# **DATASHEET**

# **General Description**

The 8T349316 is a 2.5V differential clock buffer with sixteen LVDS outputs. The fanout from a differential input to the sixteen LVDS outputs reduces loading on the preceding driver and provides an efficient clock distribution network. The 8T349316 can act as a translator from a differential HSTL, LVPECL, CML or LVDS input to LVDS output signals. A single-ended 3.3V, 2.5V LVCMOS/LVTTL input can also be used to translate to LVDS outputs. The redundant input capability allows for an asynchronous change-over from a primary clock source to a secondary clock source. Selectable reference inputs are controlled by SEL. The 8T349316 outputs can be asynchronously enabled/disabled. When disabled, the outputs will drive to the value selected by the GL pin. Multiple power and grounds reduce noise.The extended temperature range supports wireless infrastructure, telecommunication and networking end equipment requirements. The device is a member of the high-performance clock family from IDT.

# **Block Diagram**



### **Features**

- **•** Clock signal selection and fanout to 16 LVDS outputs
- **•** Guaranteed Low Skew < 50ps (max)
- **•** Low output pulse skew < 125ps (max)
- **•** Propagation delay < 1.75ns (max)
- **•** Up to 1GHz clock signal operation
- **•** Support the following input types: HSTL, LVPECL, HCSL, LVTTL
- **•** Selectable differential input
- **•** Power-down mode
- **•** Full 2.5V power supply
- **•** -40°C to +85°C ambient operating temperature
- **•** Lead-free (RoHS 6) 52-lead VFQFN-P packaging
- **•** Replacement device for the 5T9316

# **Pin Assignment**



# **Pin Description and Pin Characteristic Tables**

### **Table 1: Pin Descriptions**

![](_page_1_Picture_362.jpeg)

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### **Table 1: Pin Descriptions**

![](_page_2_Picture_221.jpeg)

#### **Table 2. Pin Characteristics**

![](_page_2_Picture_222.jpeg)

# **Logic Truth Tables**

### <span id="page-3-3"></span>**Table 3A: Input Signal Source Select1**

![](_page_3_Picture_121.jpeg)

1. Asynchronous control.

#### <span id="page-3-2"></span>**Table 3B. Device Power-down control1**

![](_page_3_Picture_122.jpeg)

1. Asynchronous control.

2. Disable outputs by setting  $nG1 = nG2 = 1$  before entering power-down mode and while in power-down mode. To enter normal device operation, first enable the outputs by setting  $nG1 = nG2 = 0$  before setting  $nPD = 1$ .

#### <span id="page-3-0"></span>**Table 3C. Output Q[1:8] Enable Control1**

![](_page_3_Picture_123.jpeg)

1. Asynchronous controls.

### <span id="page-3-1"></span>**Table 3D. Output Q[9:16] Enable Control1**

![](_page_3_Picture_124.jpeg)

1. Asynchronous controls.

# **Absolute Maximum Ratings**

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of the product at these conditions or any conditions beyond those listed in the *DC Characteristics or AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

#### **Table 4. Absolute Maximum Ratings**

![](_page_4_Picture_113.jpeg)

#### **Table 5. Recommended Operating Range**

![](_page_4_Picture_114.jpeg)

### **DC Electrical Characteristics**

![](_page_5_Picture_256.jpeg)

### **Table 6A. Power Supply DC Characteristics,**  $V_{DD} = 2.5V \pm 10\%$ **,**  $T_A = -40\degree C$  **to**  $+85\degree C$

### **Table 6B. LVCMOS/LVTTL DC Characteristics,**  $V_{DD} = 2.5V \pm 10\%$ **,**  $T_A = -40\degree C$  **to**  $+85\degree C$

![](_page_5_Picture_257.jpeg)

# **Table 6C. Differential Input DC Characteristics,**  $V_{DD} = 2.5V \pm 10\%$ ,  $T_A = -40\degree C$  to  $+85\degree C$  <sup>1</sup>

![](_page_5_Picture_258.jpeg)

1.  $V_{IL}$  should not be less than -0.3V.

2. Common mode input voltage is defined at the crosspoint.

# **Table 6D. LVDS DC Characteristics,**  $V_{DD} = 2.5V \pm 10\%$ ,  $T_A = -40\degree C$  to  $+85\degree C$

![](_page_5_Picture_259.jpeg)

# **AC Electrical Characteristics**

![](_page_6_Picture_176.jpeg)

# **Table 7. AC Characteristics,**  $V_{DD} = 2.5V \pm 10\%$ **,**  $T_A = -40\degree C$  **to**  $+85\degree C^1$

1. Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

2. This parameter is defined in accordance with JEDEC standard 65.

3. Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the differential cross point.

4. Defined as skew between outputs on different devices operating at the same supply voltage, same temperature, same frequency and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross point.

# **Parameter Measurement Information**

![](_page_7_Figure_2.jpeg)

![](_page_7_Figure_3.jpeg)

![](_page_7_Figure_4.jpeg)

**Propagation Delay**

![](_page_7_Figure_6.jpeg)

**Part-to-Part Skew**

![](_page_7_Figure_8.jpeg)

**Offset Voltage Setup**

![](_page_7_Figure_10.jpeg)

**Differential Input Level** 

![](_page_7_Figure_12.jpeg)

![](_page_7_Figure_13.jpeg)

![](_page_7_Figure_14.jpeg)

**Pulse Skew**

![](_page_7_Figure_16.jpeg)

**Differential Output Voltage Setup**

# **Applications Information**

### **Recommendations for Unused Input and Output Pins**

#### **Inputs:**

#### **Ax/nAx Clock Inputs**

For applications not requiring the use of the differential input, Ax should be pulled up with a  $10k\Omega$  resistor and nAx pulled down with a 10k $\Omega$  resistor.

#### **LVCMOS Control Pins**

All control pins have internal pullup or pulldown resistors; additional resistance is not required but can be added for additional protection. A 1 $k\Omega$  resistor can be used.

### **Outputs:**

### **LVDS Outputs**

All unused LVDS output pairs can be either left floating or terminated with 100 $\Omega$  across. If they are left floating, there should be no trace attached.

# **Wiring the Differential Input to Accept Single-Ended Levels**

*[Figure 1](#page-9-0)* shows how a differential input can be wired to accept single ended levels. The reference voltage  $V_1= V_{DD}/2$  is generated by the bias resistors R1 and R2. The bypass capacitor (C1) is used to help filter noise on the DC bias. This bias circuit should be located as close to the input pin as possible. The ratio of R1 and R2 might need to be adjusted to position the  $V_1$ in the center of the input voltage swing. For example, if the input clock swing is 2.5V and  $V_{DD} = 3.3V$ , R1 and R2 value should be adjusted to set  $V_1$  at 1.25V. The values below are for when both the single ended swing and  $V_{DD}$  are at the same voltage. This configuration requires that the sum of the output impedance of the driver (Ro) and the series resistance (Rs) equals the transmission line impedance. In addition, matched termination at the input will attenuate the signal in half. This can be done in one of two ways. First, R3 and R4 in parallel should equal the transmission line

impedance. For most 50 $\Omega$  applications, R3 and R4 can be 100 $\Omega$ . The values of the resistors can be increased to reduce the loading for slower and weaker LVCMOS driver. When using single-ended signaling, the noise rejection benefits of differential signaling are reduced. Even though the differential input can handle full rail LVCMOS signaling, it is recommended that the amplitude be reduced. The datasheet specifies a lower differential amplitude, however this only applies to differential signals. For single-ended applications, the swing can be larger, however  $V_{IL}$  cannot be less than -0.3V and  $V_{\text{IH}}$  cannot be more than  $V_{\text{DD}}$  + 0.3V. Though some of the recommended components might not be used, the pads should be placed in the layout. They can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a differential signal.

![](_page_9_Figure_4.jpeg)

<span id="page-9-0"></span>**Figure 1. Recommended Schematic for Wiring a Differential Input to Accept Single-ended Levels.**

# **LVDS Driver Termination**

For a general LVDS interface, the recommended value for the termination impedance  $(Z_T)$  is between 90 $\Omega$  and 132 $\Omega$ . The actual value should be selected to match the differential impedance  $(Z_0)$  of your transmission line. A typical point-to-point LVDS design uses a  $100\Omega$  parallel resistor at the receiver and a 100 $\Omega$  differential transmission-line environment. In order to avoid any transmission-line reflection issues, the components should be surface mounted and must be placed as close to the receiver as possible. IDT offers a full line of LVDS compliant devices with two types of output structures: current source and voltage source. The standard termination schematic as shown in *[Figure 2A](#page-10-0)* can be used with either type of output structure. *[Figure 2B](#page-10-1)*, which can also be used with both output types, is an optional termination with center tap capacitance to help filter common mode noise. The capacitor value should be approximately 50pF. If using a non-standard termination, it is recommended to contact IDT and confirm if the output structure is current source or voltage source type. In addition, since these outputs are LVDS compatible, the input receiver's amplitude and common-mode input range should be verified for compatibility with the output.

![](_page_10_Figure_4.jpeg)

<span id="page-10-0"></span>**Figure 2A. Standard LVDS Termination**

![](_page_10_Figure_6.jpeg)

<span id="page-10-1"></span>**Figure 2B. Optional LVDS Termination**

# **VFQFN EPAD Thermal Release Path**

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *[Figure 3.](#page-11-0)* The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as "heat pipes". The number of vias (i.e. "heat pipes") are application specific

and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor's Thermally/ Electrically Enhance Leadframe Base Package, Amkor Technology.

![](_page_11_Figure_5.jpeg)

<span id="page-11-0"></span>**Figure 3. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale).**

![](_page_12_Picture_0.jpeg)

# **Power Considerations**

This section provides information on power dissipation and junction temperature for the 8T349316. Equations and example calculations are also provided.

#### **1. Power Dissipation.**

The total power dissipation for the 8T349316 is the sum of the core power plus the output power dissipation due to the load. The following is the power dissipation for  $V_{DD} = 2.5V + 10\% = 2.700V$ , which gives worst case results.

The maximum current at 85°C is as follows:

 $I_{DD$  MAX = 360mA

• Power  $_{(core)MAX}$  =  $V_{DD\_MAX}$  \*  $I_{DD\_MAX}$  = 2.7V \* 360mA = **972mW** Total Power <sub>MAX</sub> = 972mW

#### **2. Junction Temperature.**

Junction temperature, Tj, is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, Tj, to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for Tj is as follows: Tj =  $\theta_{JA}$  \* Pd\_total + TA

Tj = Junction Temperature

 $\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

Pd\_total = Total Device Power Dissipation (example calculation is in section 1 above)

 $T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 33°C/W per Table 6 below.

Therefore, Tj for an ambient temperature of 85°C with all outputs switching is:

 $85^{\circ}$ C + 0.972W  $*$  33°C/W = 117.1°C. This is below the limit of 125°C.

This calculation is only an example. Tj will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

#### Table 6. Thermal Resistance  $\theta_{JA}$  for 52-Lead VFQFN, Forced Convection

![](_page_12_Picture_154.jpeg)

# **Reliability Information**

### Table 8.  $\theta_{JA}$  vs. Air Flow Table for a 52-lead VFQFN Package

![](_page_13_Picture_47.jpeg)

### **Transistor Count**

The transistor count for 8T349316 is: 1821

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# **52-Lead VFQFN Package Outline**

![](_page_14_Figure_2.jpeg)

# **52-Lead VFQFN Package Outline, (continued)**

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![](_page_15_Picture_20.jpeg)

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![](_page_16_Figure_1.jpeg)

# **52-Lead VFQFN Package Outline, (continued)**

# **Ordering Information**

### **Table 9. Ordering Information**

![](_page_17_Picture_30.jpeg)

![](_page_18_Picture_0.jpeg)

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