General Description

The 843252-45 is a 2 LVPECL output Synthesizer optimized to generate Ethernet reference clock frequencies. Using a 25MHz, 18pF parallel resonant crystal, the following frequencies can be generated: 156.25MHz and 125MHz. The 843252-45 uses IDT's 3rd generation low phase noise VCO technology and can achieve 1ps or lower typical rms phase jitter, easily meeting Ethernet jitter requirements. The 843252-45 is packaged in a small 16-pin TSSOP package.

Features

- **•** Two differential LVPECL output pairs
- **•** Crystal oscillator interface designed for a 25MHz, 18pF parallel resonant crystal
- **•** A 25MHz crystal generates output frequencies of: 156.25MHz and 125MHz
- **•** VCO frequency: 625MHz
- **•** RMS Phase Jitter @ 156.25MHz, (1.875MHz 20MHz) using a 25MHz crystal: 0.54ps (typical)
- **•** Full 3.3V supply mode
- **•** 0°C to 70°C ambient operating temperature
- **•** Available in lead-free (RoHS 6) package
- **•** For functional replacement part us 8T49N241

Block Diagram

Pin Assignment

16-Lead TSSOP 4.4mm x 5.0mm x 0.925mm package body G Package

Table 1. Pin Descriptions

NOTE: *Pullup* refers to internal input resistors. See Table 2, *Pin Characteristics,* for typical values.

Table 2. Pin Characteristics

Function Tables

Table 3A. CLK_EN_A Function Table **Table 3B. CLK_EN_B Function Table**

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

DC Electrical Characteristics

Table 4A. Power Supply DC Characteristics, $V_{CC} = V_{CCOA} = V_{CCOB} = 3.3V \pm 5\%$, $T_A = 0^{\circ}C$ to 70°C

Table 4B. LVCMOS/LVTTL DC Characteristics, $V_{CC} = V_{CCOA} = V_{CCOB} = 3.3V \pm 5\%$, $T_A = 0^{\circ}C$ to 70^oC

Table 4C. LVPECL DC Characteristics, $V_{CC} = V_{CCOA} = V_{CCOB} = 3.3V \pm 5\%$, $T_A = 0^{\circ}C$ to 70^oC

NOTE 1: Output termination with 50 Ω to $V_{CCOAB} - 2V$.

Table 5. Crystal Characteristics

NOTE: Characterized using an 18pF parallel resonant crystal.

AC Electrical Characteristics

Table 6. AC Characteristics, $V_{CC} = V_{CCOA} = V_{CCOB} = 3.3V \pm 5\%$, $T_A = 0^{\circ}C$ to 70°C

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. Device will meet specifications after thermal equilibrium has been reached under these conditions.

Using a 25MHz, 18pF quartz crystal.

NOTE 1: Please refer to the Phase Noise plots.

Typical Phase Noise at 125MHz

Typical Phase Noise at 156.25MHz

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Parameter Measurement Information

3.3V LVPECL Output Load AC Test Circuit

Output Rise/Fall Time

RMS Phase Jitter

Output Duty Cycle/Pulse Width/Period

Application Information

Recommendations for Unused Input Pins

Inputs:

LVCMOS Control Pins

All control pins have internal pullups; additional resistance is not required but can be added for additional protection. A 1 $k\Omega$ resistor can be used.

Power Supply Filtering Technique

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The 843252-45 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. V_{CC} , V_{CCAA} , V_{CCOA} , and V_{CCOB} should be individually connected to the power supply plane through vias, and 0.01µF bypass capacitors should be used for each pin. *Figure 1* illustrates this for a generic V_{CC} pin and also shows that V_{CCA} requires that an additional 10 Ω resistor along with a 10 μ F bypass capacitor be connected to the V_{CCA} pin.

Outputs:

LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

Figure 1. Power Supply Filtering

Crystal Input Interface

The 843252-45 has been characterized with 18pF parallel resonant crystals. The capacitor values, C1 and C2, shown in *Figure 2* below were determined using a 25MHz, 18pF parallel

Figure 2. Crystal Input Interface

resonant crystal and were chosen to minimize the ppm error. The optimum C1 and C2 values can be slightly adjusted for different board layouts.

LVCMOS to XTAL Interface

The XTAL_IN input can accept a single-ended LVCMOS signal through an AC coupling capacitor. A general interface diagram is shown in *Figure 3.* The XTAL_OUT pin can be left floating. The input edge rate can be as slow as 10ns. For LVCMOS inputs, it is recommended that the amplitude be reduced from full swing to half swing in order to prevent signal interference with the power rail and to reduce noise. This configuration requires that the output

impedance of the driver (Ro) plus the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First, R1 and R2 in parallel should equal the transmission line impedance. For most 50Ω applications, R1 and R2 can be 100 Ω . This can also be accomplished by removing R1 and making R2 50 Ω .

Figure 3. General Diagram for LVCMOS Driver to XTAL Input Interface

Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

 $3.3V$ $3.3V$ $Z_0 = 50\Omega$ Input LVPECL $Z_0 = 50\Omega$ $R1$ $B₂$ ≲ 50_S 50Ω $V_{CC} - 2V$ **RTT RTT**

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

Figure 4A. 3.3V LVPECL Output Termination Figure 4B. 3.3V LVPECL Output Termination

Schematic Example

Figure 5 shows an example of 843252-45 application schematic. In this example, the device is operated at $V_{CC} = 3.3V$. The 18pF parallel resonant 25MHz crystal is used. The C1 = 27pF and C2 = 27pF are recommended for frequency accuracy. For different board layouts, the C1 and C2 may be slightly adjusted for

optimizing frequency accuracy. Two examples of LVPECL terminations are shown in this schematic. Additional termination approaches are shown in the *LVPECL Termination Application Note.*

Power Considerations

This section provides information on power dissipation and junction temperature for the 843252-45. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the 843252-45 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core) $_{MAX}$ = V_{CC} $_{MAX}$ $*$ I_{EE} $_{MAX}$ = 3.465V $*$ 75mA = 259.9mW
- Power (outputs)_{MAX} = 30mW/Loaded Output pair If all outputs are loaded, the total power is 2 * 30mW = **60mW**

Total Power_{$-MAX$} (3.3V, with all outputs switching) = 259.9mW + 60mW = 319.9mW

2. Junction Temperature.

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS devices is 125°C.

The equation for Tj is as follows: Tj = θ_{JA} * Pd_total + T_A

Tj = Junction Temperature

 θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_total = Total Device Power Dissipation (example calculation is in section 1 above)

 T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 92.4°C/W per Table 7 below.

Therefore, Tj for an ambient temperature of 70°C with all outputs switching is:

 70° C + 0.320W * 92.4°C/W = 99.6°C. This is well below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 7. Thermal Resistance θ_{JA} for 16 Lead TSSOP, Forced Convection

3. Calculations and Equations.

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The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in *Figure 6.*

Figure 6. LVPECL Driver Circuit and Termination

To calculate worst case power dissipation into the load, use the following equations which assume a 50 Ω load, and a termination voltage of $V_{CCO} - 2V$.

- \bullet For logic high, V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} 0.9V $(V_{\text{CCO_MAX}} - V_{\text{OH_MAX}}) = 0.9V$
- \bullet For logic low, V_{OUT} = $\text{V}_{\text{OL_MAX}}$ = $\text{V}_{\text{CCO_MAX}}$ 1.7V $(V_{\text{CCO_MAX}} - V_{\text{OL_MAX}}) = 1.7V$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

Pd_H = [(V_{OH_MAX} – (V_{CCO_MAX} – 2V))/R_L] * (V_{CCO_MAX} – V_{OH_MAX}) = [(2V – (V_{CCO_MAX} – V_{OH_MAX}))/R_L] * (V_{CCO_MAX} – V_{OH_MAX}) = $[(2V – 0.9V)/50 Ω] * 0.9V = 19.8mW$

Pd_L = [(V_{OL_MAX} – (V_{CCO_MAX} – 2V))/R_L] * (V_{CCO_MAX} – V_{OL_MAX}) = [(2V – (V_{CCO_MAX} – V_{OL_MAX}))/R_{L]} * (V_{CCO_MAX} – V_{OL_MAX}) = $[(2V – 1.7V)/50^Ω] * 1.7V = 10.2mW$

Total Power Dissipation per output pair = Pd_H + Pd_L = **30mW**

Reliability Information

Table 8. θ_{JA} vs. Air Flow Table for a 16 Lead TSSOP

Transistor Count

The transistor count for 843252-45 is: 2039

Package Outline and Package Dimensions

Package Outline - G Suffix for 16-Lead TSSOP Table 9. Package Dimensions for 16 Lead TSSOP

Reference Document: JEDEC Publication 95, MO-153

Ordering Information

Table 10. Ordering Information

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

Revision History Sheet

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