**DATA SHEET** 

### **General Description**

The 843051 is a Gigabit Ethernet Clock Generator. The 843051 can synthesize 10 Gigabit Ethernet, SONET, or Serial ATA reference clock frequencies with the appropriate choice of crystal and output divider. The 843051 has excellent phase jitter performance and is packaged in a small 8-pin TSSOP, making it ideal for use in systems with limited board space.

#### **Features**

- One differential 3.3V LVPECL output
- Crystal oscillator interface designed for 18pF parallel resonant crystals
- RMS phase jitter at:
   155.52MHz (12kHz 20MHz): 0.74ps (typical)
   156.25MHz (1.875MHz 20MHz): 0.43ps (typical)
   161.13MHz (1.933Hz 20MHz): 0.43ps (typical)

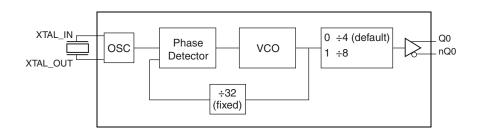
#### 

- Full 3.3V output supply mode
- 0°C to 70°C ambient operating temperature
- Available in lead-free (RoHS 6) package

#### **Frequency Table**

Inputs		
Crystal Frequency (MHz)	FREQ_SEL	Output Frequency Range (MHz)
20.141601	0	161.132812
20.141601	1	80.566406
19.53125	0	156.25
19.53125	1	78.125
19.44	0	155.52
19.44	1	77.76
18.75	0	150
18.75	1	75

# **Block Diagram**



# **Pin Assignment**





**Table 1. Pin Descriptions** 

Number	Name	T	ype	Description
1	V <sub>CCA</sub>	Power		Analog supply pin
2	V <sub>EE</sub>	Power		Negative supply pin.
3, 4	XTAL_OUT XTAL_IN	Input		Crystal oscillator interface. XTAL_IN is the input, XTAL_OUT is the output.
5	FREQ_SEL	Input	Pulldown	Frequency select pin. LVCMOS/LVTTL interface levels.
6, 7	nQ0, Q0	Output		Differential output pair. LVPECL interface levels.
8	V <sub>CC</sub>	Power		Core supply pin.

NOTE: Pulldown refers to internal input resistors. See Table 2, Pin Characteristics, for typical values.

### **Table 2. Pin Characteristics**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
C <sub>IN</sub>	Input Capacitance			4		pF
R <sub>PULLDOWN</sub>	Input Pulldown Resistor			51		kΩ

# **Absolute Maximum Ratings**

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, V <sub>CC</sub>	4.6V
Inputs, V <sub>I</sub>	-0.5V to V <sub>CC</sub> + 0.5V
Outputs, I <sub>O</sub> Continuous Current Surge Current	50mA 100mA
Package Thermal Impedance, $\theta_{JA}$	101.7°C/W (0 mps)
Storage Temperature, T <sub>STG</sub>	-65°C to 150°C

# **DC Electrical Characteristics**

Table 3A. Power Supply DC Characteristics,  $V_{CC} = 3.3V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = 0^{\circ}C$  to  $70^{\circ}C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>CC</sub>	Core Supply Voltage		3.135	3.3	3.465	V
V <sub>CCA</sub>	Analog Supply Voltage		3.135	3.3	3.465	V
I <sub>CC</sub>	Power Supply Current				70	mA
I <sub>CCA</sub>	Analog Supply Current				15	mA
I <sub>EE</sub>	Power Supply Current				85	mA



Table 3B. LVCMOS/LVTTL DC Characteristics,  $V_{CC} = 3.3V \pm 5\%, \ V_{EE} = 0V, \ T_A = 0^{\circ}C$  to  $70^{\circ}C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>IH</sub>	Input High Voltage		2		V <sub>CC</sub> + 0.3	V
V <sub>IL</sub>	Input Low Voltage		-0.3		0.8	V
I <sub>IH</sub>	Input High Current	V <sub>CC</sub> = V <sub>IN</sub> = 3.465V			150	μΑ
I <sub>IL</sub>	Input Low Current	V <sub>CC</sub> = 3.465V, V <sub>IN</sub> = 0V	-5			μΑ

### Table 3C. LVPECL DC Characteristics, $V_{CC}$ = 3.3V $\pm$ 5%, $V_{EE}$ = 0V, $T_A$ = 0°C to 70°C

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
V <sub>OH</sub>	Output High Voltage; NOTE 1		V <sub>CC</sub> – 1.4		V <sub>CC</sub> - 0.9	V
V <sub>OL</sub>	Output Low Voltage; NOTE 1		V <sub>CC</sub> - 2.0		V <sub>CC</sub> – 1.7	V
V <sub>SWING</sub>	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs termination with 50  $\!\Omega$  to  $V_{CC}$  – 2V.

#### **Table 4. Crystal Characteristics**

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation		Fundamental			
Frequency		12		40	MHz
Equivalent Series Resistance (ESR)				50	Ω
Shunt Capacitance				7	pF

### **AC Electrical Characteristics**

Table 5. AC Characteristics,  $V_{CC} = 3.3V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = 0^{\circ}C$  to  $70^{\circ}$ 

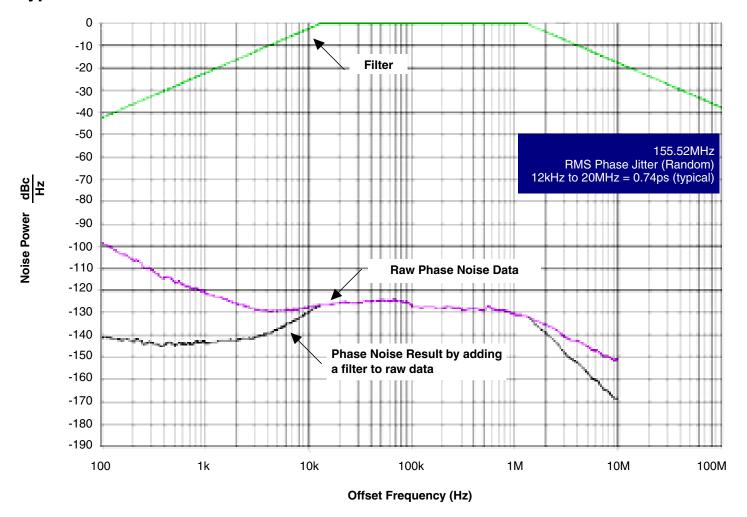
Symbol	Parameter		Minimum	Typical	Maximum	Units
				155.52		MHz
$f_{OUT}$	Output Frequency			156.25		MHz
				161.13		MHz
		155.52MHz, Integration Range: 12kHz – 20MHz		0.74		ps
		156.25MHz, Integration Range: 1.875MHz – 20MHz		0.43		ps
tjit(Ø)	RMS Phase Jitter, Random: NOTE 1	156.25MHz, Integration Range: 12kHz – 20MHz		0.75		ps
	riandom, rio i E i	161.13MHz, Integration Range: 1.933MHz – 20MHz		0.43		ps
		161.13MHz, Integration Range: 12kHz – 20MHz		0.72		ps
$t_R / t_F$	Output Rise/Fall Time	20% to 80%	325		600	ps
odc	Output Duty Cycle		49		51	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Please refer to Phase Noise Plot.

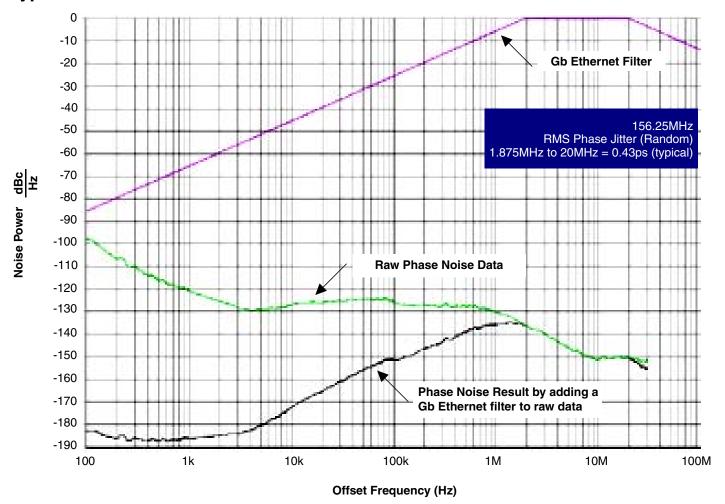


# Typical Phase Noise at 155.52MHz



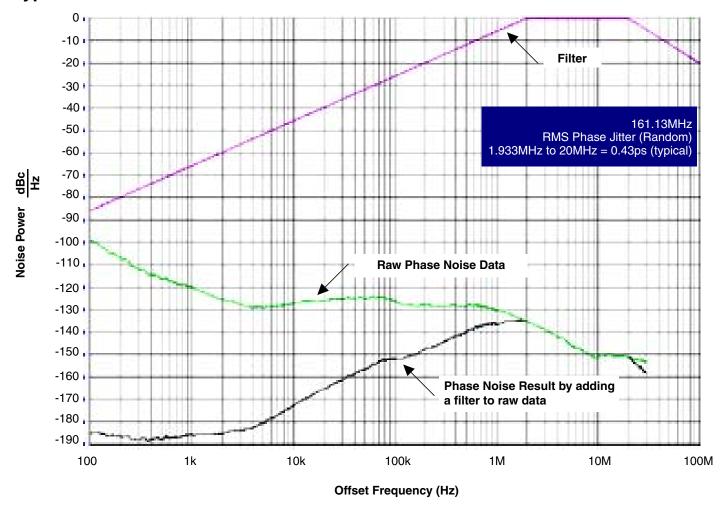


# Typical Phase Noise at 156.25MHz



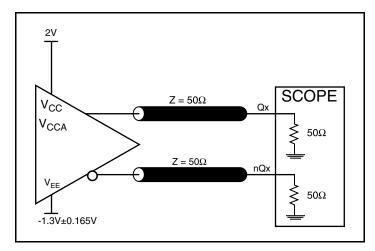


# **Typical Phase Noise at 161.13MHz**

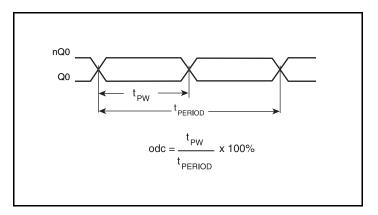




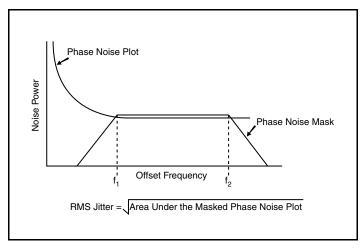
## **Parameter Measurement Information**



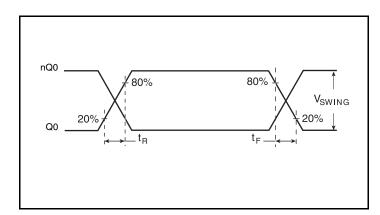
3.3V LVPECL Output Load AC Test Circuit



**Output Duty Cycle/Pulse Width/Period** 



**RMS Phase Jitter** 



**Output Rise/Fall Time** 



### **Applications Information**

### **Power Supply Filtering Technique**

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The 843051 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL.  $V_{CC}$  and  $V_{CCA}$  should be individually connected to the power supply plane through vias, and  $0.01\mu F$  bypass capacitors should be used for each pin. Figure 1 illustrates this for a generic  $V_{CC}$  pin and also shows that  $V_{CCA}$  requires that an additional  $10\Omega$  resistor along with a  $10\mu F$  bypass capacitor be connected to the  $V_{CCA}$  pin.

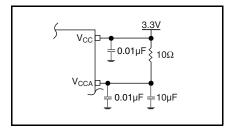


Figure 1. Power Supply Filtering

### **Crystal Input Interface**

The 843051 has been characterized with 18pF parallel resonant crystals. The capacitor values, C1 and C2, shown in *Figure 2* below were determined using a 26.04167MHz, 18pF parallel

X1 C1 33pF

18pF Parallel Crystal

C2 27pF

XTAL\_IN

XTAL\_OUT

Figure 2. Crystal Input Interface

resonant crystal and were chosen to minimize the ppm error. The optimum C1 and C2 values can be slightly adjusted for different board layouts.



### **Overdriving the XTAL Interface**

The XTAL\_IN input can be overdriven by an LVCMOS driver or by one side of a differential driver through an AC coupling capacitor. The XTAL\_OUT pin can be left floating. The amplitude of the input signal should be between 500mV and 1.8V and the slew rate should not be less than 0.2V/ns. For 3.3V LVCMOS inputs, the amplitude must be reduced from full swing to at least half the swing in order to prevent signal interference with the power rail and to reduce internal noise. *Figure 3A* shows an example of the interface diagram for a high speed 3.3V LVCMOS driver. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This

can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most  $50\Omega$  applications, R1 and R2 can be  $100\Omega$ . This can also be accomplished by removing R1 and changing R2 to  $50\Omega$ . The values of the resistors can be increased to reduce the loading for a slower and weaker LVCMOS driver. *Figure 3B* shows an example of the interface diagram for an LVPECL driver. This is a standard LVPECL termination with one side of the driver feeding the XTAL\_IN input. It is recommended that all components in the schematics be placed in the layout. Though some components might not be used, they can be utilized for debugging purposes. The datasheet specifications are characterized and quaranteed by using a quartz crystal as the input.

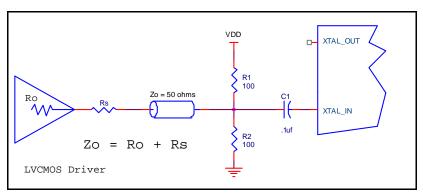


Figure 3A. General Diagram for LVCMOS Driver to XTAL Input Interface

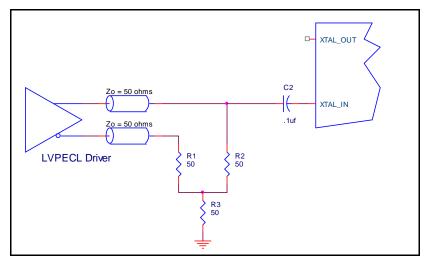


Figure 3B. General Diagram for LVPECL Driver to XTAL Input Interface



### **Termination for 3.3V LVPECL Outputs**

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive  $50\Omega$ 

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

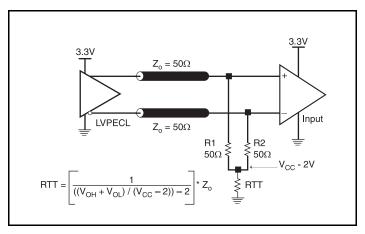


Figure 4A. 3.3V LVPECL Output Termination

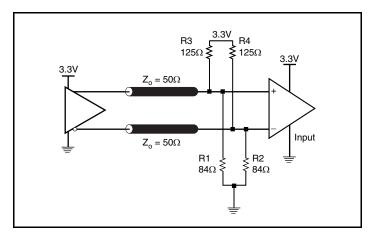


Figure 4B. 3.3V LVPECL Output Termination



### **Schematic Example**

Figure 5A shows a schematic example of the 843051. An example of LVEPCL termination is shown in this schematic. Additional LVPECL termination approaches are shown in the LVPECL Termination Application Note. In this example, an 18pF parallel resonant crystal

is used. The C1 = 27pF and C2 = 33pF are recommended for frequency accuracy. The C1 and C2 values may be slightly adjusted for optimizing frequency accuracy.

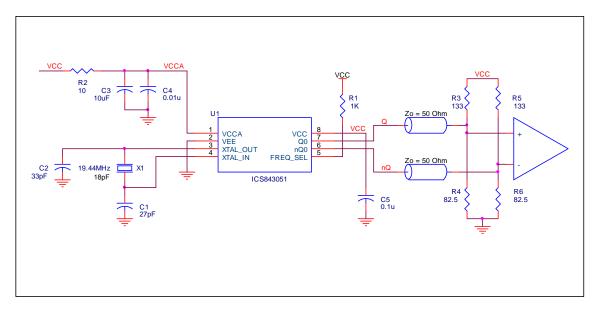
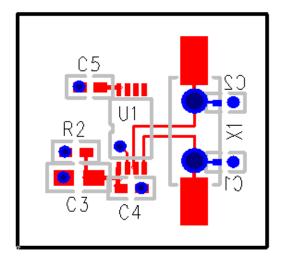


Figure 5A. 843051 Schematic Example

### **PC Board Layout Example**

Figure 5B shows an example of 843051 P.C. board layout. The crystal X1 footprint shown in this example allows installation of either surface mount HC49S or through-hole HC49 package. The footprints of other components in this example are listed in the *Table 6*. There should be

Figure 5B. 843051 PC Board Layout Example



at least one decoupling capacitor per power pin. The decoupling capacitors should be located as close as possible to the power pins. The layout assumes that the board has clean analog power ground plane.

**Table 6. Footprint Table** 

Reference	Size
C1, C2	0402
C3	0805
C4, C5	0603
R2	0603

NOTE: Table 6 lists component sizes shown in this layout example.



### **Power Considerations**

This section provides information on power dissipation and junction temperature for the 843051. Equations and example calculations are also provided.

#### 1. Power Dissipation.

The total power dissipation for the 843051 is the sum of the core power plus the power dissipated in the load(s).

The following is the power dissipation for  $V_{CC} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> = V<sub>CC MAX</sub> \* I<sub>EE MAX</sub> = 3.465V \* 85mA = 294.5mW
- Power (outputs)<sub>MAX</sub> = 30mW/Loaded Output pair

Total Power\_MAX (3.465V, with all outputs switching) = 294.5mW + 30mW = 324.5mW

#### 2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA}$  \* Pd\_total + T<sub>A</sub>

Tj = Junction Temperature

 $\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

T<sub>A</sub> = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming a moderate air flow of 1 meter per second and a multi-layer board, the appropriate value is 90.5°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is:

 $70^{\circ}\text{C} + 0.324\text{W} * 90.5^{\circ}\text{C/W} = 99.3^{\circ}\text{C}$ . This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resitance  $\theta_{JA}$  for 8 Lead TSSOP, Forced Convection

$\theta_{JA}$ vs. Air Flow				
Meters per Second	0	1	2.5	
Multi-Layer PCB, JEDEC Standard Test Boards	101.7°C/W	90.5°C/W	89.8°C/W	



#### 3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pair.

LVPECL output driver circuit and termination are shown in Figure 6.

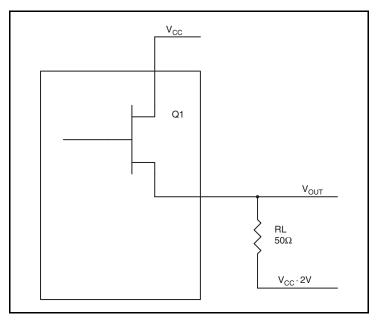


Figure 6. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a  $50\Omega$  load, and a termination voltage of  $V_{CC}$  – 2V.

- For logic high,  $V_{OUT} = V_{OH\_MAX} = V_{CC\_MAX} 0.9V$   $(V_{CC\_MAX} V_{OH\_MAX}) = 0.9V$
- For logic low,  $V_{OUT} = V_{OL\_MAX} = V_{CC\_MAX} 1.7V$  $(V_{CC\_MAX} - V_{OL\_MAX}) = 1.7V$

Pd\_H is power dissipation when the output drives high.

Pd\_L is the power dissipation when the output drives low.

$$Pd\_H = [(V_{OH\_MAX} - (V_{CC\_MAX} - 2V))/R_L] * (V_{CC\_MAX} - V_{OH\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OH\_MAX}))/R_L] * (V_{CC\_MAX} - V_{OH\_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = \textbf{19.8mW}$$

$$Pd\_L = [(V_{OL\_MAX} - (V_{CC\_MAX} - 2V))/R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OL\_MAX}))/R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = \textbf{10.2mW}$$

Total Power Dissipation per output pair = Pd\_H + Pd\_L = 30mW



# **Reliability Information**

Table 8.  $\theta_{\mbox{\scriptsize JA}}$  vs. Air Flow Table for a 8 Lead TSSOP

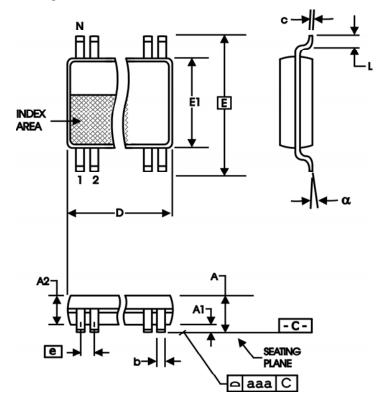
$\theta_{\sf JA}$ vs. Air Flow				
Meters per Second	0	1	2.5	
Multi-Layer PCB, JEDEC Standard Test Boards	101.7°C/W	90.5°C/W	89.8°C/W	

### **Transistor Count**

The transistor count for 843051 is: 1892

# **Package Outline and Package Dimensions**

Package Outline - G Suffix for 8 Lead TSSOP



**Table 9. Package Dimensions** 

All Dimensions in Millimeters						
Symbol	Minimum	Maximum				
N	8					
Α		1.20				
<b>A</b> 1	0.5	0.15				
A2	0.80	1.05				
b	0.19	0.30				
С	0.09	0.20				
D	2.90	3.10				
Е	6.40 Basic					
E1	4.30	4.50				
е	0.65 Basic					
L	0.45	0.75				
α	0°	8°				
aaa		0.10				

Reference Document: JEDEC Publication 95, MO-153



# **Ordering Information**

## **Table 10. Ordering Information**

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
843051AGLF	051AL	"Lead-Free" 8 Lead TSSOP	Tube	0°C to 70°C
843051AGLFT	051AL	"Lead-Free" 8 Lead TSSOP	Tape & Reel	0°C to 70°C



# **Revision History Sheet**

Rev	Table	Page	Description of Change	Date
Α	T10	14	Ordering Information Table - corrected count from 154 per tube to 100 per tube.	
Α	T10	1 14	Added Lead-Free bullet in Features section. Ordering Information Table - added "Lead-Free" part.	
Α	T10	14	Ordering Information Table/Shipping Packaging column - deleted tube quantity.	
Α		9	Added LVCMOS to XTAL Interface section.	3/5/08
Α	T3C T5	3 3 9	LVPECL DC Characteristics Table - corrected V <sub>OH</sub> /V <sub>OL</sub> parameters from "Current" to "Voltage" and units from "uA" to "V".  AC Characteristics Table - added thermal note.  Updated "Overdriving the Crystal Interface" section.  Updated header/footer.	10/13/10
В	T10	1 9 15	Updated header/footer through the datasheet.  Deleted IDT prefix from part number.  Features Section - deleted leaded information in the last bullet.  Added outline box around Block Diagram.  Updated Overdriving the XTAL Interface application note.  Ordering Information table - deleted leaded parts rows and note.	10/15/15



#### **IMPORTANT NOTICE AND DISCLAIMER**

RENESAS ELECTRONICS CORPORATION AND ITS SUBSIDIARIES ("RENESAS") PROVIDES TECHNICAL SPECIFICATIONS AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS OR IMPLIED, INCLUDING, WITHOUT LIMITATION, ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, OR NON-INFRINGEMENT OF THIRD-PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for developers who are designing with Renesas products. You are solely responsible for (1) selecting the appropriate products for your application, (2) designing, validating, and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. Renesas grants you permission to use these resources only to develop an application that uses Renesas products. Other reproduction or use of these resources is strictly prohibited. No license is granted to any other Renesas intellectual property or to any third-party intellectual property. Renesas disclaims responsibility for, and you will fully indemnify Renesas and its representatives against, any claims, damages, costs, losses, or liabilities arising from your use of these resources. Renesas' products are provided only subject to Renesas' Terms and Conditions of Sale or other applicable terms agreed to in writing. No use of any Renesas resources expands or otherwise alters any applicable warranties or warranty disclaimers for these products.

(Disclaimer Rev.1.01)

#### **Corporate Headquarters**

TOYOSU FORESIA, 3-2-24 Toyosu, Koto-ku, Tokyo 135-0061, Japan www.renesas.com

#### **Trademarks**

Renesas and the Renesas logo are trademarks of Renesas Electronics Corporation. All trademarks and registered trademarks are the property of their respective owners.

#### **Contact Information**

For further information on a product, technology, the most up-to-date version of a document, or your nearest sales office, please visit <a href="https://www.renesas.com/contact-us/">www.renesas.com/contact-us/</a>.