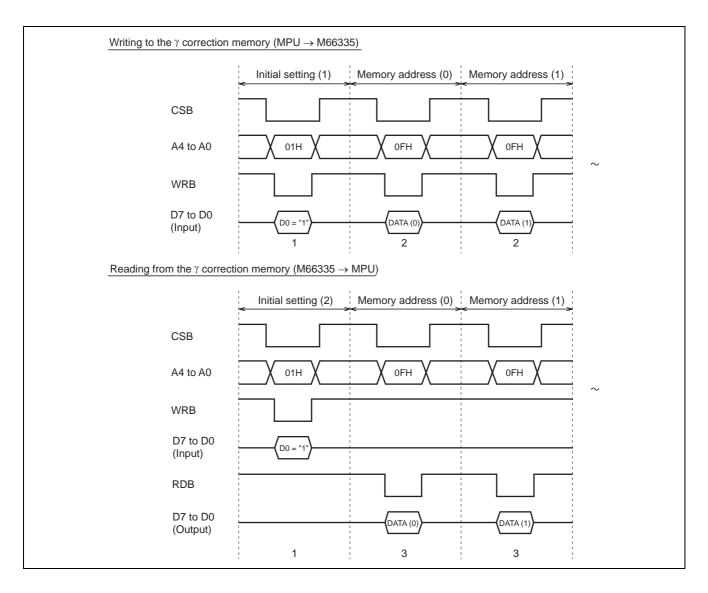
A0	A1	A2	А3	
A4	A5	A6	A7	
A8	A9	A10	A11	
A12	A12 A13		A15	
A16	A17	A18	A19	
A20	A21	A22	A23	
A24	A25	A26	A27	
A28	A28 A29		A31	

4 × 8 Matrix

A0	A1	A2	A3	A4	A5	A6	A7
A8	A9	A10	A11	A12	A13	A14	A15
A16	A17	A18	A19	A20	A21	A22	A23
A24	A25	A26	A27	A28	A29	A30	A31
A32	A33	A34	A35	A36	A37	A38	A39
A40	A41	A42	A43	A44	A45	A46	A47
A48	A49	A50	A51	A52	A53	A54	A55
A56	A57	A58	A59	A60	A61	A62	A63

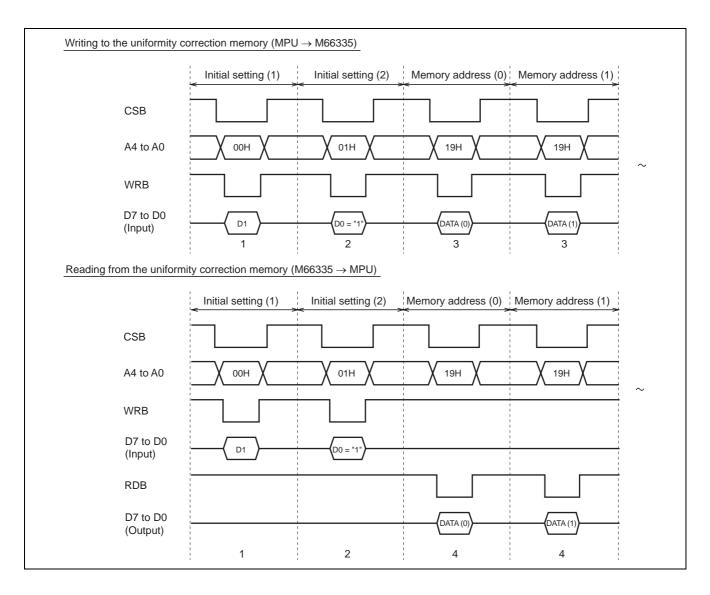
8 × 8 Matrix

The sequences of writing γ correction table to and reading it from SRAM with a configuration of 64 words \times 6 bits which is built in the M66335 for γ correction are shown below.



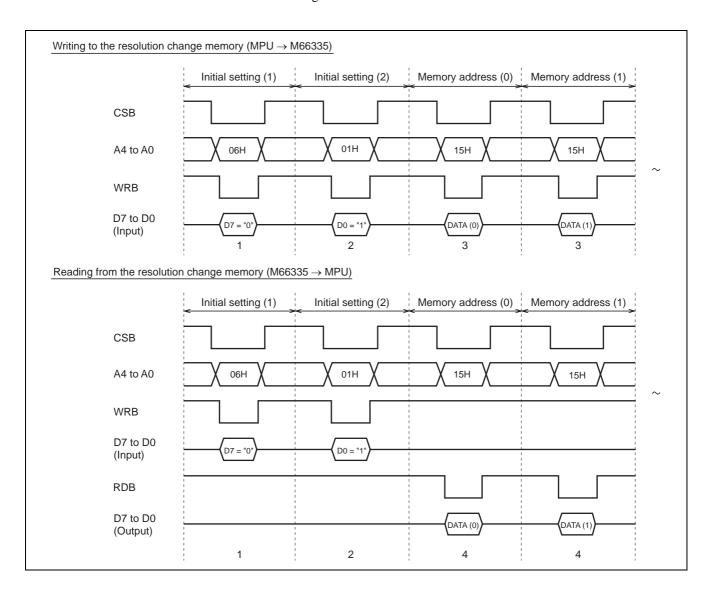
- 1. D0 (CNTRST) of the register 01 is set at "1" to reset the address counter of the γ correction memory.
- 2. GAMMA_D is selected in the register 0F, and DATA (0) of the MPU bus (D5 to D0) is written in the memory. The address counter of the γ correction memory is incremented at the edge of the first transition of WRB. (For writing)
- 3. GAMMA_D is selected in the register 0F, and DATA (0) of the γ correction memory is read into the MPU bus (D5 to D0). The address counter of the γ correction memory is incremented at the edge of the first transition of RDB. (For reading)

Uniformity correction data can be written to and read from SRAM for uniformity correction built in the M66335 through the MPU bus. With this operation, the uniformity data can be temporarily saved in the backup memory when the power is off. The sequences of writing and reading uniformity correction data are shown below.



- 1. "0" (black correction) or "1" (white correction) is set in D1 (Umode) of the register 00.
- 2. D0 (CNTRST) of the register 01 is set at "1" to reset the address counter of the uniformity correction memory.
- 3. UNIF_D is selected in the register 19, and DATA (0) of the MPU bus (D5 to D0) is written in the memory. The address counter of the uniformity correction memory is incremented at the edge of the first transition of WRB. (For writing)
- 4. UNIF_D is selected in the register 19, and DATA (0) of the uniformity correction memory is read into the MPU bus (D5 to D0). The address counter of the uniformity correction memory is incremented at the edge of the first transition of RDB. (For reading)

The sequences of writing a resolution change table to and reading it from SRAM with a configuration of 100 words \times 1 bit which is built in the M66335 for resolution change are shown below.



- 1. "0" (horizontal scan) is set in D7 (MSSEL) of the register 06.
- 2. D0 (CNTRST) of the register 01 is set at "1" to reset the address counter of the resolution change memory.
- 3. CNV_D is selected in the register 15, and DATA (0) of the MPU bus (D0) is written in the memory. The address counter of the resolution change memory is incremented at the edge of the first transition of WRB. (For writing)
- 4. CNV_D is selected in the register 15, and DATA (0) of the resolution change memory is read into the MPU bus (D0). The address counter of the resolution change memory is incremented at the edge of the first transition of RDB. (For reading)

List of the M66335FP Registers

R/W	A4 to A0	Default	D7	D6	D5	D4	D3	D2	D1	D0		
R/W	00H	00H	RESET	SENS	SENS_W	AGC	UNIF	SCAN	UMODE	"L"		
R/W	01H	00H	SOURCE	S/H_W	SH_W	UNIFS	P_O	M_B	UNIFM	CNTRST		
W	02H	00H	RES	LCMPS	BLS	BLCMPS	CCD	CIS3	CIS2	CIS1		
W	03H	00H				PRE_DA	ATA (7:0)					
W	04H	00H					PRE_DA	TA (13:8)				
W	05H	00H				ST_P	L (7:0)					
W	06H	00H	MSSEL	AVE	CONV	< <1:0>	CONV'	Y <1:0>	GAMM	A <1:0>		
W	07H	00H	POL	DITH	<1:0>	MODE	<1:0>		SLICE <2:0>			
W	08H	00H			ERROF	R <1:0>	MTF_C	C <1:0>	MTF_	<1:0>		
W	09H	00H					SEPA_	_A (5:0)				
W	0AH	00H					SEPA_	_B (5:0)				
W	0BH	00H					SEPA_	_C (5:0)				
W	0CH	00H			MAX_U	P <1:0>	MAX_DO	WN <1:0>	MIN_U	P <1:0>		
W	0DH	1FH					UL_MII	N <5:0>				
W	0EH	20H					LL_MA	X <5:0>				
R/W	0FH	_					GAMMA	_D (5:0)				
R/W	10H	_					DITH_	D (5:0)				
W	11H	00H				OFFSE	T <7:0>					
W	12H	00H					C	FFSET <12:8	3>			
W	13H	00H			OUTLENGTH <7:0>							
W	14H	00H			OUTLENGTH <12:8>							
W	15H	_								CNV_D		
R/W	16H	00H						AGCSTP	SRDYS	SRDYB		
W	17H	_		INTCLR								
R/W	18H	00H				GAIN	<7:0>					
R/W	19H	00H					UNIF_I	D <5:0>				

Register Structure

Address	R/W	Description										
00 _H	R/W	D7 D6 D5 D4 D3 D2 D1										
		RESET	SENS	SENS_W	AGC	UNIF	SCAN	UMODE	"L"	(Default value: 00 _H)		
			_							_		
		D7	RI	stem Rese	et .	Wit	th D7 = 1, t	he syste	em is reset during the			
			Normal m			· <u>·</u>				oulse is "L".		
		1	Reset mod				(*)	Write only				
							1					
		D6	9	SENS: Ser	sor Type		1					
		0	CCD									
		1		of clock d	uty)							
							<u>.</u>					
		D5	SENS W	/: Reading	. Width of	the Sens	or					
		0	A4		, ,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,,		<u>. </u>					
		1	B4									
		D4		AGC: AG	C Mode		1 Co	ntrols start	stop of	the AGC mode.		
		0	Stop	7.00.7.0	- Inouc							
		1	Start									
							_					
		D3		UNIE: UN	F: UNIF Mode			Controls start/stop of the UNIF mode.				
		0	Stop	Oltin . Olt	ii iiioac		1					
		1	Start				-					
							1					
		D2		SCAN: SC	AN Mode] Co	ntrols start	stop of	the SCAN mode.		
		0	Stop	SCAN. SC	AN WIOGE		-	THE OIL CLAIR	otop of	and Corar mode.		
		1	Start				1					
							J					
				UMODE: U	Iniformity	, Correcti	on in the	LINIE Mo	do			
		D1		orrection				ly White C		on		
		0	Black corr			0110011011		ily milito c	70110011	<u> </u>		
		1	White corr				White	e correction	n			
01 _H	R/W	D7	D6	D5	D4	D3	D2	D1	D0			
		SOURC			UNIFS	P_O	M_B	UNIFM	CNTRS	T (Default value: 00 _H)		
					1	1				(= 0.000.000.000.000.000.000.000.000.000.		
		D7	SOLIB	RCE: Read	ing Width	of the O	riginal					
		0	A4	CE. Reau	ing width	or the Or	iginai					
		1	B4									
		'	54									
		D.C.	T	0/11 14	CAN Deal	- \A/: -14!-		7				
		D6	Normal		S/W Puls		rolo)	-				
		1		(quadruple multiplied b		II CIOCK CY	cie)					

Address	R/W	Description													
01 _H	R/W	D5		SH W.	SH Pulse	- Widtl	`		1						
		0	Normal (*					<u> </u>							
		1	Reverse					<u>"</u>							
			11010100	or morrina	manapin	ou by 2			_						
		D4	UNIF	S: Unifo	rmity Co	rrection	on .								
		0	Valid												
		1	Invalid												
								_							
		D3			MA Outp	out				•	he form o	f LSB and D7 in the			
		0	Without D		out			_ 10	orm c	of MSB.					
		1	With DM	A output											
		Do	M_B: Processing Mode					¬ ,	ال ۱۸/ith	ho multiva	luo coloo	ted, data (6-bit) after			
		D2		_B: Proc	essing	wode						can be output through			
		1	Binary Multivalue	3						MA transfe		9			
		'_	ividitivatu												
		D1 UNIF: Uniformity Correction in SCAN							1						
		0 White correction													
		1	Black cor	rection +	white co	rrectio	n								
		<u> </u>													
		D0							With D0 = 1, the counter is reset during the						
		0	Normal mode					period that the write pulse is "L". All the built-in RAM addresses are reset.							
		1	Reset mo	de						ite only	Aivi addie	esses are reset.			
02 _H	W									-					
		D7	D6	D5			D3		D2	D1	D0] ₍₂ , , , , , , , , , , , , , , , , , , ,			
		RES	LCMP	S BLS	BLC	MPS	CCD	С	IS3	CIS2	CIS1	(Default value: 00 _H)			
		D7	DI	C. Dane			Г	DC	1	LON	IDC: Line	Clamping			
		D7	200 dpi	S: Resc	olution		-	D6	Inv	alid	IPS: Line	Clamping			
		1	400 dpi				-	1	Val						
			.00 ap.				<u> </u>	•							
		D5 BLS: Bit Clamping						D4	В	LCMPS: E	Black Lev	vel Line Clamping			
		0	Invalid					0	+	alid		<u> </u>			
		1	Valid					1	Val	lid					
		D3	D2	D1	D0	S	ensor	s Co	mpat	ible with	Image Se	ensor Interfaces			
		0	0	0	1							or higher			
		0	0	1	0					ne input le		ler 2 V			
		0	1	0	0			ors c	apab	le of line of	lamping				
		1	0	0	0	CCD)								

Address R/W Description										
03 _H	W	D7	D6	D5	D4	D3	D2	D1	D0	
						TA <7:0>				(Default value: 00 _H)
		D7 to D	0: PRE_[ounter value						
04 _H	W	D7	D0	D.F.	D.4	D0		D4	D0	
		D7	D6	D5	D4	D3 PRE_DAT	D2	D1	D0	(Default value: 00 _H)
		D5 to D	0: PRF 1	DATA <13:8	s> the highe			ne single-li	ne cycle	counter value
05 _H	W		_							
OOH	VV	D7	D6	D5	D4	D3	D2	D1	D0	
					ST_PL	_<7:0>				(Default value: 00 _H)
		D7 to D	0: ST_PL	<7:0> start	t pulse pos	ition to the	sensor			
			Set S7	_PL = (dun	nmy pixels	of the sen	sor + 2).			
06 _H	W	D7	DC	Dr	D4	Do	Do	D4	Do	
		D7 MSSEL	D6 AVE	D5	D4 DNVX	D3 CON	D2	D1	D0 MMA	(Default value: 00 _H)
		IVISSEL	- AVE		JNVA	COI	NV I	GAI	VIIVIA	(Default value: 00 _H)
		D7		L: Horizon	tal and Ve	rtical Sett	ing			
		0	Horizon							
		1	Vertical							
		De	A\/E. A	waraning F)reessing	. Whe	n "with a	veraging" (salactac	ı.
		D6 AVE: Averaging Processing When "with averaging" sell For enlargement: inserted								
		1		averaging			_	prece	ding one	e and the current one.
			· · · · · · · · · · · · · · · · · · ·	avoraging		— For i	eduction			ines from removed lines
								the curre	_	of the removed one and
				CONVY	Enlargeme	ont/Poduc	stion Mo	do	RES =	. 1
		D5	D4		orizontal S				_	ne setting of 400 dpi,
		0	0	Original so	ale				enlarg	ement cannot be set.
		0	1	Enlargeme	ent					
		1	0	Reduction						
		1	1							
					Enlargeme					
		D3	D2		orizontal S	scanning	Directio	n		
		0	0	Original so Enlargeme						
		1	0	Reduction						
		1	1	ACCUCION						
		<u>'</u>	_ '							

Address	R/W					Des	cription						
06 _H	W	D1	D0		GAMMA:	/ Correcti	on Proc	essina]			
		0	0	Characte	r, photo: $\gamma =$			<u>-</u>					
		0	1		r, photo: $\gamma =$		value			-			
		1	0		$\gamma = 1$; phot			alue		-			
		1	1		$\gamma = \text{downlo}$					-			
		Note: II	ıdament						of image	zone separation.			
		14010. 00	adginioni	DOTWOOT OF	iaraotor ario	i prioto is i	basea on	Ture result (Ji iiilage	, zone separation.			
07 _H	W	D7	D6	D5	D4	D3	D2	D1	D0				
		POL		DITH	MC	DE		SLICE		(Default value: 00 _H)			
					!					_			
		D7 POL: Conversion-to-Binary Output Mode											
		0		t; black: 0	n-to-binary	Output	viode						
		1); black: 0); black: 1									
		'	vviiite. (D, DIACK. I									
		D6	D5		H: Dither N	latrix Size	9						
		0	0	4 × 4									
		0	1	4 × 8									
		1	0	8 × 8									
			1	_									
		D4	D3	MODE:	Selection o	f the Con	version-	to-Binary	Mode				
		0	0	Simple bi	nary								
		0	1	Organize	d dither								
		1	0	Image zo	ne separatio	n (simple	binary +	error diffus	ion)				
		1	1	Error diffu	ısion								
		D2	D1	D0	SLICE	: Thresh	old Fact	or for Con	ersion	to Binary			
		0	0	0	6/16					-			
		0	0	1	7/16								
		0	1	0	8/16								
		0	1	1	9/16								
		1	0	0	10/16								
		1	0	1	11/16								
		1	1	0	12/16								
		1	1	1	13/16								

Address	R/W					Des	cription			
08 _H	W	D7	D6	D5	D4	D3	D2	D1	D0	
					RROR		F_C	MTI		(Default value: 00 _H)
						_				
						ERRC	R			7
		D5	D4	Error (Base)	_				
		0	0	Strong (7/						
		0	1	Strong (7/	8)	Strong (1	/4)			
		1	0	Weak (3/4) Weak (1/8)						
		1	1	Weak (3/4	!)	Strong (1	/4)			
		D3	D2	MTF_C: I	MTF Com	pensation	Factor			
		0	0	1/4						
		0	1	1/2						
		1	0	1						
		1	1	0						
		Note: Th	nis is vali	d when MC	DE is sim	ple binary o	or image z	zone separa	ation (c	haracter).
		D1	D0	MTF_I: N	ITF Comp	ensation	Factor			
		0	0	1/4						
		0	1	1/2						
		1	0	1						
		1	1	0						
		Note: Th	nis is vali	d when MC	DE is orga	anized dithe	er, error d	iffusion or i	image z	zone separation (photo).
09 _H	W									
		D7	D6	D5	D4	D3	D2	D1	D0	7.5 ()
							PA_A			(Default value: 00 _H)
		D5 to D0): SEPA	_A Image z	zone sepa	ration parai	meter (dif	ferential)		
0A _H	W	D7	D6	D5	D4	D3	D2	D1	D0	
							PA_B			(Default value: 00 _H)
		D5 to D0	D: SEPA	_B Image z						
0B _H	W	D7	D6	D5	D4	D3	D2	D1	D0	
		D7	D0	D3	D4		PA_C	וט	טט	(Default value: 00 _H)
		D5 to D/	J. SED^	C Imaga	7000 0000	ration para		ovimum\		(Bolault value, OOH)
		טט נט ט(J. SEPA	_C image 2	zone sepa	ration para	meter (ma	axiiiiulii)		

Address	R/W					Desc	cription			
0C _H	W	D7	D6	D5	D4	D3	D2	D1	D0	
					X_UP	MAX_C		MIN		(Default value: 00 _H)
				100.0	<u> </u>	101.01_2	,0,111	IVIIIV	_0.	(Doladit Value: OOH)
		D5	D4					n Clock	for the	Up Counter
		0	0	Ordinary (7						
		0	1	Slow $(T = (T + T)^T)$						
		1	0	Fast (T = (
		1	1	Fastest (T	= (single p	ixei cycie)	× 8)			
		D3	D2	MAX DO	NN: Backo	round Le	vel Detec	tion Clo	ck for t	he Down Counter
		0	0	Ordinary (
		0	1	Slow (T = (
		1	0	Fast (T = (
		1	1	Fastest (T						
				`			•			
		D1	D0	MIN_UI	P: Charact	er Level D	Detection	Clock fo	or the U	p Counter
		0	0	Ordinary (7	Γ = (single	pixel cycle) × 128)			
		0	1	Slow (T = (single pixe	l cycle) × 2	256)			
		1	0	Fast (T = (single pixe	l cycle) × 6	64)			
		1	1	Fastest (T	= (single p	ixel cycle)	× 32)			
0D _H	W	D7	D6	D5	D4	D3	D2	D1	D0	
						UL_I	MIN			(Default value: 1F _H)
		D5 to D0	D: UL MI	N Detection	n of backar	ound/char	acter leve	ls		<u> </u>
			_		limit of cha					
0E _H	W									
		D7	D6	D5	D4	D3	D2	D1	D0	
						LL_N				(Default value: 20 _H)
		D5 to D0): LL_M/	AX Detectio	_			ls		
					imit of back	-				. (111 - 141)
		Lowest	imit of ba	ackground is	eveis (LL_i\	/IAX) > nig	nest limit	of charac	ter leve	ls (UL_MIN)
0F _H	R/W	D7	D6	D5	D4	D3	D2	D1	D0	
						GAMMA_				
		D5 to D0): GAMM	IA_D Built-i	n γ memor					<u> </u>
10 _H	R/W	D7	D6	D5	D4	D3	D2	D1	D0	
		707	00	סט	<i>υ</i> 4	DITH_C		וט	טט	
		DE 1: 50). DIT''	D D.::14 :	:41= = :		<0.0>			
		טט נס ט(): חווט 	D Built-in d	ither memo	ory data				

Address	R/W					Des	cription						
11 _H	W	D7	D6	D5	D4	D3	D2	D1	D0				
					OFFSE	T <7:0>				(Default value: 00 _H)			
		D7 to D0:	OFFSET	<7:0> Off	set for cut	t-out Lowe	est order 8	3 bits					
12 _H	W	D7	D6	D5	D4	D3	D2	D1	D0				
			OFFSET <12:8> (Default value: 00 _H)										
		D3 to D0:	to D0: OFFSET <12:8> Offset for cut-out Highest order 5 bits										
13 _H	W	D7	D6	D5	D4	D3	D2	D1	D0	_			
			OUTLENGTH <7:0> (Default value: 00 _H)										
		D7 to D0:	OUTLEN	GTH <7:0	> No. of o	utput pixe	ls Lowest	t order 8 k	oits	_			
14 _H	W	D7	D6	D5	D4	D3	D2	D1	D0				
						OUTI	LENGTH <	12:8>		(Default value: 00 _H)			
		D3 to D0:	OUTLEN	GTH <12:	8> No. of	xia tuatuo	els Highe	est order 5	5 bits	_			
										els is not a multiple of 8,			
			e remainde			-				,			
15 _H	R/W	D7	D6	D5	D4	D3	D2	D1	D0				
									CNV_D				
		DO: CNIV	D0: CNV_D Indication of enlargement/reduction										
		DU. CINV_	_ט ווועונעו	IOH OF EIN	aigeillelli/	TEGUCION							

Address	R/W					Des	cription						
16 _H	R/W	D7	D6	D5	D4	D3	D2	D1	D0				
							AGCSTP	SRDYS	-	(Default value: 00 _H)			
								l					
							İ						
		D2			ontrol Co	unter							
		0	Gain contro Gain fixed.		valid.								
		1	Gain fixed.										
		D1	ę pi	NG. GDI	OY Contro	N.							
		0	SRDY conf										
		1	SRDY con										
		<u> </u>	OND FOOT	aror amoug	jii tiio oxto	orrial pill							
		D0	SRDYB: [Data Tran	sfer Start	Ready	In t	he case	of data co	ntrol through the			
		0	Transfer al							nput pin must be			
		1	Transfer no	ot allowed			always set at "H".						
									-	th the register, the be controlled line by			
							line		ilei iiiusi i	be controlled lifte by			
								 Write on	ly				
17 _H	W	D7	D6	D5	D4	D3	D2	D1	D0				
						CLR							
		INT sig	nals are neg	ated by a			ress.			I			
10	R/W												
18 _H	K/VV	D7	D6	D5	D4	D3	D2	D1	D0				
					GAIN	<7:0>							
		In readi	ing: the curre	ent gain v	alue of the	e gain cor	trol counte	r can be	read.				
		In writin	ng: the gain		-			set.					
			Howeve	r, this is v	alid only i	f AGCSTF	P = 1.						
19 _H	R/W	D7	D6	D5	D4	D3	D2	D1	D0	With UMODE = 0,			
		D7	D6	DS	D4		<5:0>	וט	D0	access to the			
		DE to D	O. LINIE D	Duilt in u	niformity o			***		uniformity correction			
		D5 t0 D	00: UNIF_D	Buiit- in u	niiormity c	correction	memory da	ita		memory for black			
										correction is available.			
										With UMODE = 1, access to the			
										uniformity correction			
										memory for white			
										correction is available.			
	1												

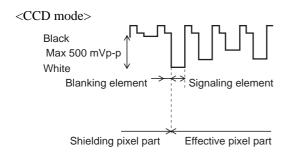
Description of the Operations of the Analog Circuits

The configuration of the analog processing circuits is shown in figure 17.

(1) Sensor Selection Circuit

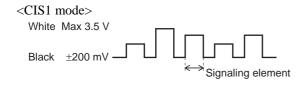
The four types of sensors in the table can be connected to the circuit.

Register 02 _H	Sensor Type
CCD	CCD sensor
CIS1	CIS sensor which outputs light voltages (white voltage) of 3.5 V or lower
CIS2	CIS sensor which outputs light voltages (white voltage) of 2 V or lower
CIS3	CIS sensor which output shielding pixels for each line



The amplitudes of sensor signals are multiplied by –4 through the two operating amplifiers directly after the switch to select the CCD mode. (The waveforms of the signals are inverted at the same time.)

As a result, the sensor signals input to the sample and hold circuit have a dark voltage of 2.2 V.



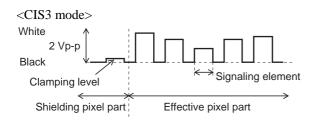
The amplitude of signals input from the sensor are halved. Then, their reference potential is shifted up to 2.2 V.

As a result, the sensor signals input to the sample and hold circuit have a dark voltage of 2.2 V.



The reference potential of signals input from the sensor is shifted up to 2.2 V.

As a result, the sensor signals input to the sample and hold circuit have a dark voltage of 2.2 V.



Sensor signals with a dark voltage of 2.2 V clamped by line clamping input are directly input to the sample and hold circuit.

(2) Line Clamping Circuit

This circuit is used for CCD (line clamping mode) and CIS3.

The reference voltage (dark voltage) output in the shielding pixel part of the sensor is sampled by LCMP (line clamping pulses) and shifted up to the internal reference voltage of 2.2 V. This is not used for the CIS1 or CIS2 input sensor (set off constantly).

: register 02 (LCMPS)

(3) Sample and Hold Circuit and Bit Clamping Circuit

In the CCD mode, bit clamping, as well as line clamping, can be performed. The blanking elements of each pixel of sensor output is sampled by BTCMP (bit clamping pulses). The differences of signals from the reference potential sampled by the bit clamping circuit are input to the gain control circuit of next step as signaling elements. To turn off bit clamping, set BLS invalid, so that the reference potential will be fixed at the internal reference potential of 2.2 V.

: register 02 (BLS)

(4) Gain Control Circuit

The amplifying factor (gain) must be adjusted so that the amplitudes of sensor signals can come within the dynamic range of the A/D converter. The gain is set through the automatic gain control in the AGC mode (register 00) or directly through the register 18 (GAIN <7:0>).

The gain changes within the following ranges according to the sensor used.

Mode	Amplifying Factor of Signals (Gain)						
CCD	4 to 20						
CIS1	0.5 to 2.5						
CIS2	1 to 5						
CIS3	1 to 5						

In the AGC mode, the gain control counter is set at the greatest gain in the initial state and then counted down each time an overflow bit is output from the A/D converter. The count (gain) of the gain control counter is directly read/written through the register 18 (GAIN <7:0>). The counting operation of the counter can be controlled through the register 16 (AGCSTP).

(5) Internal Reference Voltage

Internal reference voltage source for the analog circuits:

this generates the reference voltage (2.2 V) for the line clamping circuit, the sample and hold circuit, and the bit clamping circuit.

A/D converter reference voltage generation circuit:

this generates VWL (white level reference voltage of 3.8 V) and VBL (black level reference voltage of 1.8 V) for the A/D converter.

(6) Black Level Clamping Circuit

This circuit adjust the level of reference voltage to the A/D converter from analog circuits.

The black clamping circuit is used in the CCD or CID3 mode. (See figure 18, 19 and 22) The GCAO pin and the BCMI pin are capacity-coupled. The output reference potential in the shielding pixel part of sensor signals are applied to the BCMV pin as the VBL (black level reference voltage of 1.8 V) for the A/D converter.

BLCMP (black level clamping pulses) are generated concurrently with the shielding pixel part of each line. To turn off this circuit, set BLCMPS invalid and apply the black level reference voltage of the A/D converter to the BCMV pin.

: register 02 (BLCMPS)

In the CIS1 or CIS2 mode, the LEVAJ pin is used. (See figure 20 and 21) Voltage is applied to the LEVAJ pin so that the reference potential of output at the GCAO pin can be adjusted to the VBL (black level reference voltage of 1.8 V) of the A/D converter. Set voltage input to the LEVAJ pin as follows.

VLEVAJ = VVBL
$$- A \times G_V + 0.2 [V]$$

VGCAO = VLEVAJ + $G_V \times V_{IN} [V]$

where,

A: the lowest limit of dark voltage of the sensor [V]

G_V: gain (multiplying factor) of the gain control circuit

 V_{IN} : signals input from the sensor [V]

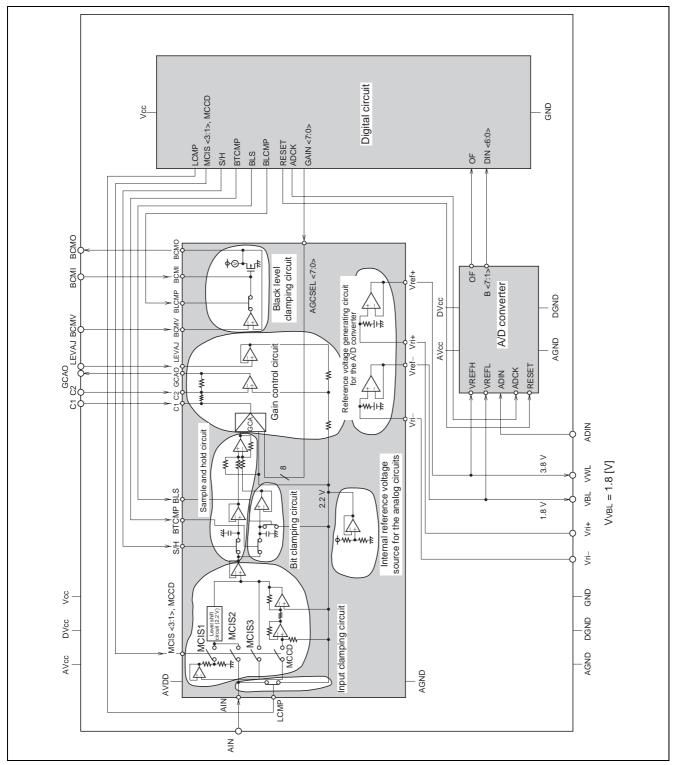
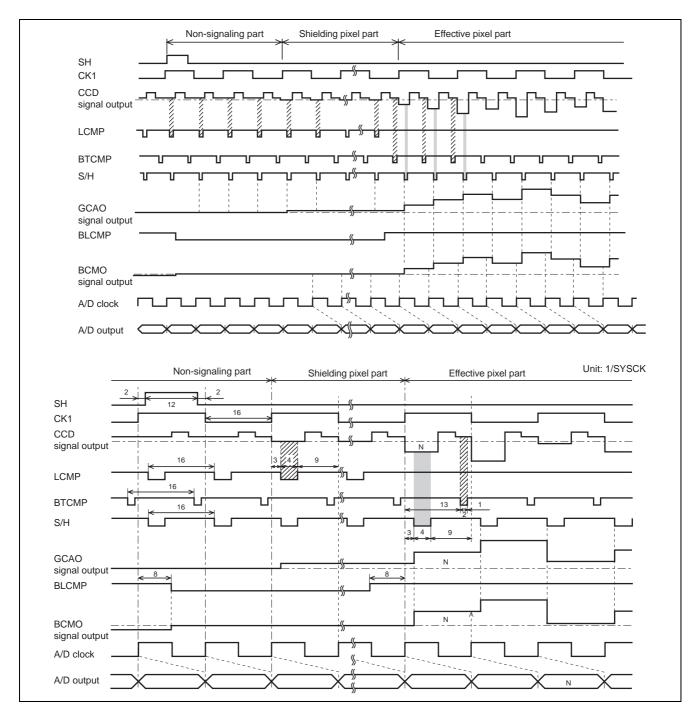


Figure 17 Circuit Configuration of the Analog Part of the M66335FP

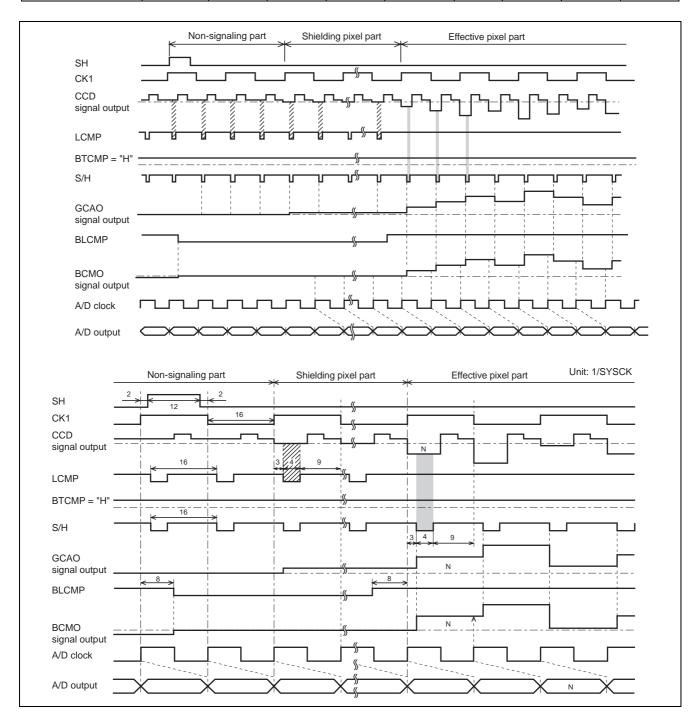
Analog Circuit Timing Chart (for CCD Mode/Bit Clamping)

Register	Address	00H	02H							
	Bit	D6	D6	D5	D4	D3	D2	D1	D0	
Mode	Signal	SENS	LCMPS	BLS	BLCMPS	CCD	CIS3	CIS2	CIS1	
CCD (bit clamping)	Setting	1	1	1	1	1	0	0	0	



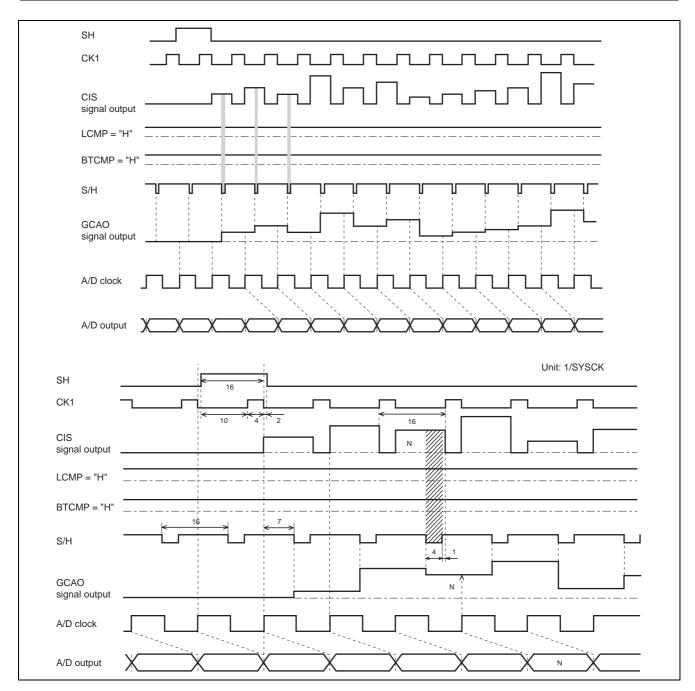
Analog Circuit Timing Chart (for CCD Mode/Line Clamping)

Register	Address	00H	02H							
	Bit	D6	D6	D5	D4	D3	D2	D1	D0	
Mode	Signal	SENS	LCMPS	BLS	BLCMPS	CCD	CIS3	CIS2	CIS1	
CCD (line clamping)	Setting	1	1	0	1	1	0	0	0	



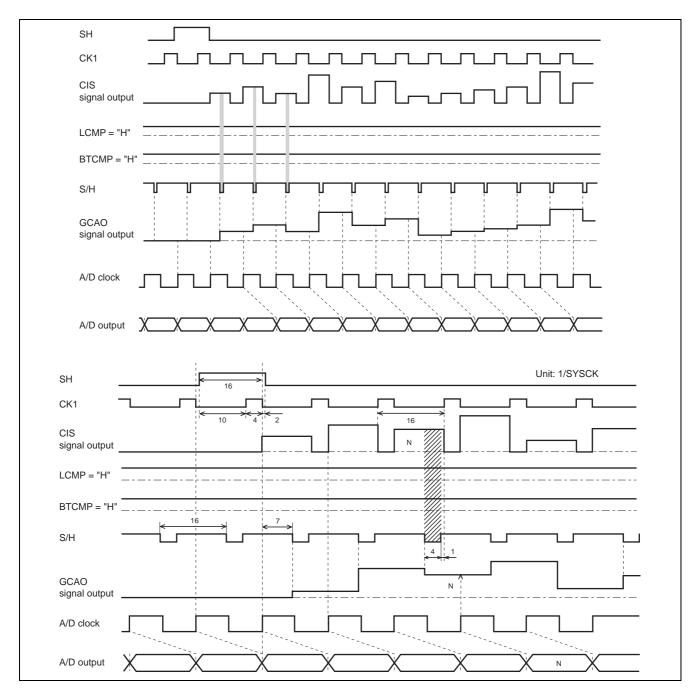
Analog Circuit Timing Chart (for CIS1 Mode)

Register	Address	00H	02H							
	Bit	D6	D6	D5	D4	D3	D2	D1	D0	
Mode	Signal	SENS	LCMPS	BLS	BLCMPS	CCD	CIS3	CIS2	CIS1	
CIS1	Setting	0	0	0	0	0	0	0	1	



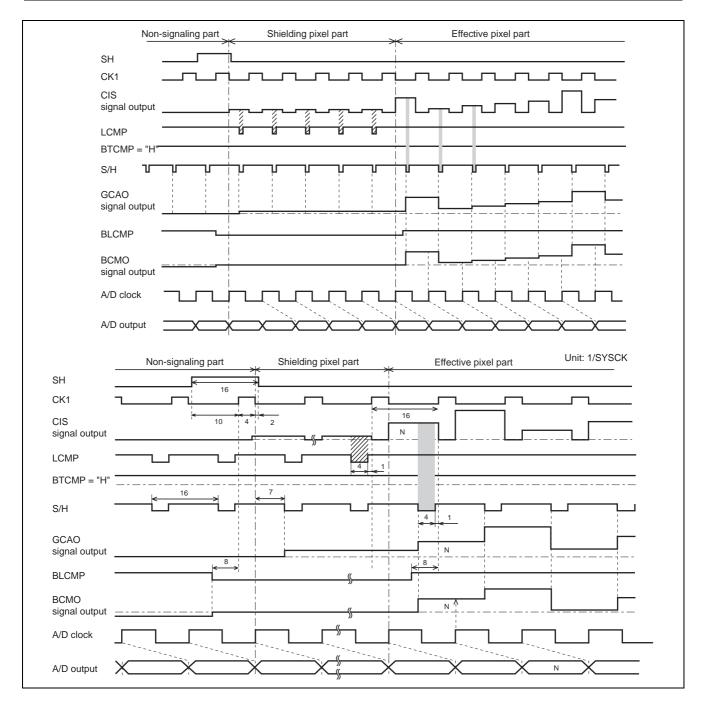
Analog Circuit Timing Chart (for CIS2 Mode)

Re	egister	Address	00H	02H							
	_	Bit	D6	D6	D5	D4	D3	D2	D1	D0	
Mode		Signal	SENS	LCMPS	BLS	BLCMPS	CCD	CIS3	CIS2	CIS1	
CIS2		Setting	0	0	0	0	0	0	1	0	



Analog Circuit Timing Chart (for CIS3 Mode)

Register	Address	00H	02H								
	Bit	D6	D6	D5	D4	D3	D2	D1	D0		
Mode	Signal	SENS	LCMPS	BLS	BLCMPS	CCD	CIS3	CIS2	CIS1		
CIS3	Setting	1	1	0	1	0	1	0	0		



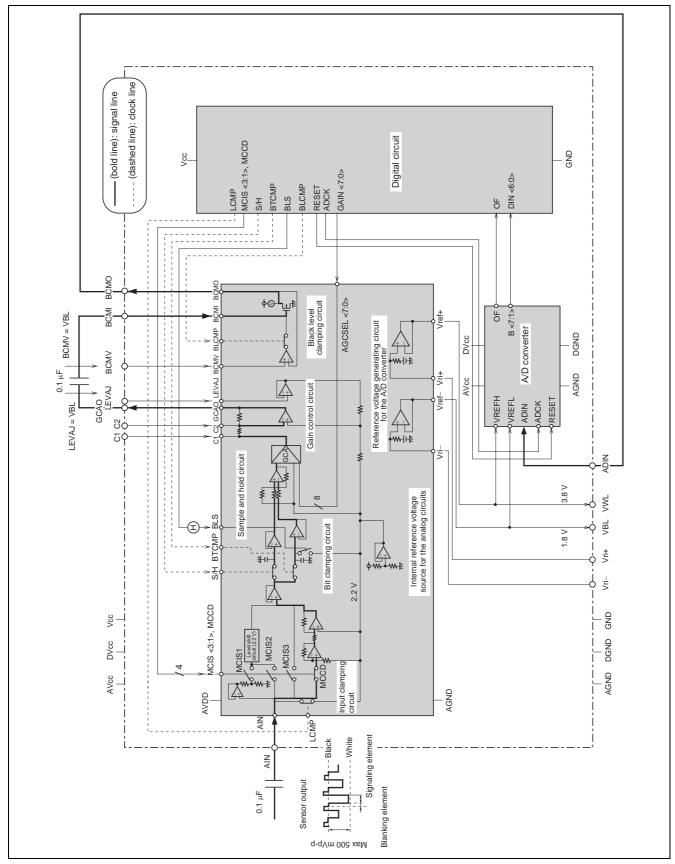


Figure 18 External Pin Connections of the Analog Part (for the CCD Mode/Bit Clamping)

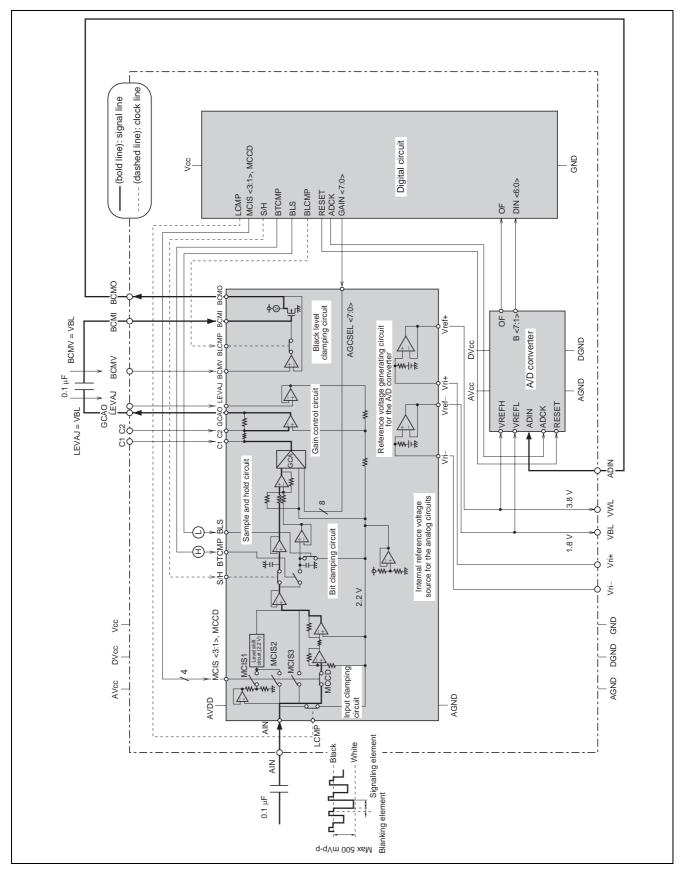


Figure 19 External Pin Connections of the Analog Part (for the CCD Mode/Line Clamping)

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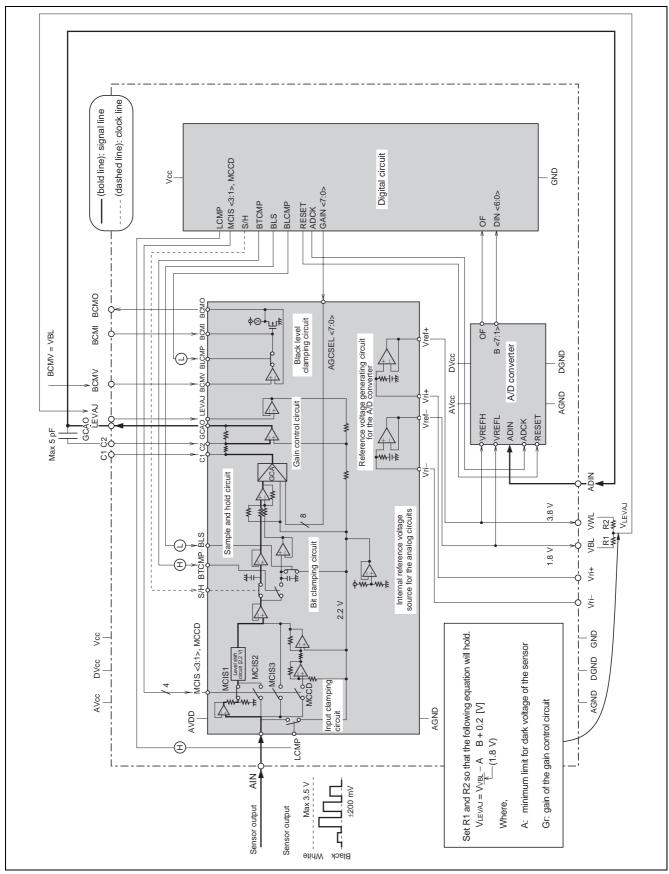


Figure 20 External Pin Connections of the Analog Part (for the CIS1 Mode)

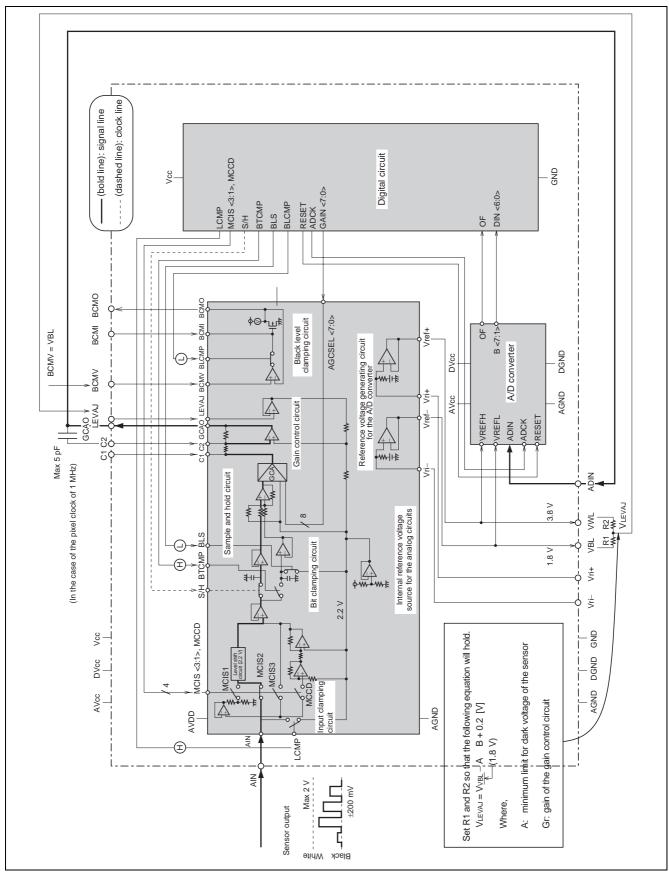


Figure 21 External Pin Connections of the Analog Part (for the CIS2 Mode)

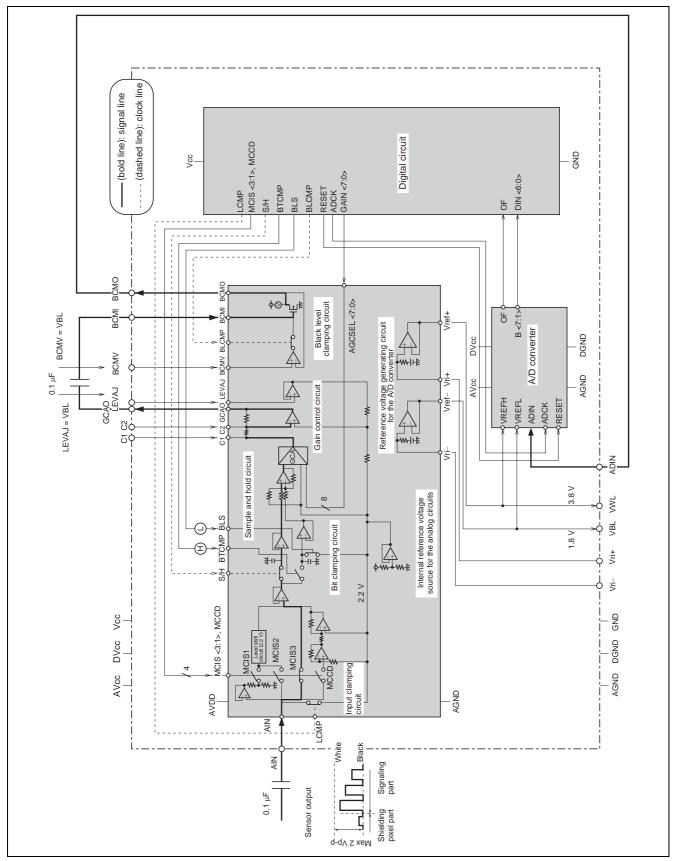
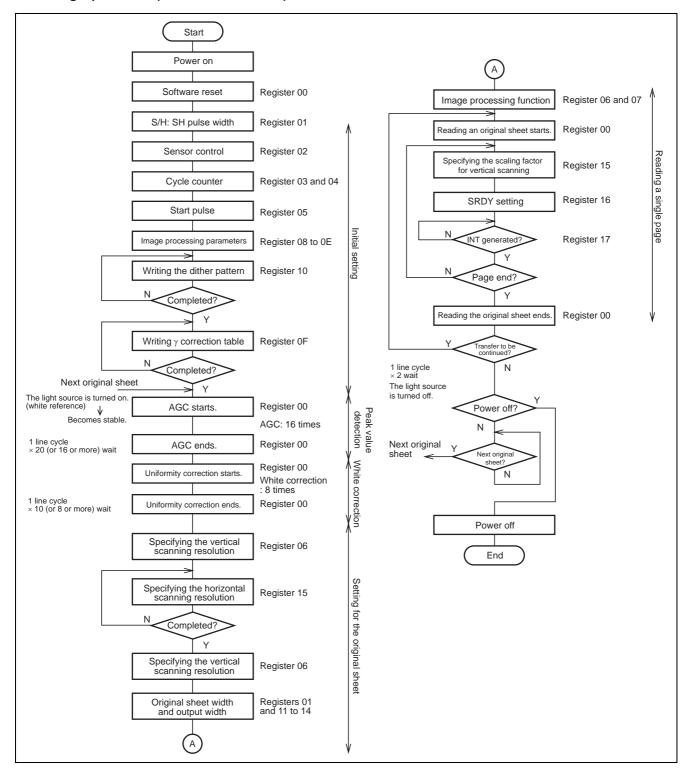


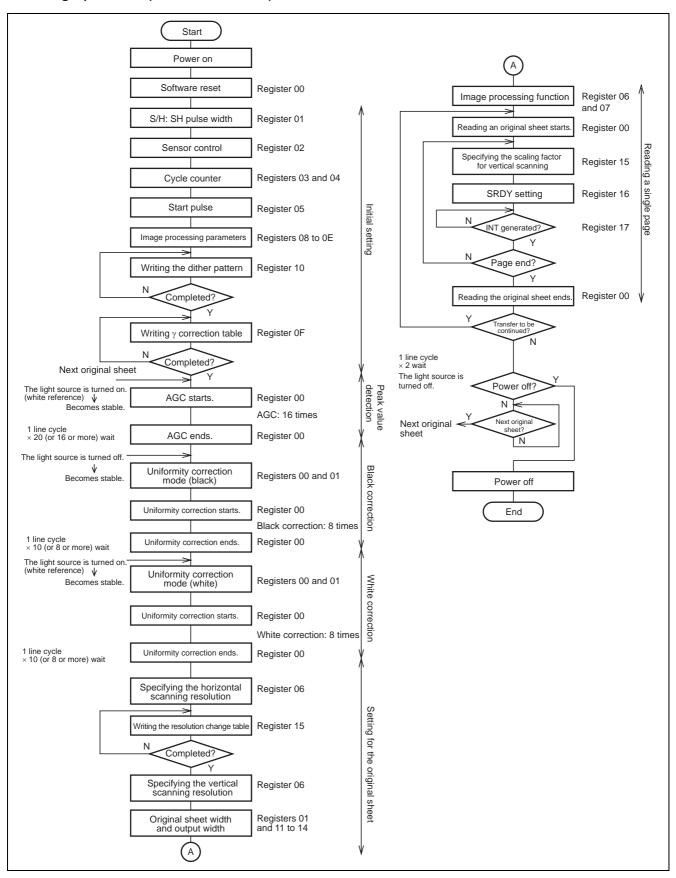
Figure 22 External Pin Connections of the Analog Part (for the CIS3 Mode)

Flowchart

Reading Operations (for the CCD Sensor)

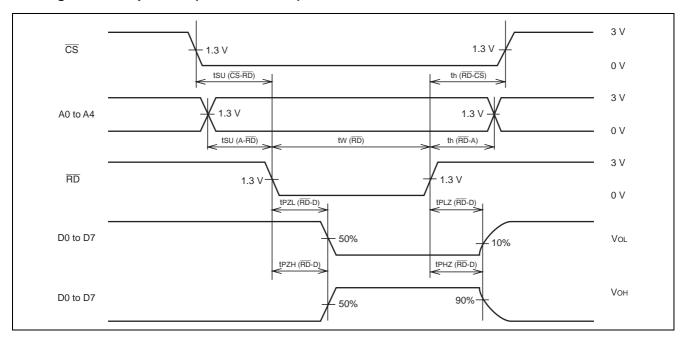


Reading Operations (for the CIS Sensor)

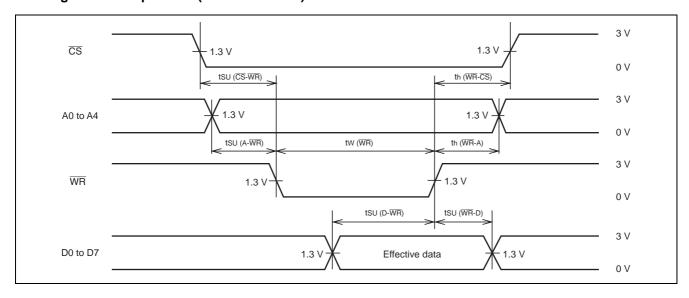


MPU Interface

Timing for Read Operation (M66335 → MPU)

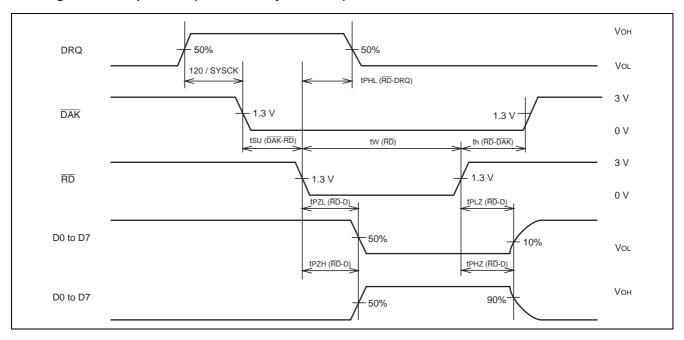


Timing for Write Operation (MPU \rightarrow M66335)

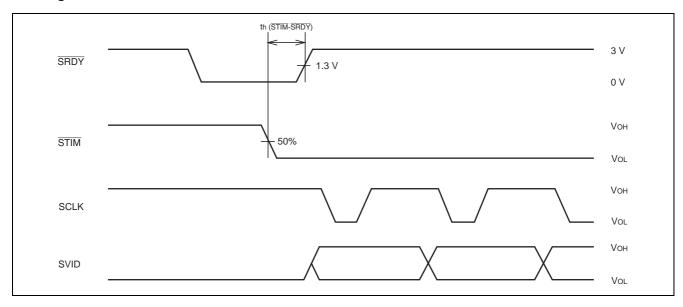


DMA Timing

Timing for Read Operation (M66335 → System Bus)



Timing of CODEC

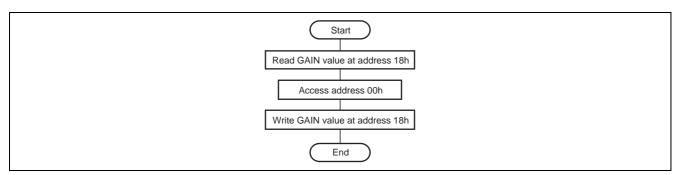


Cautions for Use

(1) Access to Address 00h

To gain access to address 00h, the value of built-in GCC (gain control counter) may be set to FFh.

This requires to read GAIN value at address 18h before access to address 00h and write the GAIN value at address 18h after the access (see flowchart A).



Flowchart A Address 00h Access Flow

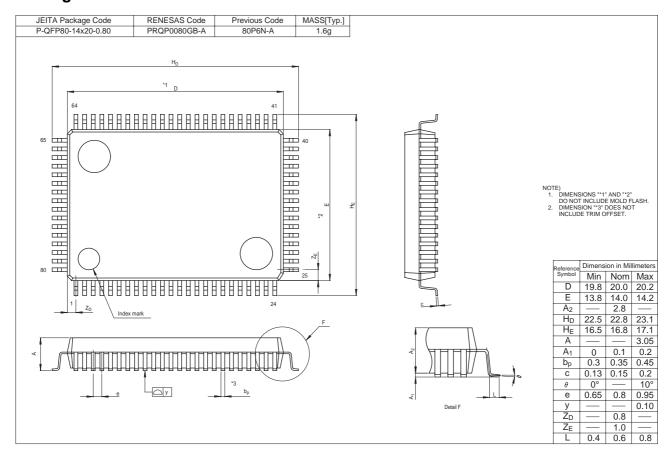
(2) Reset

The M66335FP adopts the two types of reset. These reset functions are provided in table A.

Table A Reset Functions

Function	Register	Internal F/F	GCC
Reset Type	Initialization	Initialization	Initialization
Hardware reset (RESET)	0	0	0
Software reset register 0 (RESET)		0	0

Package Dimensions



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