

Leading performance 120-MHz Arm® Cortex®-M4 core, up to 512-KB of code flash memory, 64-KB SRAM, security and safety features, and advanced analog.

Features

■ Arm Cortex-M4 Core with Floating Point Unit (FPU)

- Armv7E-M architecture with DSP instruction set
- Maximum operating frequency: 120 MHz
- Support for 4-GB address space
- On-chip debugging system: JTAG, SWD, and ETM
- Boundary scan and Arm Memory Protection Unit (Arm MPU)

■ Memory

- Up to 512-KB code flash memory (40 MHz zero wait states)
- 8-KB data flash memory (125,000 erase/write cycles)
- 64-KB SRAM
- Flash Cache (FCACHE)
- Memory Protection Units (MPU)
- Memory Mirror Function (MMF)
- 128-bit unique ID

■ Connectivity

- Serial Communications Interface (SCI) with FIFO × 7
- Serial Peripheral Interface (SPI) × 2
- I²C bus interface (IIC) × 2
- CAN module (CAN) × 1
- IrDA interface

■ Analog

- 12-bit A/D Converter (ADC12) with 3 sample-and-hold circuits each × 2
- 12-bit D/A Converter (DAC12) × 2
- High-Speed Analog Comparator (ACMPHS) × 6
- Programmable Gain Amplifier (PGA) × 6
- Temperature Sensor (TSN)

■ Timers

- General PWM Timer 32-bit Enhanced High Resolution (GPT32EH) × 4
- General PWM Timer 32-bit Enhanced (GPT32E) × 4
- General PWM Timer 32-bit (GPT32) × 5
- Low Power Asynchronous General-Purpose Timer (AGT) × 2
- Watchdog Timer (WDT)

■ Safety

- SRAM parity error check
- Flash area protection
- ADC self-diagnosis function
- Clock Frequency Accuracy Measurement Circuit (CAC)
- Cyclic Redundancy Check (CRC) calculator
- Data Operation Circuit (DOC)
- Port Output Enable for GPT (POEG)
- Independent Watchdog Timer (IWDT)
- GPIO readback level detection
- Register write protection
- Main oscillator stop detection
- Illegal memory access

■ System and Power Management

- Low power modes
- Event Link Controller (ELC)
- DMA Controller (DMAC) × 8
- Data Transfer Controller (DTC)
- Key Interrupt Function (KINT)
- Power-on reset
- Low Voltage Detection (LVD) with voltage settings

■ Security and Encryption

- AES128/192/256
- 3DES/ARC4
- SHA1/SHA224/SHA256/MD5
- GHASH
- RSA/DSA/ECC
- True Random Number Generator (TRNG)

■ Multiple Clock Sources

- Main clock oscillator (MOSC) (8 to 24 MHz)
- Sub-clock oscillator (SOSC) (32.768 kHz)
- High-speed on-chip oscillator (HOCO) (16/18/20 MHz)
- Middle-speed on-chip oscillator (MOCO) (8 MHz)
- Low-speed on-chip oscillator (LOCO) (32.768 kHz)
- IWDT-dedicated on-chip oscillator (15 kHz)
- Clock trim function for HOCO/MOCO/LOCO
- Clock out support

■ General-Purpose I/O Ports

- Up to 76 input/output pins
 - Up to 9 CMOS input
 - Up to 67 CMOS input/output
 - Up to 14 input/output 5 V tolerant
 - Up to 13 high current (20 mA)

■ Operating Voltage

- VCC: 2.7 to 3.6 V

■ Operating Temperature and Packages

- Ta = -40°C to +105°C
 - 100-pin LQFP (14 mm × 14 mm, 0.5 mm pitch)
 - 64-pin LQFP (10 mm × 10 mm, 0.5 mm pitch)

1. Overview

The MCU integrates multiple series of software- and pin-compatible Arm[®]-based 32-bit cores that share a common set of Renesas peripherals to facilitate design scalability and efficient platform-based product development.

The MCU in this series incorporates a high-performance Arm Cortex[®]-M4 core running up to 120 MHz with the following features:

- Up to 512-KB code flash memory
- 64-KB SRAM
- Security and safety features
- 12-bit A/D Converter (ADC12)
- 12-bit D/A Converter (DAC12)
- Analog peripherals.

1.1 Function Outline

Table 1.1 Arm core

Feature	Functional description
Arm Cortex-M4 core	<ul style="list-style-type: none"> • Maximum operating frequency: up to 120 MHz • Arm Cortex-M4 core: <ul style="list-style-type: none"> - Revision: r0p1-01rel0 - Armv7E-M architecture profile - Single precision floating-point unit compliant with the ANSI/IEEE Std 754-2008. • Arm Memory Protection Unit (Arm MPU): <ul style="list-style-type: none"> - Armv7 Protected Memory System Architecture - 8 protect regions. • SysTick timer: <ul style="list-style-type: none"> - Driven by SYSTICCLK (LOCO) or ICLK.

Table 1.2 Memory

Feature	Functional description
Code flash memory	Up to 512-KB code flash memory. See section 41, Flash Memory in User's Manual.
Data flash memory	8-KB data flash memory. See section 41, Flash Memory in User's Manual.
Memory Mirror Function (MMF)	The Memory Mirror Function (MMF) can be configured to mirror the target application image load address in code flash memory to the application image link address in the 23-bit unused memory space (memory mirror space addresses). Your application code is developed and linked to run from this MMF destination address. Your application code does not need to know the load location where it is stored in code flash memory. See section 5, Memory Mirror Function (MMF) in User's Manual.
Option-setting memory	The option-setting memory determines the state of the MCU after a reset. See section 7, Option-Setting Memory in User's Manual.
SRAM	On-chip high-speed SRAM. See section 40, SRAM in User's Manual.

Table 1.3 System (1 of 3)

Feature	Functional description
Operating modes	Two operating modes: <ul style="list-style-type: none"> • Single-chip mode • SCI boot mode. See section 3, Operating Modes in User's Manual.

Table 1.3 System (2 of 3)

Feature	Functional description
Resets	<p>14 resets:</p> <ul style="list-style-type: none"> • RES pin reset • Power-on reset • Voltage monitor 0 reset • Voltage monitor 1 reset • Voltage monitor 2 reset • Independent watchdog timer reset • Watchdog timer reset • Deep Software Standby reset • SRAM parity error reset • Bus master MPU error reset • Bus slave MPU error reset • Stack pointer error reset • Software reset. <p>See section 6, Resets in User's Manual.</p>
Low Voltage Detection (LVD)	<p>The Low Voltage Detection (LVD) function monitors the voltage level input to the VCC pin, and the detection level can be selected using a software program. See section 8, Low Voltage Detection (LVD) in User's Manual.</p>
Clocks	<ul style="list-style-type: none"> • Main clock oscillator (MOSC) • Sub-clock oscillator (SOSC) • High-speed on-chip oscillator (HOCO) • Middle-speed on-chip oscillator (MOCO) • Low-speed on-chip oscillator (LOCO) • PLL frequency synthesizer • IDWT-dedicated on-chip oscillator • Clock out support. <p>See section 9, Clock Generation Circuit in User's Manual.</p>
Clock Frequency Accuracy Measurement Circuit (CAC)	<p>The Clock Frequency Accuracy Measurement Circuit (CAC) counts pulses of the clock to be measured (measurement target clock) within the time generated by the clock to be used as a measurement reference (measurement reference clock), and determines the accuracy depending on whether the number of pulses is within the allowable range.</p> <p>When measurement is complete or the number of pulses within the time generated by the measurement reference clock is not within the allowable range, an interrupt request is generated.</p> <p>See section 10, Clock Frequency Accuracy Measurement Circuit (CAC) in User's Manual.</p>
Interrupt Controller Unit (ICU)	<p>The Interrupt Controller Unit (ICU) controls which event signals are linked to the NVIC/DTC module and DMAC module. The ICU also controls NMI interrupts. See section 13, Interrupt Controller Unit (ICU) in User's Manual.</p>
Key Interrupt Function (KINT)	<p>A key interrupt can be generated by setting the Key Return Mode Register (KRM) and inputting a rising or falling edge to the key interrupt input pins. See section 20, Key Interrupt Function (KINT) in User's Manual.</p>
Low power modes	<p>Power consumption can be reduced in multiple ways, such as by setting clock dividers, stopping modules, selecting power control mode in normal operation, and transitioning to low power modes. See section 11, Low Power Modes in User's Manual.</p>
Register write protection	<p>The register write protection function protects important registers from being overwritten because of software errors. See section 12, Register Write Protection in User's Manual.</p>
Memory Protection Unit (MPU)	<p>Four Memory Protection Units (MPUs) and a CPU stack pointer monitor function are provided for memory protection. See section 15, Memory Protection Unit (MPU) in User's Manual.</p>
Watchdog Timer (WDT)	<p>The Watchdog Timer (WDT) is a 14-bit down-counter that can be used to reset the MCU when the counter underflows because the system has run out of control and is unable to refresh the WDT. In addition, a non-maskable interrupt or interrupt can be generated by an underflow. A refresh-permitted period can be set to refresh the counter and used as the condition for detecting when the system runs out of control. See section 25, Watchdog Timer (WDT) in User's Manual.</p>

Table 1.3 System (3 of 3)

Feature	Functional description
Independent Watchdog Timer (IWDT)	The Independent Watchdog Timer (IWDT) consists of a 14-bit down-counter that must be serviced periodically to prevent counter underflow. The IWDT provides functionality to reset the MCU or to generate a non-maskable interrupt or interrupt for a timer underflow. Because the timer operates with an independent, dedicated clock source, it is particularly useful in returning the MCU to a known state as a fail-safe mechanism when the system runs out of control. The IWDT can be triggered automatically on a reset, underflow, or refresh error, or by a refresh of the count value in the registers. See section 26, Independent Watchdog Timer (IWDT) in User's Manual.

Table 1.4 Event link

Feature	Functional description
Event Link Controller (ELC)	The Event Link Controller (ELC) uses the interrupt requests generated by various peripheral modules as event signals to connect them to different modules, enabling direct interaction between the modules without CPU intervention. See section 18, Event Link Controller (ELC) in User's Manual.

Table 1.5 Direct memory access

Feature	Functional description
Data Transfer Controller (DTC)	A Data Transfer Controller (DTC) module is provided for transferring data when activated by an interrupt request. See section 17, Data Transfer Controller (DTC) in User's Manual.
DMA Controller (DMAC)	An 8-channel DMA Controller (DMAC) module is provided for transferring data without the CPU. When a DMA transfer request is generated, the DMAC transfers data stored at the transfer source address to the transfer destination address. See section 16, DMA Controller (DMAC) in User's Manual.

Table 1.6 Timers

Feature	Functional description
General PWM Timer (GPT)	The General PWM Timer (GPT) is a 32-bit timer with 13 channels. PWM waveforms can be generated by controlling the up-counter, down-counter, or up- and down-counter. In addition, PWM waveforms can be generated for controlling brushless DC motors. The GPT can also be used as a general-purpose timer. See section 22, General PWM Timer (GPT) in User's Manual.
Port Output Enable for GPT (POEG)	Use the Port Output Enable for GPT (POEG) function to place the General PWM Timer (GPT) output pins in the output disable state. See section 21, Port Output Enable for GPT (POEG) in User's Manual.
Low Power Asynchronous General-Purpose Timer (AGT)	The Low Power Asynchronous General-Purpose Timer (AGT) is a 16-bit timer that can be used for pulse output, external pulse width or period measurement, and counting of external events. This 16-bit timer consists of a reload register and a down-counter. The reload register and the down-counter are allocated to the same address, and can be accessed with the AGT register. See section 24, Low Power Asynchronous General-Purpose Timer (AGT) in User's Manual.

Table 1.7 Communication interfaces

Feature	Functional description
Serial Communications Interface (SCI)	<p>The Serial Communications Interface (SCI) is configurable to five asynchronous and synchronous serial interfaces:</p> <ul style="list-style-type: none"> • Asynchronous interfaces (UART and Asynchronous Communications Interface Adapter (ACIA)) • 8-bit clock synchronous interface • Simple IIC (master-only) • Simple SPI • Smart card interface. <p>The smart card interface complies with the ISO/IEC 7816-3 standard for electronic signals and transmission protocol.</p> <p>Each SCI has FIFO buffers to enable continuous and full-duplex communication, and the data transfer speed can be configured independently using an on-chip baud rate generator. See section 27, Serial Communications Interface (SCI) in User's Manual.</p>
IrDA Interface (IrDA)	<p>The IrDA interface sends and receives IrDA data communication waveforms in cooperation with the SCI1 based on the IrDA (Infrared Data Association) standard 1.0. See section 28, IrDA Interface in User's Manual.</p>
I ² C bus interface (IIC)	<p>The 2-channel I²C bus interface (IIC) conforms with and provides a subset of the NXP I²C (Inter-Integrated Circuit) bus interface functions. See section 29, I²C Bus Interface (IIC) in User's Manual.</p>
Serial Peripheral Interface (SPI)	<p>Two independent Serial Peripheral Interface (SPI) channels are capable of high-speed, full-duplex synchronous serial communications with multiple processors and peripheral devices. See section 31, Serial Peripheral Interface (SPI) in User's Manual.</p>
Controller Area Network (CAN) module	<p>The Controller Area Network (CAN) module provides functionality to receive and transmit data using a message-based protocol between multiple slaves and masters in electromagnetically-noisy applications.</p> <p>The CAN module complies with the ISO 11898-1 (CAN 2.0A/CAN 2.0B) standard and supports up to 32 mailboxes, which can be configured for transmission or reception in normal mailbox and FIFO modes. Both standard (11-bit) and extended (29-bit) messaging formats are supported. See section 30, Controller Area Network (CAN) Module in User's Manual.</p>

Table 1.8 Analog

Feature	Functional description
12-bit A/D Converter (ADC12)	<p>Two units of successive approximation 12-bit A/D Converter (ADC12) are provided. Analog input channels are selectable up to 11 in unit 0 and up to 8 in unit 1. Each 2 analog inputs of unit 0 and 1 are assigned to same port (AN005/AN105, AN006/AN106), up to 17 ports are available as analog input. The temperature sensor output and an internal reference voltage are selectable for conversion of each unit 0 and 1. The A/D conversion accuracy is selectable from 12-bit, 10-bit, and 8-bit conversion, making it possible to optimize the tradeoff between speed and resolution in generating a digital value. See section 35, 12-Bit A/D Converter (ADC12) in User's Manual.</p>
12-bit D/A Converter (DAC12)	<p>A 12-bit D/A Converter (DAC12) converts data and includes an output amplifier. See section 36, 12-Bit D/A Converter (DAC12) in User's Manual.</p>
Temperature Sensor (TSN)	<p>The on-chip Temperature Sensor (TSN) can determine and monitor the die temperature for reliable operation of the device. The sensor outputs a voltage directly proportional to the die temperature, and the relationship between the die temperature and the output voltage is linear. The output voltage is provided to the ADC12 for conversion and can also be used by the end application. See section 37, Temperature Sensor (TSN) in User's Manual.</p>
High-Speed Analog Comparator (ACMPHS)	<p>The High-Speed Analog Comparator (ACMPHS) compares a test voltage with a reference voltage and provides a digital output based on the conversion result.</p> <p>Both the test and reference voltages can be provided to the comparator from internal sources such as the DAC12 output and internal reference voltage, and an external source with or without an internal PGA.</p> <p>Such flexibility is useful in applications that require go/no-go comparisons to be performed between analog signals without necessarily requiring A/D conversion. See section 38, High-Speed Analog Comparator (ACMPHS) in User's Manual.</p>

Table 1.9 Data processing

Feature	Functional description
Cyclic Redundancy Check (CRC) calculator	The Cyclic Redundancy Check (CRC) calculator generates CRC codes to detect errors in the data. The bit order of CRC calculation results can be switched for LSB-first or MSB-first communication. Additionally, various CRC-generating polynomials are available. The snoop function allows monitoring reads from and writes to specific addresses. This function is useful in applications that require CRC code to be generated automatically in certain events, such as monitoring writes to the serial transmit buffer and reads from the serial receive buffer. See section 32, Cyclic Redundancy Check (CRC) Calculator in User's Manual.
Data Operation Circuit (DOC)	The Data Operation Circuit (DOC) compares, adds, and subtracts 16-bit data. See section 39, Data Operation Circuit (DOC) in User's Manual.

Table 1.10 Security

Feature	Functional description
Secure Crypto Engine 7 (SCE7)	<ul style="list-style-type: none"> • Security algorithms: <ul style="list-style-type: none"> - Symmetric algorithms: AES, 3DES, and ARC4 - Asymmetric algorithms: RSA, DSA, and ECC. • Other support features: <ul style="list-style-type: none"> - TRNG (True Random Number Generator) - Hash-value generation: SHA1, SHA224, SHA256, GHASH, and MD5 - 128-bit unique ID.

Table 1.11 I/O ports

Feature	Functional description
I/O ports	<ul style="list-style-type: none"> • I/O ports for the 100-pin LQFP <ul style="list-style-type: none"> - I/O pins: 67 - Input pins: 9 - Pull-Up resistors: 68 - N-ch open-drain outputs: 67 - 5-V tolerance: 14 • I/O ports for the 64-pin LQFP <ul style="list-style-type: none"> - I/O pins: 35 - Input pins: 5 - Pull-Up resistors: 36 - N-ch open-drain outputs: 35 - 5-V tolerance: 9

1.2 Block Diagram

Figure 1.1 shows a block diagram of the MCU superset, some individual devices within the group have a subset of the features.

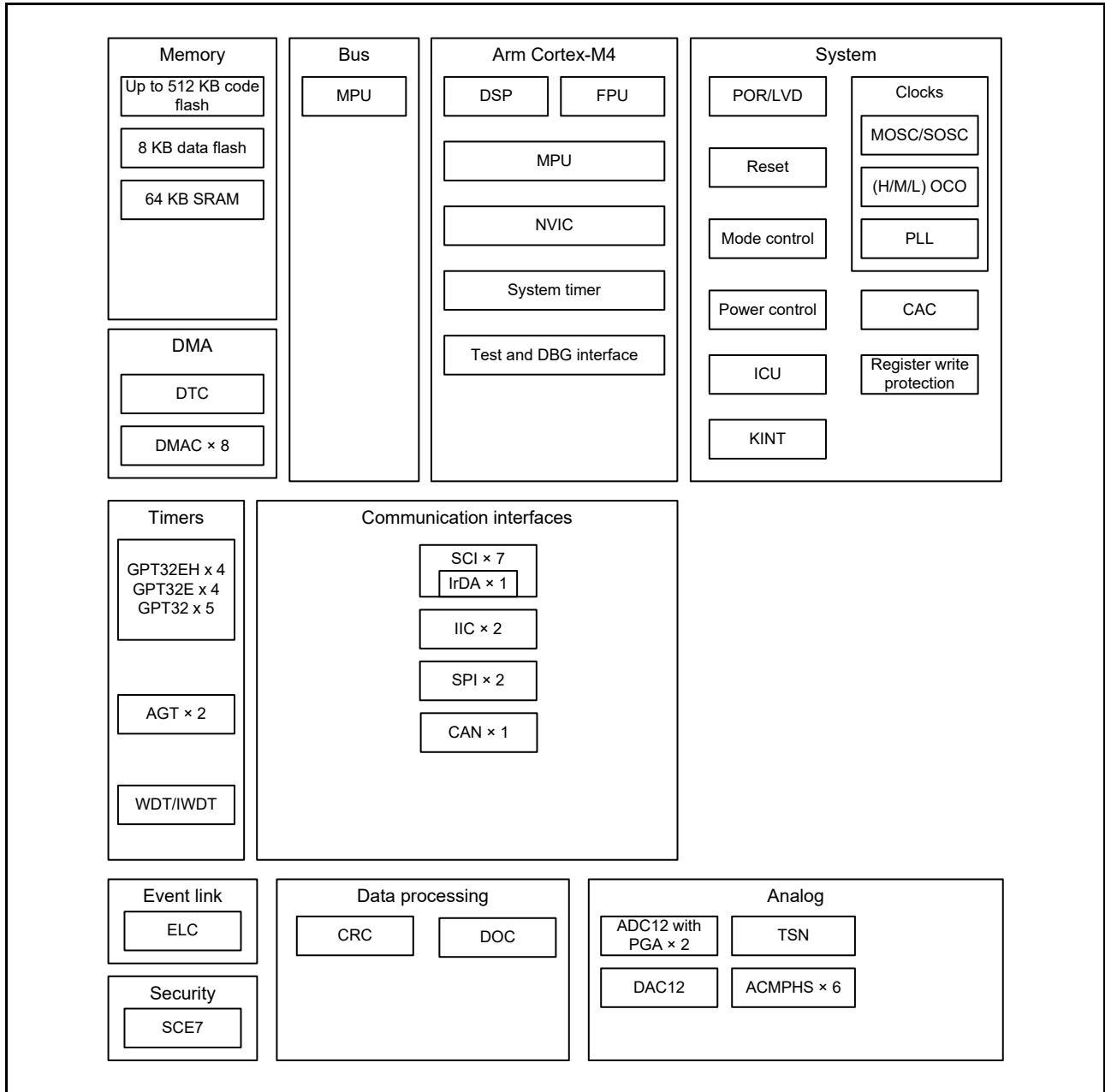


Figure 1.1 Block diagram

1.3 Part Numbering

Figure 1.2 shows the product part number information, including memory capacity and package type. Table 1.12 shows a list of products.

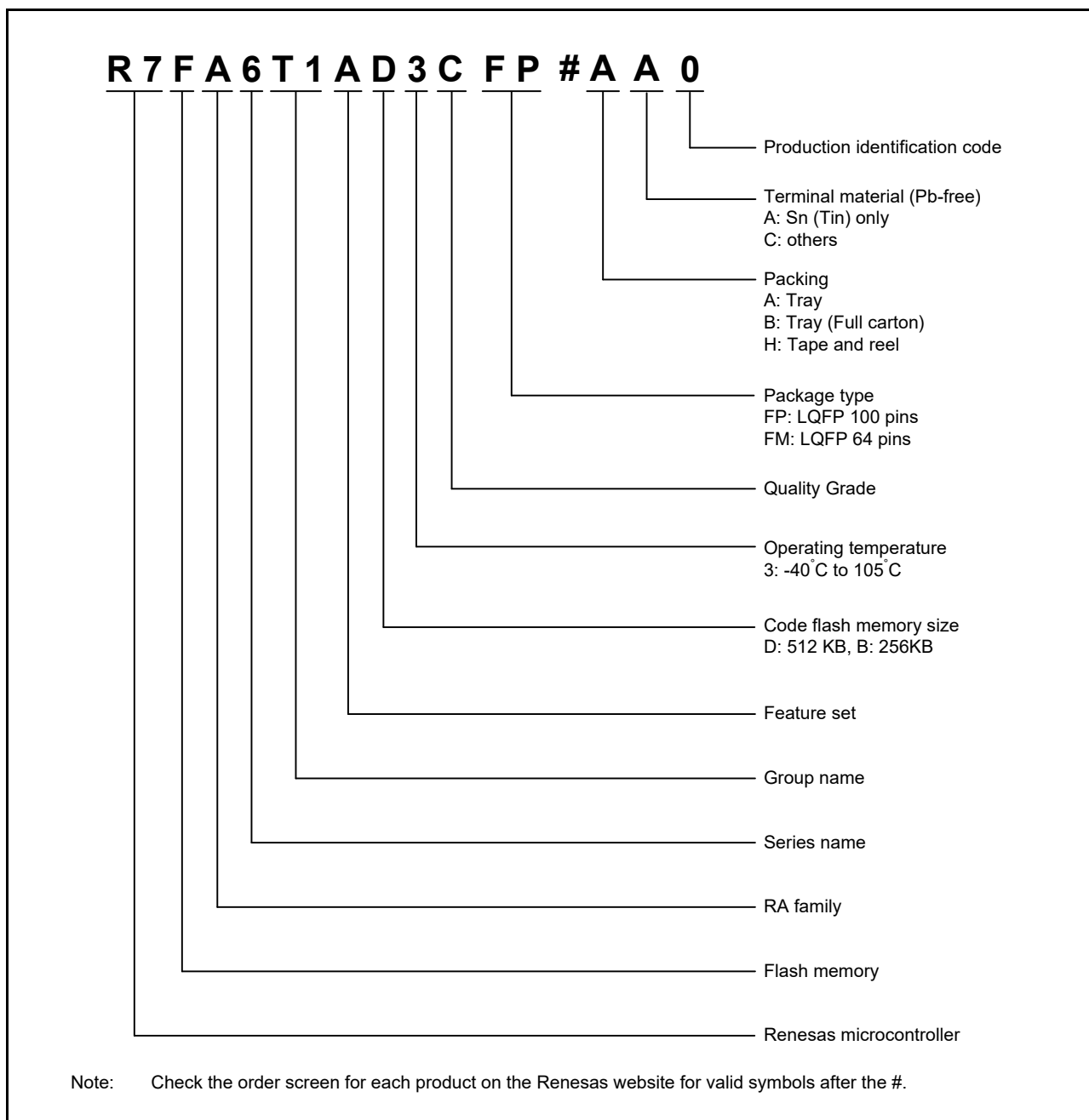


Figure 1.2 Part numbering scheme

Table 1.12 Product list

Product part number	Package code	Code flash	Data flash	SRAM	Operating temperature
R7FA6T1AD3CFP	PLQP0100KB-B	512 KB	8 KB	64 KB	-40 to +105°C
R7FA6T1AB3CFP	PLQP0100KB-B	256 KB			-40 to +105°C
R7FA6T1AD3CFM	PLQP0064KB-C	512 KB			-40 to +105°C
R7FA6T1AB3CFM	PLQP0064KB-C	256 KB			-40 to +105°C

1.4 Function Comparison

Table 1.13 Functional comparison

Function	Part numbers			
	R7FA6T1AD3CFP	R7FA6T1AB3CFP	R7FA6T1AD3CFM	R7FA6T1AB3CFM
Pin count	100		64	
Package	LQFP			
Code flash memory	512 KB	256 KB	512 KB	256 KB
Data flash memory	8 KB			
SRAM	64 KB			
	Parity	64 KB		
System	CPU clock	120 MHz		
	Backup registers	512 Bytes		
	ICU	Yes		
	KINT	8		
Event link	ELC	Yes		
DMA	DTC	Yes		
	DMAC	8		
Timers	GPT32EH	4		
	GPT32E	4	3	
	GPT32	5	4	
	AGT	2		
	WDT/IWDT	Yes		
Communication	SCI	7		
	IIC	2		
	SPI	2		
	CAN	1		
Analog	ADC12	Unit0: 11 Unit1: 8 Shared channel pin: 2*1		Unit0: 7 Unit1: 3 Shared channel pin: 2*1
	3ch-S/H	Unit0: 1 (3ch) Unit1: 1 (3ch)		Unit0: 1 (3ch)
	PGA	Unit0: 3 Unit1: 3		Unit0: 3
	DAC12	2		
	ACMPHS	6		
	TSN	Yes		
Data processing	CRC	Yes		
	DOC	Yes		
Security	SCE7			
I/O Ports	I/O Pins	67		35
	Input pins	9		5
	Pull-up resistors	68		36
	open-drain outputs	67		35
	5-V tolerance	14		9

Note 1. Some input channels of the ADC units are sharing same port pin.

1.5 Pin Functions

Table 1.14 Pin functions (1 of 3)

Function	Signal	I/O	Description
Power supply	VCC	Input	Power supply pin. This is used as the digital power supply for the respective modules and internal voltage regulator, and used to monitor the voltage of the POR/LVD. Connect this pin to the system power supply. Connect it to VSS by a 0.1- μ F capacitor. Place the capacitor close to the pin.
	VCL0	Input	Connect this pin to VSS through a 0.1- μ F smoothing capacitor used to stabilize the internal power supply. Place the capacitor close to the pin.
	VCL	Input	
	VSS	Input	Ground pin. Connect to the system power supply (0 V).
Clock	XTAL	Output	Pins for a crystal resonator. An external clock signal can be input through the XTAL pin.
	EXTAL	Input	
	XCIN	Input	Input/output pins for the sub-clock oscillator. Connect a crystal resonator between XCOU and XCIN.
	XCOU	Output	
	CLKOUT	Output	Clock output pin
Operating mode control	MD	Input	Pin for setting the operating mode. The signal level on this pin must not be changed during operation mode transition on release from the reset state.
System control	RES	Input	Reset signal input pin. The MCU enters the reset state when this signal goes low.
CAC	CACREF	Input	Measurement reference clock input pin
Interrupt	NMI	Input	Non-maskable interrupt request pin
	IRQ0 to IRQ13	Input	Maskable interrupt request pins
KINT	KR00 to KR07	Input	A key interrupt can be generated by inputting a falling edge to the key interrupt input pins
On-chip emulator	TMS	I/O	On-chip emulator or boundary scan pins
	TDI	Input	
	TCK	Input	
	TDO	Output	
	TCLK	Output	This pin outputs the clock for synchronization with the trace data
	TDATA0 to TDATA3	Output	Trace data output
	SWDIO	I/O	Serial wire debug data input/output pin
	SWCLK	Input	Serial wire clock pin
	SWO	Output	Serial wire trace output pin
	GPT	GTETRGA, GTETRGB, GTETRGC, GTETRGD	Input
GTIOC0A to GTIOC12A, GTIOC0B to GTIOC12B		I/O	Input capture, output compare, or PWM output pins
GTIU		Input	Hall sensor input pin U
GTIV		Input	Hall sensor input pin V
GTIW		Input	Hall sensor input pin W
GTOUUP		Output	3-phase PWM output for BLDC motor control (positive U phase)
GTOULO		Output	3-phase PWM output for BLDC motor control (negative U phase)
GTOVUP		Output	3-phase PWM output for BLDC motor control (positive V phase)
GTOVLO		Output	3-phase PWM output for BLDC motor control (negative V phase)
GTOWUP		Output	3-phase PWM output for BLDC motor control (positive W phase)
GTOWLO		Output	3-phase PWM output for BLDC motor control (negative W phase)
AGT	AGTEE0, AGTEE1	Input	External event input enable signals
	AGTIO0, AGTIO1	I/O	External event input and pulse output pins
	AGTO0, AGTO1	Output	Pulse output pins
	AGTOA0, AGTOA1	Output	Output compare match A output pins
	AGTOB0, AGTOB1	Output	Output compare match B output pins

Table 1.14 Pin functions (2 of 3)

Function	Signal	I/O	Description
SCI	SCK0 to SCK4, SCK8, SCK9	I/O	Input/output pins for the clock (clock synchronous mode)
	RXD0 to RXD4, RXD8, RXD9	Input	Input pins for received data (asynchronous mode/clock synchronous mode)
	TXD0 to TXD4, TXD8, TXD9	Output	Output pins for transmitted data (asynchronous mode/clock synchronous mode)
	CTS0_RTS0 to CTS4_RTS4, CTS8_RTS8, CTS9_RTS9	I/O	Input/output pins for controlling the start of transmission and reception (asynchronous mode/clock synchronous mode), active-low
	SCL0 to SCL4, SCL8, SCL9	I/O	Input/output pins for the IIC clock (simple IIC mode)
	SDA0 to SDA4, SDA8, SDA9	I/O	Input/output pins for the IIC data (simple IIC mode)
	SCK0 to SCK4, SCK8, SCK9	I/O	Input/output pins for the clock (simple SPI mode)
	MISO0 to MISO4, MISO8, MISO9	I/O	Input/output pins for slave transmission of data (simple SPI mode)
	MOSI0 to MOSI4, MOSI8, MOSI9	I/O	Input/output pins for master transmission of data (simple SPI mode)
	SS0 to SS4, SS8, SS9	Input	Chip-select input pins (simple SPI mode), active-low
IIC	SCL0, SCL1	I/O	Input/output pins for the clock
	SDA0, SDA1	I/O	Input/output pins for data
SPI	RSPCKA, RSPCKB	I/O	Clock input/output pin
	MOSIA, MOSIB	I/O	Input or output pins for data output from the master
	MISOA, MISOB	I/O	Input or output pins for data output from the slave
	SSLA0, SSLB0	I/O	Input or output pin for slave selection
	SSLA1 to SSLA3, SSLB1 to SSLB3	Output	Output pins for slave selection
CAN	CRX0	Input	Receive data
	CTX0	Output	Transmit data
Analog power supply	AVCC0	Input	Analog voltage supply pin. This is used as the analog power supply for the respective modules. Supply this pin with the same voltage as the VCC pin.
	AVSS0	Input	Analog ground pin. This is used as the analog ground for the respective modules. Supply this pin with the same voltage as the VSS pin.
	VREFH0	Input	Analog reference voltage supply pin for the ADC12 (unit 0). Connect this pin to VCC when not using the ADC12 (unit 0) and sample-and-hold circuit for AN000 to AN002.
	VREFL0	Input	Analog reference ground pin for the ADC12. Connect this pin to VSS when not using the ADC12 (unit 0) and sample-and-hold circuit for AN000 to AN002
	VREFH	Input	Analog reference voltage supply pin for the ADC12 (unit 1) and D/A Converter. Connect this pin to VCC when not using the ADC12 (unit 1), sample-and-hold circuit for AN100 to AN102, and D/A Converter.
	VREFL	Input	Analog reference ground pin for the ADC12 and D/A Converter. Connect this pin to VSS when not using the ADC12 (unit 1), sample-and-hold circuit for AN100 to AN102, and D/A Converter.
ADC12	AN000 to AN003, AN005 to AN007, AN016 to AN018, AN020	Input	Input pins for the analog signals to be processed by the ADC12. AN005 & AN105 and AN006 & AN106 are assigned to same port pin
	AN100 to AN102, AN105 to AN107, AN116, AN117	Input	
	ADTRG0	Input	Input pins for the external trigger signals that start the A/D conversion
	ADTRG1	Input	
	PGAVSS000, PGAVSS100	Input	Pseudo-differential input pins

Table 1.14 Pin functions (3 of 3)

Function	Signal	I/O	Description
DAC12	DA0, DA1	Output	Output pins for the analog signals processed by the D/A converter
ACMPHS	VCOUT	Output	Comparator output pin
	IVREF0 to IVREF3	Input	Reference voltage input pins for comparator
	IVCMP0 to IVCMP3	Input	Analog voltage input pins for comparator
I/O ports	P000 to P007	Input	General-purpose input pins
	P008, P014, P015	I/O	General-purpose input/output pins
	P100 to P115	I/O	General-purpose input/output pins
	P200	Input	General-purpose input pin
	P201, P205 to P214	I/O	General-purpose input/output pins
	P300 to P307	I/O	General-purpose input/output pins
	P400 to P415	I/O	General-purpose input/output pins
	P500 to P504, P508	I/O	General-purpose input/output pins
	P600 to P602, P608 to P610	I/O	General-purpose input/output pins
	P708	I/O	General-purpose input/output pin

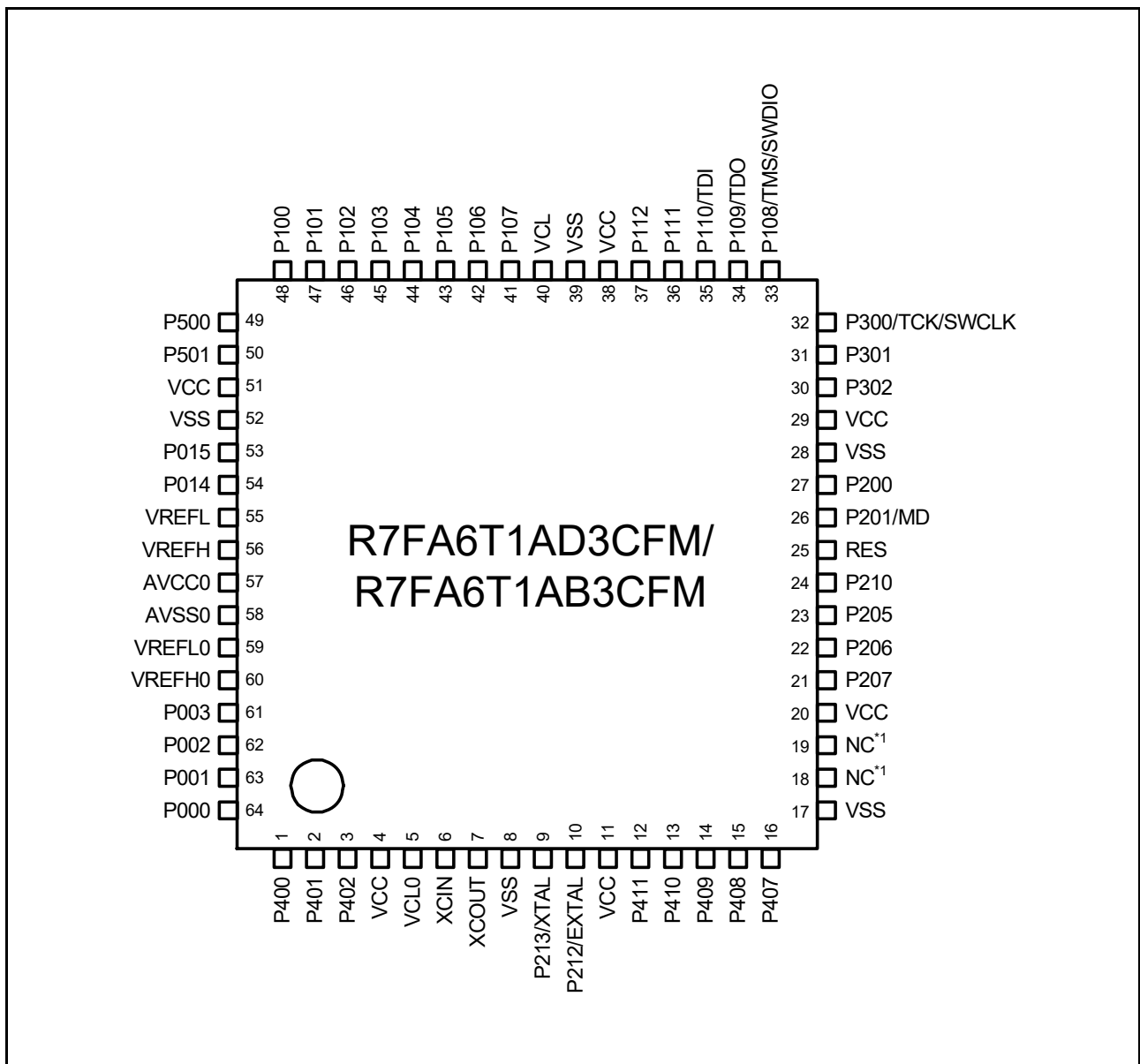


Figure 1.4 Pin assignment for 64-pin LQFP (top view)

Note 1. This pin should be left floating.

1.7 Pin Lists

Pin number		Power, System, Clock, Debug, CAC	Interrupt	I/O port	Timers			Communication interfaces					Analog	
LQFP100	LQFP64				AGT	GPT	GPT	CAN	SCI0,2,4,8 (30 MHz)	SCI1,3,9 (30 MHz)	IIC	SPI	ADC12	DAC12, ACMPHS
1	1	-	IRQ0	P400	AGTIO1	-	GTIOC6A	-	SCK4	-	SCL0_A	-	ADTRG1	-
2	2	-	IRQ5-DS	P401	-	GTETRGA	GTIOC6B	CTX0	CTS4_RTS4/S4	-	SDA0_A	-	-	-
3	3	CACREF	IRQ4-DS	P402	AGTIO0/AGTIO1	-	-	CRX0	-	-	-	-	-	-
4	-	-	-	P403	AGTIO0/AGTIO1	-	GTIOC3A	-	-	-	-	-	-	-
5	-	-	-	P404	-	-	GTIOC3B	-	-	-	-	-	-	-
6	-	-	-	P405	-	-	GTIOC1A	-	-	-	-	-	-	-
7	-	-	-	P406	-	-	GTIOC1B	-	-	-	-	-	-	-
8	4	VCC	-	-	-	-	-	-	-	-	-	-	-	-
9	5	VCL0	-	-	-	-	-	-	-	-	-	-	-	-
10	6	XCIN	-	-	-	-	-	-	-	-	-	-	-	-
11	7	XCOUT	-	-	-	-	-	-	-	-	-	-	-	-
12	8	VSS	-	-	-	-	-	-	-	-	-	-	-	-
13	9	XTAL	IRQ2	P213	-	GTETRGC	GTIOC0A	-	-	TXD1/MOSI1/S-DA1	-	ADTRG1	-	-
14	10	EXTAL	IRQ3	P212	AGTEE1	GTETRGD	GTIOC0B	-	-	RXD1/MISO1/S-CL1	-	-	-	-
15	11	VCC	-	-	-	-	-	-	-	-	-	-	-	-
16	-	CACREF	IRQ11	P708	-	-	-	-	-	RXD1/MISO1/S-CL1	-	SSLA3_B	-	-
17	-	-	IRQ8	P415	-	-	GTIOC0A	-	-	-	-	SSLA2_B	-	-
18	-	-	IRQ9	P414	-	-	GTIOC0B	-	-	-	-	SSLA1_B	-	-
19	-	-	-	P413	-	GTOUUP	-	-	CTS0_RTS0/S0	-	-	SSLA0_B	-	-
20	-	-	-	P412	AGTEE1	GTOULO	-	-	SCK0	-	-	RSPCKA_B	-	-
21	12	-	IRQ4	P411	AGTOA1	GTOVUP	GTIOC9A	-	TXD0/MOSI0/S-DA0	CTS3_RTS3/S3	-	MOSIA_B	-	-
22	13	-	IRQ5	P410	AGTOB1	GTOVLO	GTIOC9B	-	RXD0/MISO0/S-CL0	SCK3	-	MISOA_B	-	-
23	14	-	IRQ6	P409	-	GTOUUP	GTIOC10A	-	-	TXD3/MOSI3/S-DA3	-	-	-	-
24	15	-	IRQ7	P408	-	GTOVLO	GTIOC10B	-	-	RXD3/MISO3/S-CL3	SCL0_B	-	-	-
25	16	-	-	P407	AGTIO0	-	-	-	CTS4_RTS4/S4	-	SDA0_B	-	ADTRG0	-
26	17	VSS	-	-	-	-	-	-	-	-	-	-	-	-
27	18	-	-	-	-	-	-	-	-	-	-	-	-	-
28	19	-	-	-	-	-	-	-	-	-	-	-	-	-
29	20	VCC	-	-	-	-	-	-	-	-	-	-	-	-
30	21	-	-	P207	-	-	-	-	-	-	-	-	-	-
31	22	-	IRQ0-DS	P206	-	GTIU	-	-	RXD4/MISO4/S-CL4	-	SDA1_A	-	-	-
32	23	CLKOUT	IRQ1-DS	P205	AGTO1	GTIV	GTIOC4A	-	TXD4/MOSI4/S-DA4	CTS9_RTS9/S9	SCL1_A	-	-	-
33	-	TCLK	-	P214	-	GTIU	-	-	-	-	-	-	-	-
34	-	TDATA0	-	P211	-	GTIV	-	-	-	-	-	-	-	-
35	24	TDATA1	-	P210	-	GTIW	-	-	-	-	-	-	-	-
36	-	TDATA2	-	P209	-	GTOVUP	-	-	-	-	-	-	-	-
37	-	TDATA3	-	P208	-	GTOVLO	-	-	-	-	-	-	-	-
38	25	RES	-	-	-	-	-	-	-	-	-	-	-	-
39	26	MD	-	P201	-	-	-	-	-	-	-	-	-	-
40	27	-	NMI	P200	-	-	-	-	-	-	-	-	-	-
41	-	-	-	P307	-	GTOUUP	-	-	-	-	-	-	-	-
42	-	-	-	P306	-	GTOULO	-	-	-	-	-	-	-	-
43	-	-	IRQ8	P305	-	GTOUUP	-	-	-	-	-	-	-	-
44	-	-	IRQ9	P304	-	GTOVLO	GTIOC7A	-	-	-	-	-	-	-
45	28	VSS	-	-	-	-	-	-	-	-	-	-	-	-
46	29	VCC	-	-	-	-	-	-	-	-	-	-	-	-
47	-	-	-	P303	-	-	GTIOC7B	-	-	-	-	-	-	-
48	30	-	IRQ5	P302	-	GTOUUP	GTIOC4A	-	TXD2/MOSI2/S-DA2	-	SSLB3_B	-	-	-
49	31	-	IRQ6	P301	AGTIO0	GTOULO	GTIOC4B	-	RXD2/MISO2/S-CL2	CTS9_RTS9/S9	SSLB2_B	-	-	-
50	32	TCK/SWCLK	-	P300	-	GTOUUP	GTIOC0A_A	-	-	-	SSLB1_B	-	-	-
51	33	TMS/SWDIO	-	P108	-	GTOULO	GTIOC0B_A	-	-	CTS9_RTS9/S9	SSLB0_B	-	-	-
52	34	CLKOUT/TDO/SWO	-	P109	-	GTOUUP	GTIOC1A_A	-	-	TXD9/MOSI9/S-DA9	MOSIB_B	-	-	-
53	35	TDI	IRQ3	P110	-	GTOVLO	GTIOC1B_A	-	CTS2_RTS2/S2	RXD9/MISO9/S-CL9	MISOB_B	-	VCOU	-
54	36	-	IRQ4	P111	-	-	GTIOC3A_A	-	SCK2	SCK9	-	RSPCKB_B	-	-
55	37	-	-	P112	-	-	GTIOC3B_A	-	TXD2/MOSI2/S-DA2	SCK1	-	SSLB0_B	-	-
56	-	-	-	P113	-	-	GTIOC2A	-	RXD2/MISO2/S-CL2	-	-	-	-	-
57	-	-	-	P114	-	-	GTIOC2B	-	-	-	-	-	-	-
58	-	-	-	P115	-	-	GTIOC4A	-	-	-	-	-	-	-
59	-	-	-	P608	-	-	GTIOC4B	-	-	-	-	-	-	-

Pin number		Power, System, Clock, Debug, CAC	Interrupt	I/O port	Timers			Communication interfaces					Analog	
LQFP100	LQFP64				AGT	GPT	GPT	CAN	SCI0,2,4,8 (30 MHz)	SCI1,3,9 (30 MHz)	IIC	SPI	ADC12	DAC12, ACMPHS
60	-	-	-	P609	-	-	GTIOC5A	-	-	-	-	-	-	-
61	-	-	-	P610	-	-	GTIOC5B	-	-	-	-	-	-	-
62	38	VCC	-	-	-	-	-	-	-	-	-	-	-	-
63	39	VSS	-	-	-	-	-	-	-	-	-	-	-	-
64	40	VCL	-	-	-	-	-	-	-	-	-	-	-	-
65	-	-	-	P602	-	-	GTIOC7B	-	-	-	TXD9	-	-	-
66	-	-	-	P601	-	-	GTIOC6A	-	-	-	RXD9	-	-	-
67	-	CLKOUT/CAC REF	-	P600	-	-	GTIOC6B	-	-	-	SCK9	-	-	-
68	41	-	KR07	P107	AGTOA0	-	GTIOC8A	-	CTS8_RTS8/S S8	-	-	-	-	-
69	42	-	KR06	P106	AGTOB0	-	GTIOC8B	-	SCK8	-	-	SSLA3_A	-	-
70	43	-	IRQ0/KR05	P105	-	GTETRGA	GTIOC1A	-	TXD8/MOSI8/S DA8	-	-	SSLA2_A	-	-
71	44	-	IRQ1/KR04	P104	-	GTETRGB	GTIOC1B	-	RXD8/MISO8/S CL8	-	-	SSLA1_A	-	-
72	45	-	KR03	P103	-	GTOWUP	GTIOC2A_A	CTX0	CTS0_RTS0/S S0	-	-	SSLA0_A	-	-
73	46	-	KR02	P102	AGTO0	GTOWLO	GTIOC2B_A	CRX0	SCK0	-	-	RSPCKA_A	ADTRG0	-
74	47	-	IRQ1/KR01	P101	AGTEE0	GTETRGB	GTIOC5A	-	TXD0/MOSI0/S DA0	CTS1_RTS1/S S1	SDA1_B	MOSIA_A	-	-
75	48	-	IRQ2/KR00	P100	AGTIO0	GTETRGA	GTIOC5B	-	RXD0/MISO0/S CL0	SCK1	SCL1_B	MISOA_A	-	-
76	49	-	-	P500	AGTOA0	GTIU	GTIOC11A	-	-	-	-	-	AN016	IVREF0
77	50	-	IRQ11	P501	AGTOB0	GTIV	GTIOC11B	-	-	-	-	-	AN116	IVREF1
78	-	-	IRQ12	P502	-	GTIW	GTIOC12A	-	-	-	-	-	AN017	IVCMP0
79	-	-	-	P503	-	GTETRGC	GTIOC12B	-	-	-	-	-	AN117	-
80	-	-	-	P504	-	GTETRGD	-	-	-	-	-	-	AN018	-
81	-	-	-	P508	-	-	-	-	-	-	-	-	AN020	-
82	51	VCC	-	-	-	-	-	-	-	-	-	-	-	-
83	52	VSS	-	-	-	-	-	-	-	-	-	-	-	-
84	53	-	IRQ13	P015	-	-	-	-	-	-	-	-	AN006/AN106	DA1/IVCMP1
85	54	-	-	P014	-	-	-	-	-	-	-	-	AN005/AN105	DA0/IVREF3
86	55	VREFL	-	-	-	-	-	-	-	-	-	-	-	-
87	56	VREFH	-	-	-	-	-	-	-	-	-	-	-	-
88	57	AVCC0	-	-	-	-	-	-	-	-	-	-	-	-
89	58	AVSS0	-	-	-	-	-	-	-	-	-	-	-	-
90	59	VREFL0	-	-	-	-	-	-	-	-	-	-	-	-
91	60	VREFH0	-	-	-	-	-	-	-	-	-	-	-	-
92	-	-	IRQ12-DS	P008	-	-	-	-	-	-	-	-	AN003	-
93	-	-	-	P007	-	-	-	-	-	-	-	-	PGAVSS100/AN107	-
94	-	-	IRQ11-DS	P006	-	-	-	-	-	-	-	-	AN102	IVCMP2
95	-	-	IRQ10-DS	P005	-	-	-	-	-	-	-	-	AN101	IVCMP2
96	-	-	IRQ9-DS	P004	-	-	-	-	-	-	-	-	AN100	IVCMP2
97	61	-	-	P003	-	-	-	-	-	-	-	-	PGAVSS000/AN007	-
98	62	-	IRQ8-DS	P002	-	-	-	-	-	-	-	-	AN002	IVCMP2
99	63	-	IRQ7-DS	P001	-	-	-	-	-	-	-	-	AN001	IVCMP2
100	64	-	IRQ6-DS	P000	-	-	-	-	-	-	-	-	AN000	IVCMP2

Note: Some pin names have the added suffix of _A and _B. When assigning the GPT, IIC, and SPI functionality, select the functional pins with the same suffix.

2. Electrical Characteristics

Unless otherwise specified, the electrical characteristics of the MCU are defined under the following conditions:

- $VCC = AVCC0 = 2.7$ to 3.6 V
- $2.7 \leq VREFH0/VREFH \leq AVCC0$
- $VSS = AVSS0 = VREFL0/VREFL = 0$ V
- $T_a = T_{opr}$

Figure 2.1 shows the timing conditions.

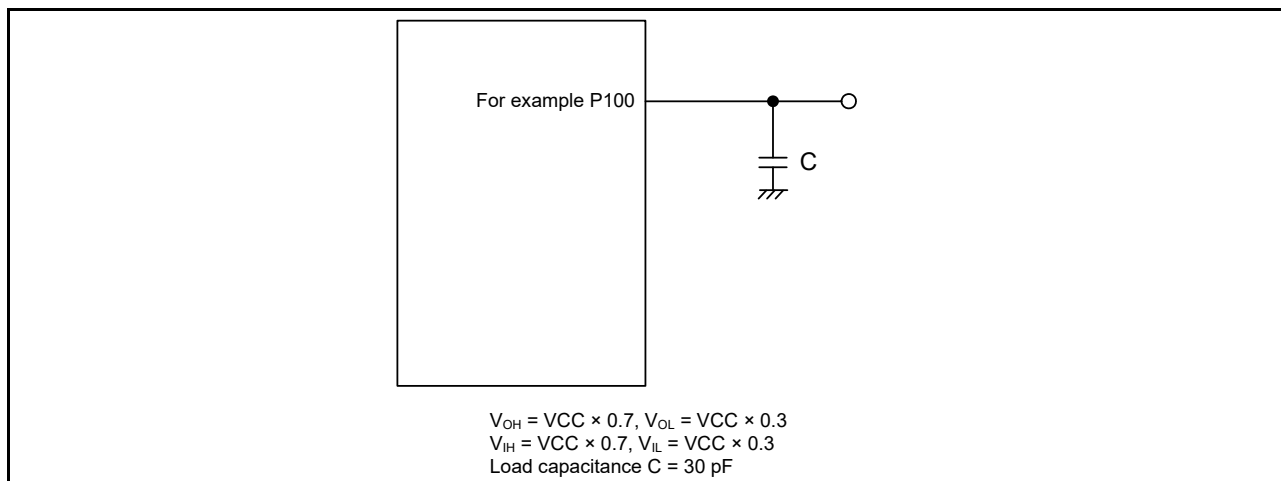


Figure 2.1 Input or output timing measurement conditions

The measurement conditions for the timing specification of each peripheral are recommended for the best peripheral operation. However, make sure to adjust the driving abilities of each pin to meet the conditions of your system.

Each function pin used for the same function must select the same drive ability. If the I/O drive ability of each function pin is mixed, the A/C specification of each function is not guaranteed.

2.1 Absolute Maximum Ratings

Table 2.1 Absolute maximum ratings

Parameter	Symbol	Value	Unit
Power supply voltage	VCC	-0.3 to +4.0	V
Input voltage (except for 5 V-tolerant ports*1)	V_{in}	-0.3 to VCC + 0.3	V
Input voltage (5 V-tolerant ports*1)	V_{in}	-0.3 to + VCC + 4.0 (max. 5.8)	V
Reference power supply voltage	VREFH/VREFH0	-0.3 to AVCC0 + 0.3	V
Analog power supply voltage	AVCC0 *2	-0.3 to +4.0	V
Analog input voltage (except for P000 to P007)	V_{AN}	-0.3 to AVCC0 + 0.3	V
Analog input voltage (P000 to P007) when PGA pseudo-differential input is disabled	V_{AN}	-0.3 to AVCC0 + 0.3	V
Analog input voltage (P000 to P002, P004 to P006) when PGA pseudo-differential input is enabled	V_{AN}	-1.3 to AVCC0 + 0.3	V
Analog input voltage (P003, P007) when PGA pseudo-differential input is enabled	V_{AN}	-0.8 to AVCC0 + 0.3	V
Operating temperature*3, *4	T_{opr}	-40 to +105	°C
Storage temperature	T_{stg}	-55 to +125	°C

Caution: Permanent damage to the MCU might result if absolute maximum ratings are exceeded.

Note 1. Ports P205, P206, P400, P401, P407 to P415, and P708 are 5 V tolerant.

Note 2. Connect AVCC0 to VCC.

Note 3. See section 2.2.1, T_j/T_a Definition.

Note 4. Contact Renesas Electronics sales office for information on derating operation when $T_a = +85^\circ\text{C}$ to $+105^\circ\text{C}$. Derating is the systematic reduction of load for improved reliability.

Table 2.2 Recommended operating conditions

Parameter	Symbol	Min	Typ	Max	Unit
Power supply voltages	VCC	2.7	-	3.6	V
	VSS	-	0	-	V
Analog power supply voltages	AVCC0*1	-	VCC	-	V
	AVSS0	-	0	-	V

Note 1. Connect AVCC0 to VCC. When the A/D converter, the D/A converter, or the comparator are not in use, do not leave the AVCC0, VREFH/VREFH0, AVSS0, and VREFL/VREFL0 pins open. Connect the AVCC0 and VREFH/VREFH0 pins to VCC, and the AVSS0 and VREFL/VREFL0 pins to VSS, respectively.

2.2 DC Characteristics

2.2.1 T_j/T_a Definition

Table 2.3 DC characteristics

Conditions: Products with operating temperature (T_a) -40 to $+105^\circ\text{C}$.

Parameter		Symbol	Typ	Max	Unit	Test conditions
Permissible junction temperature	100-pin LQFP	T_j	-	125	$^\circ\text{C}$	High-speed mode Low-speed mode Subosc-speed mode.
	64-pin LQFP					

Note: Make sure that $T_j = T_a + \theta_{ja} \times \text{total power consumption (W)}$,
where total power consumption = $(V_{CC} - V_{OH}) \times \Sigma I_{OH} + V_{OL} \times \Sigma I_{OL} + I_{CCmax} \times V_{CC}$.

2.2.2 I/O V_{IH} , V_{IL}

Table 2.4 I/O V_{IH} , V_{IL} (1 of 2)

Parameter			Symbol	Min	Typ	Max	Unit
Input voltage (except for Schmitt trigger input pins)	Peripheral function pin	EXTAL(external clock input), SPI (except RSPCK)	V_{IH}	$V_{CC} \times 0.8$	-	-	V
			V_{IL}	-	-	$V_{CC} \times 0.2$	
		IIC (SMBus)*1	V_{IH}	2.1	-	-	
			V_{IL}	-	-	0.8	
		IIC (SMBus)*2	V_{IH}	2.1	-	$V_{CC} + 3.6$ (max 5.8)	
			V_{IL}	-	-	0.8	
Schmitt trigger input voltage	IIC (except for SMBus)*1	V_{IH}	$V_{CC} \times 0.7$	-	-		
		V_{IL}	-	-	$V_{CC} \times 0.3$		
		ΔV_T	$V_{CC} \times 0.05$	-	-		
		IIC (except for SMBus)*2	V_{IH}	$V_{CC} \times 0.7$	-	$V_{CC} + 3.6$ (max 5.8)	
			V_{IL}	-	-	$V_{CC} \times 0.3$	
			ΔV_T	$V_{CC} \times 0.05$	-	-	
	5 V-tolerant ports*3, *7	V_{IH}	$V_{CC} \times 0.8$	-	$V_{CC} + 3.6$ (max 5.8)		
		V_{IL}	-	-	$V_{CC} \times 0.2$		
		ΔV_T	$V_{CC} \times 0.05$	-	-		

Table 2.4 I/O V_{IH} , V_{IL} (2 of 2)

Parameter			Symbol	Min	Typ	Max	Unit
Schmitt trigger input voltage	Peripheral function pin	P402/AGTIO0,1 P403/AGTIO0,1	V_{IH}	$VCC \times 0.8$	-	$VCC + 0.3$	V
			V_{IL}	-	-	$VCC \times 0.2$	
			ΔV_T	$VCC \times 0.05$	-	-	
		Other input pins*4	V_{IH}	$VCC \times 0.8$	-	-	
			V_{IL}	-	-	$VCC \times 0.2$	
			ΔV_T	$VCC \times 0.05$	-	-	
	Ports	5 V-tolerant ports*5, *7	V_{IH}	$VCC \times 0.8$	-	$VCC + 3.6$ (max 5.8)	
			V_{IL}	-	-	$VCC \times 0.2$	
		Other input pins*6	V_{IH}	$VCC \times 0.8$	-	-	
V_{IL}			-	-	$VCC \times 0.2$		

Note 1. SCL1_B, SDA1_B (total 2 pins).

Note 2. SCL0_A, SDA0_A, SCL0_B, SDA0_B, SCL1_A, SDA1_A (total 6 pins).

Note 3. RES and peripheral function pins associated with P205, P206, P400, P401, P407 to P415, P708 (total 15 pins).

Note 4. All input pins except for the peripheral function pins already described in the table.

Note 5. P205, P206, P400, P401, P407 to P415, P708 (total 14 pins).

Note 6. All input pins except for the ports already described in the table.

Note 7. When VCC is less than 2.7 V, the input voltage of 5 V-tolerant ports should be less than 3.6 V, otherwise breakdown may occur because 5 V-tolerant ports are electrically controlled so as not to violate the breakdown voltage.

2.2.3 I/O I_{OH} , I_{OL}

Table 2.5 I/O I_{OH} , I_{OL} (1 of 2)

Parameter			Symbol	Min	Typ	Max	Unit
Permissible output current (average value per pin)	Ports P008, P201	-	I_{OH}	-	-	-2.0	mA
			I_{OL}	-	-	2.0	mA
	Ports P014, P015	-	I_{OH}	-	-	-4.0	mA
			I_{OL}	-	-	4.0	mA
	Ports P205, P206, P407 to P415, P602, P708 (total 13 pins)	Low drive*1	I_{OH}	-	-	-2.0	mA
			I_{OL}	-	-	2.0	mA
		Middle drive*2	I_{OH}	-	-	-4.0	mA
			I_{OL}	-	-	4.0	mA
		High drive*3	I_{OH}	-	-	-20	mA
			I_{OL}	-	-	20	mA
	Other output pins*4	Low drive*1	I_{OH}	-	-	-2.0	mA
			I_{OL}	-	-	2.0	mA
		Middle drive*2	I_{OH}	-	-	-4.0	mA
			I_{OL}	-	-	4.0	mA
		High drive*3	I_{OH}	-	-	-16	mA
			I_{OL}	-	-	16	mA

Table 2.5 I/O I_{OH} , I_{OL} (2 of 2)

Parameter			Symbol	Min	Typ	Max	Unit
Permissible output current (max value per pin)	Ports P008, P201	-	I_{OH}	-	-	-4.0	mA
			I_{OL}	-	-	4.0	mA
	Ports P014, P015	-	I_{OH}	-	-	-8.0	mA
			I_{OL}	-	-	8.0	mA
	Ports P205, P206, P407 to P415, P602, P708 (total 13 pins)	Low drive*1	I_{OH}	-	-	-4.0	mA
			I_{OL}	-	-	4.0	mA
		Middle drive*2	I_{OH}	-	-	-8.0	mA
			I_{OL}	-	-	8.0	mA
		High drive*3	I_{OH}	-	-	-40	mA
			I_{OL}	-	-	40	mA
	Other output pins*4	Low drive*1	I_{OH}	-	-	-4.0	mA
			I_{OL}	-	-	4.0	mA
Middle drive*2		I_{OH}	-	-	-8.0	mA	
		I_{OL}	-	-	8.0	mA	
High drive*3		I_{OH}	-	-	-32	mA	
		I_{OL}	-	-	32	mA	
Permissible output current (max value of total of all pins)	Maximum of all output pins		$\Sigma I_{OH} (\text{max})$	-	-	-80	mA
			$\Sigma I_{OL} (\text{max})$	-	-	80	mA

Caution: To protect the reliability of the MCU, the output current values should not exceed the values in this table. The average output current indicates the average value of current measured during 100 μs .

- Note 1. This is the value when low driving ability is selected in the Port Drive Capability bit in the PmnPFS register. The selected driving ability is retained in Deep Software Standby mode.
- Note 2. This is the value when middle driving ability is selected in the Port Drive Capability bit in the PmnPFS register. The selected driving ability is retained in Deep Software Standby mode.
- Note 3. This is the value when high driving ability is selected in the Port Drive Capability bit in the PmnPFS register. The selected driving ability is retained in Deep Software Standby mode.
- Note 4. Except for P000 to P007, P200, which are input ports.

2.2.4 I/O V_{OH} , V_{OL} , and Other Characteristics

Table 2.6 I/O V_{OH} , V_{OL} , and other characteristics (1 of 2)

Parameter			Symbol	Min	Typ	Max	Unit	Test conditions
Output voltage	IIC		V_{OL}	-	-	0.4	V	$I_{OL} = 3.0 \text{ mA}$
			V_{OL}	-	-	0.6		$I_{OL} = 6.0 \text{ mA}$
	IIC*1		V_{OL}	-	-	0.4		$I_{OL} = 15.0 \text{ mA}$ (ICFER.FMPE = 1)
			V_{OL}	-	0.4	-		$I_{OL} = 20.0 \text{ mA}$ (ICFER.FMPE = 1)
	Ports P205, P206, P407 to P415, P602, P708 (total of 13 pins)*2		V_{OH}	VCC - 1.0	-	-		$I_{OH} = -20 \text{ mA}$ VCC = 3.3 V
			V_{OL}	-	-	1.0		$I_{OL} = 20 \text{ mA}$ VCC = 3.3 V
	Other output pins		V_{OH}	VCC - 0.5	-	-		$I_{OH} = -1.0 \text{ mA}$
			V_{OL}	-	-	0.5		$I_{OL} = 1.0 \text{ mA}$
Input leakage current	RES		$ I_{in} $	-	-	5.0	μA	$V_{in} = 0 \text{ V}$ $V_{in} = 5.5 \text{ V}$
	Ports P000 to P002, P004 to P006, P200			-	-	1.0		$V_{in} = 0 \text{ V}$ $V_{in} = V_{CC}$
	Ports P003, P007	Before initialization*3		-	-	45.0		$V_{in} = 0 \text{ V}$ $V_{in} = V_{CC}$
		After initialization*4		-	-	1.0		$V_{in} = 0 \text{ V}$ $V_{in} = V_{CC}$

Table 2.6 I/O V_{OH} , V_{OL} , and other characteristics (2 of 2)

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
Three-state leakage current (off state)	5 V-tolerant ports	I _{TSI}	-	-	5.0	μA $V_{in} = 0\text{ V}$ $V_{in} = 5.5\text{ V}$
			-	-	1.0	
Input pull-up MOS current	Ports P0 to P7 (except for ports P000 to P007)	I _p	-300	-	-10	μA $V_{CC} = 2.7\text{ to }3.6\text{ V}$ $V_{in} = 0\text{ V}$
Input capacitance	Ports P003, P007, P014, P015, P400, P401	C _{in}	-	-	16	pF $V_{bias} = 0\text{ V}$ $V_{amp} = 20\text{ mV}$ $f = 1\text{ MHz}$ $T_a = 25^\circ\text{C}$
	Other input pins		-	-	8	

Note 1. SCL0_A, SDA0_A (total 2 pins).

Note 2. This is the value when high driving ability is selected in the Port Drive Capability bit in the PmnPFS register. The selected driving ability is retained in Deep Software Standby mode.

Note 3. P0nPFS.ASEL(n = 3 or 7) = 1

Note 4. P0nPFS.ASEL(n = 3 or 7) = 0

2.2.5 Operating and Standby Current

Table 2.7 Operating and standby current (1 of 2)

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions							
Supply current*1	I _{CC} *3	-	-	87	mA	ICLK = 120 MHz PCLKA = 120 MHz PCLKB = 60 MHz PCLKC = 60 MHz PCLKD = 120 MHz FCLK = 60 MHz							
							Maximum*2						
							CoreMark®*5						
							Normal mode	All peripheral clocks enabled, while (1) code executing from flash*4					
								All peripheral clocks disabled, while (1) code executing from flash*5, *6					
							Sleep mode*5, *6						
							Increase during BGO operation	Data flash P/E					
								Code flash P/E					
							Low-speed mode*5	-	1.2	-	ICLK = 1 MHz		
							Subosc-speed mode*5	-	1.0	-	ICLK = 32.768 kHz		
							Software Standby mode	-	1.3	13	Ta ≤ 85°C		
								-	1.3	21	Ta ≤ 105°C		
							Deep Software Standby mode	DPSBYCR.DEEPCUT[1:0] = 00b*8	-	28	65	μA	Ta ≤ 85°C
									-	28	93	Ta ≤ 105°C	
DPSBYCR.DEEPCUT[1:0] = 01b*8	-	11.6	28	Ta ≤ 85°C									
	-	11.6	32	Ta ≤ 105°C									
DPSBYCR.DEEPCUT[1:0] = 11b*8	-	4.9	21	Ta ≤ 85°C									
	-	4.9	26	Ta ≤ 105°C									
Increase when the AGT is operating	When the low-speed on-chip oscillator (LOCO) is in use	-	4.4	-	-								
	When a crystal oscillator for low clock loads is in use	-	1.0	-	-								
When a crystal oscillator for standard clock loads is in use	-	1.4	-	-									
Analog power supply current	AI _{CC}	-	-	1.1	mA	-							
							During 12-bit A/D conversion						
							During 12-bit A/D conversion with S/H amp						
							PGA (1ch)						
							ACMPHS (1 unit)						
							Temperature sensor						
							During D/A conversion (per unit)	Without AMP output					
								With AMP output					
							Waiting for A/D, D/A conversion (all units)						
ADC12, DAC12 in standby modes (all units)*7	-	2	8	μA	-								

Table 2.7 Operating and standby current (2 of 2)

Parameter		Symbol	Min	Typ	Max	Unit	Test conditions	
Reference power supply current (VREFH0)	During 12-bit A/D conversion (unit 0)	AI _{REFH0}	-	70	120	μA	-	
	Waiting for 12-bit A/D conversion (unit 0)		-	0.07	0.5	μA	-	
	ADC12 in standby modes (unit 0)		-	0.07	0.5	μA	-	
Reference power supply current (VREFH)	During 12-bit A/D conversion (unit 1)	AI _{REFH}	-	70	120	μA	-	
	During D/A conversion (per unit)		Without AMP output	-	0.1	0.4	mA	-
			With AMP output	-	0.1	0.4	mA	-
	Waiting for 12-bit A/D (unit 1), D/A (all units) conversion		-	0.07	0.8	μA	-	
	ADC12 unit 1 in standby modes		-	0.07	0.8	μA	-	

- Note 1. Supply current values are with all output pins unloaded and all input pull-up MOS transistors in the off state.
- Note 2. Measured with clocks supplied to the peripheral functions. This does not include the BGO operation.
- Note 3. I_{CC} depends on f (ICLK) as follows. (ICLK:PCLKA:PCLKB:PCLKC:PCLKD = 2:2:1:1:2)
 I_{CC} Max. = 0.53 x f + 23 (maximum operation in High-speed mode)
 I_{CC} Typ. = 0.08 x f + 2.4 (normal operation in High-speed mode)
 I_{CC} Typ. = 0.1 x f + 1.1 (Low-speed mode)
 I_{CC} Max. = 0.09 x f + 23 (Sleep mode).
- Note 4. This does not include the BGO operation.
- Note 5. Supply of the clock signal to peripherals is stopped in this state. This does not include the BGO operation.
- Note 6. FCLK, PCLKA, PCLKB, PCLKC, and PCLKD are set to divided by 64 (3.75 MHz).
- Note 7. When the MCU is in Software Standby mode or the MSTPCRD.MSTPD16 (12-bit A/D Converter 0 Module Stop bit) and MSTPCRD.MSTPD15 (12-bit A/D Converter 1 Module Stop bit) are in the module-stop state. See section 35.6.8, Available functions and register settings of AN000 to AN002, AN007, AN100 to AN102, and AN107 in User's Manual.
- Note 8. For more information on the DBSBYCR register, see section 11.2.11, Deep Software Standby Control Register (DPSBYCR) in User's Manual.

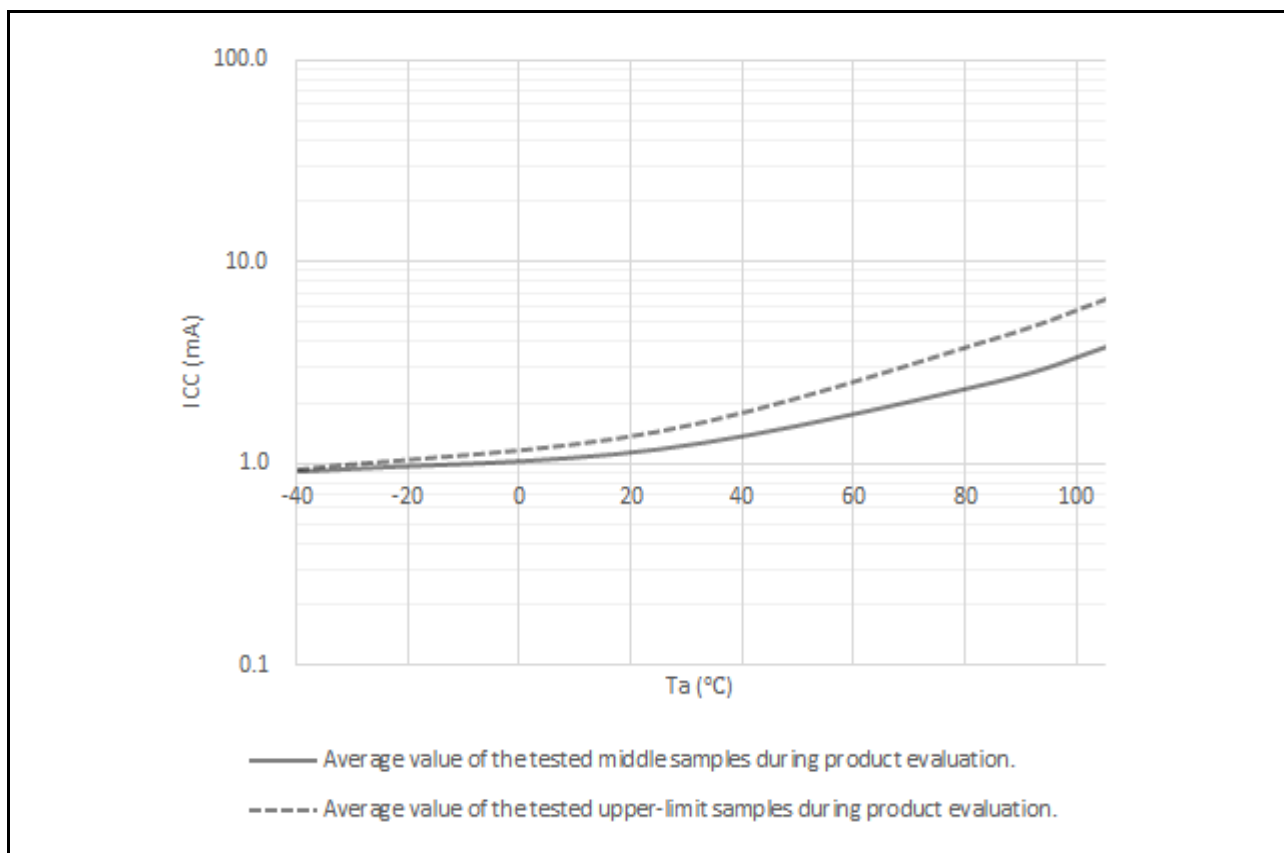


Figure 2.2 Temperature dependency in Software Standby mode (reference data)

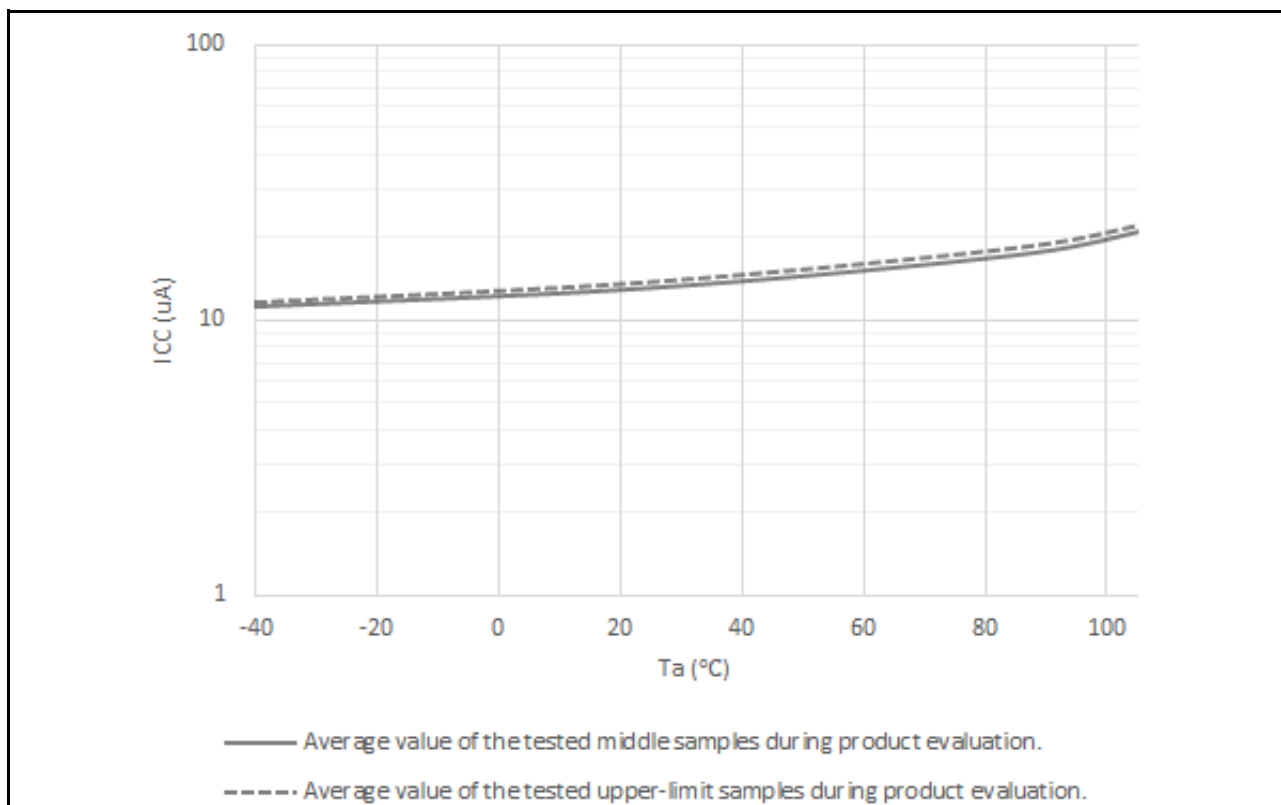


Figure 2.3 Temperature dependency in Deep Software Standby mode, power-on reset circuit low power function disabled (reference data)

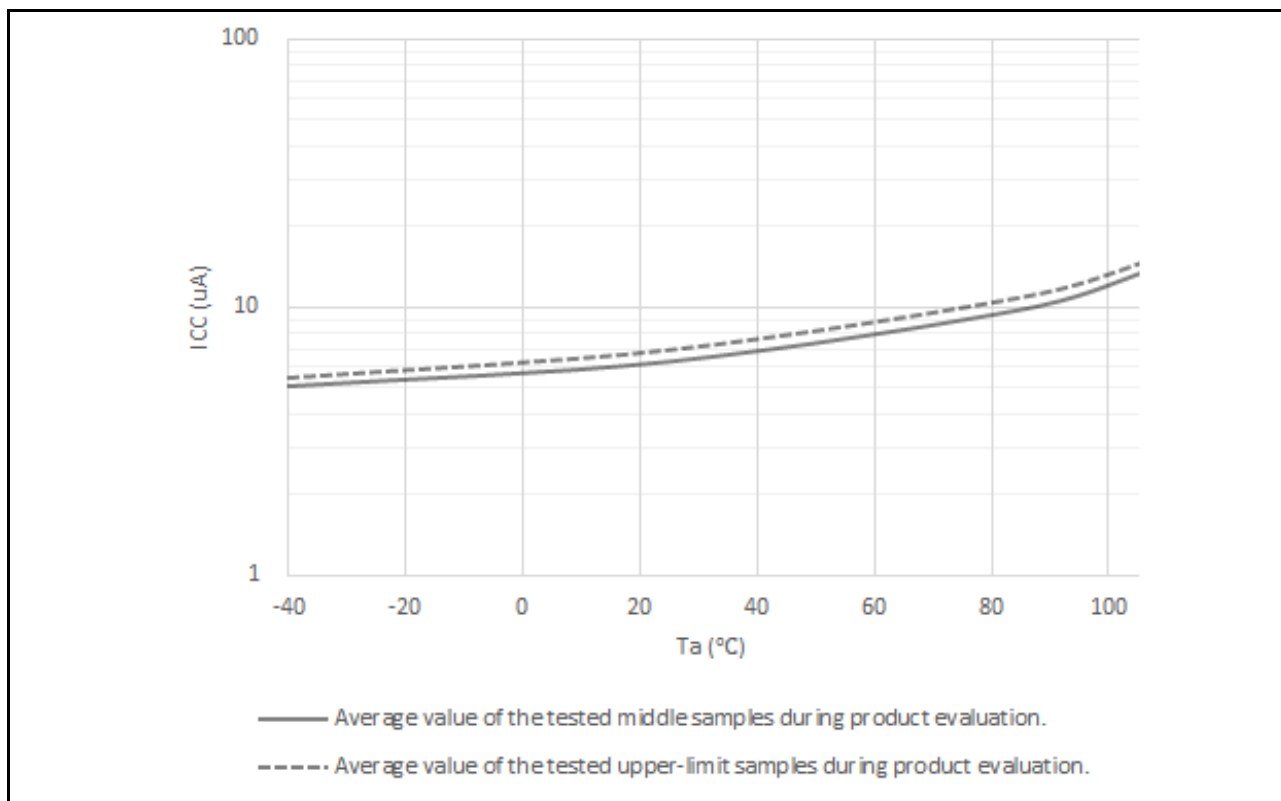


Figure 2.4 Temperature dependency in Deep Software Standby mode, power-on reset circuit low power function enabled (reference data)

2.2.6 VCC Rise and Fall Gradient and Ripple Frequency

Table 2.8 Rising gradient characteristics

Parameter		Symbol	Min	Typ	Max	Unit	Test conditions
VCC rising gradient	Voltage monitor 0 reset disabled at startup	SrVCC	0.0084	-	20	ms/V	-
	Voltage monitor 0 reset enabled at startup		0.0084	-	-		-
	SCI boot mode*1		0.0084	-	20		-

Note 1. At boot mode, the reset from voltage monitor 0 is disabled regardless of the value of the OFS1.LVDAS bit.

Table 2.9 Rise and fall gradient and ripple frequency characteristics

The ripple voltage must meet the allowable ripple frequency $f_{r(VCC)}$ within the range between the VCC upper limit (3.6 V) and lower limit (2.7 V). When the VCC change exceeds $VCC \pm 10\%$, the allowable voltage change rising and falling gradient $dt/dVCC$ must be met.

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
Allowable ripple frequency	$f_{r(VCC)}$	-	-	10	kHz	Figure 2.5 $V_{r(VCC)} \leq VCC \times 0.2$
		-	-	1	MHz	Figure 2.5 $V_{r(VCC)} \leq VCC \times 0.08$
		-	-	10	MHz	Figure 2.5 $V_{r(VCC)} \leq VCC \times 0.06$
Allowable voltage change rising and falling gradient	$dt/dVCC$	1.0	-	-	ms/V	When VCC change exceeds $VCC \pm 10\%$

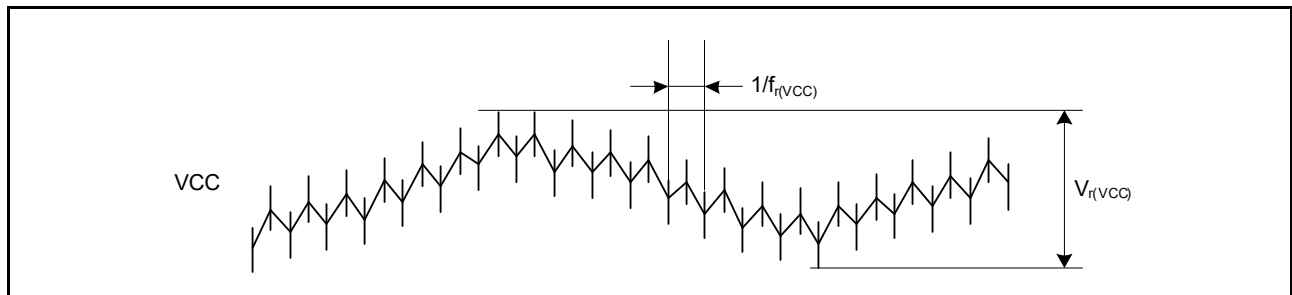


Figure 2.5 Ripple waveform

2.3 AC Characteristics

2.3.1 Frequency

Table 2.10 Operation frequency value in high-speed mode

Parameter		Symbol	Min	Typ	Max	Unit
Operation frequency	System clock (ICLK)*2	f	-	-	120	MHz
	Peripheral module clock (PCLKA)*2		-	-	120	
	Peripheral module clock (PCLKB)*2		-	-	60	
	Peripheral module clock (PCLKC)*2		-*3	-	60	
	Peripheral module clock (PCLKD)*2		-	-	120	
	Flash interface clock (FCLK)*2	-*1	-	60		

Note 1. FCLK must run at a frequency of at least 4 MHz when programming or erasing the flash memory.

Note 2. See section 9, Clock Generation Circuit in User's Manual for the relationship between the ICLK, PCLKA, PCLKB, PCLKC, PCLKD, and FCLK frequencies.

Note 3. When the ADC12 is used, the PCLKC frequency must be at least 1 MHz.

Table 2.11 Operation frequency value in low-speed mode

Parameter	Symbol	Min	Typ	Max	Unit	
Operation frequency	System clock (ICLK)*2	f	-	-	1	MHz
	Peripheral module clock (PCLKA)*2	-	-	1		
	Peripheral module clock (PCLKB)*2	-	-	1		
	Peripheral module clock (PCLKC)*2,*3	_*3	-	1		
	Peripheral module clock (PCLKD)*2	-	-	1		
	Flash interface clock (FCLK)*1,*2	-	-	1		

Note 1. Programming or erasing the flash memory is disabled in Low-speed mode.

Note 2. See section 9, Clock Generation Circuit in User's Manual for the relationship between the ICLK, PCLKA, PCLKB, PCLKC, PCLKD, and FCLK frequencies.

Note 3. When the ADC12 is used, the PCLKC frequency must be set to at least 1 MHz.

Table 2.12 Operation frequency value in Subosc-speed mode

Parameter	Symbol	Min	Typ	Max	Unit	
Operation frequency	System clock (ICLK)*2	f	29.4	-	36.1	kHz
	Peripheral module clock (PCLKA)*2	-	-	36.1		
	Peripheral module clock (PCLKB)*2	-	-	36.1		
	Peripheral module clock (PCLKC)*2,*3	-	-	36.1		
	Peripheral module clock (PCLKD)*2	-	-	36.1		
	Flash interface clock (FCLK)*1,*2	29.4	-	36.1		

Note 1. Programming or erasing the flash memory is disabled in Subosc-speed mode.

Note 2. See section 9, Clock Generation Circuit in User's Manual for the relationship between the ICLK, PCLKA, PCLKB, PCLKC, PCLKD, and FCLK frequencies.

Note 3. The ADC12 cannot be used.

2.3.2 Clock Timing

Table 2.13 Clock timing except for sub-clock oscillator (1 of 2)

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
EXTAL external clock input cycle time	t_{EXcyc}	41.66	-	-	ns	Figure 2.6
EXTAL external clock input high pulse width	t_{EXH}	15.83	-	-	ns	
EXTAL external clock input low pulse width	t_{EXL}	15.83	-	-	ns	
EXTAL external clock rise time	t_{EXr}	-	-	5.0	ns	
EXTAL external clock fall time	t_{EXf}	-	-	5.0	ns	
Main clock oscillator frequency	f_{MAIN}	8	-	24	MHz	-
Main clock oscillation stabilization wait time (crystal) *1	$t_{MAINOSCWT}$	-	-	_*1	ms	Figure 2.7
LOCO clock oscillation frequency	f_{LOCO}	29.4912	32.768	36.0448	kHz	-
LOCO clock oscillation stabilization wait time	t_{LOCOWT}	-	-	60.4	μ s	Figure 2.8
ILOCO clock oscillation frequency	f_{ILOCO}	13.5	15	16.5	kHz	-
MOCO clock oscillation frequency	F_{MOCO}	6.8	8	9.2	MHz	-
MOCO clock oscillation stabilization wait time	t_{MOCOWT}	-	-	15.0	μ s	-

Table 2.13 Clock timing except for sub-clock oscillator (2 of 2)

Parameter		Symbol	Min	Typ	Max	Unit	Test conditions
HOCO clock oscillator oscillation frequency	Without FLL	f _{HOCO16}	15.78	16	16.22	MHz	-20 ≤ Ta ≤ 105°C
		f _{HOCO18}	17.75	18	18.25		
		f _{HOCO20}	19.72	20	20.28		
	With FLL	f _{HOCO16}	15.71	16	16.29		-40 ≤ Ta ≤ -20°C
		f _{HOCO18}	17.68	18	18.32		
		f _{HOCO20}	19.64	20	20.36		
		f _{HOCO16}	15.960	16	16.040	-40 ≤ Ta ≤ 105°C Sub-clock frequency accuracy is ±50 ppm.	
		f _{HOCO18}	17.955	18	18.045		
		f _{HOCO20}	19.950	20	20.050		
HOCO clock oscillation stabilization wait time*2		t _{HOCOWT}	-	-	64.7	μs	-
FLL stabilization wait time		t _{FLLWT}	-	-	1.8	ms	-
PLL clock frequency		f _{PLL}	120	-	240	MHz	-
PLL clock oscillation stabilization wait time		t _{PLLWT}	-	-	174.9	μs	Figure 2.9

- Note 1. When setting up the main clock oscillator, ask the oscillator manufacturer for an oscillation evaluation, and use the results as the recommended oscillation stabilization time. Set the MOSCWTCR register to a value equal to or greater than the recommended value.
After changing the setting in the MOSCCR.MOSTP bit to start main clock operation, read the OSCSF.MOSCSF flag to confirm that it is 1, and then start using the main clock oscillator.
- Note 2. This is the time from release from reset state until the HOCO oscillation frequency (fHOCO) reaches the range for guaranteed operation.

Table 2.14 Clock timing for the sub-clock oscillator

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
Sub-clock frequency	f _{SUB}	-	32.768	-	kHz	-
Sub-clock oscillation stabilization wait time	t _{SUBOSCWT}	-	-	-*1	s	Figure 2.10

- Note 1. When setting up the sub-clock oscillator, ask the oscillator manufacturer for an oscillation evaluation and use the results as the recommended oscillation stabilization time.
After changing the setting in the SOSCCR.SOSTP bit to start sub-clock operation, only start using the sub-clock oscillator after the sub-clock oscillation stabilization time elapses with an adequate margin. A value that is two times the value shown is recommended.

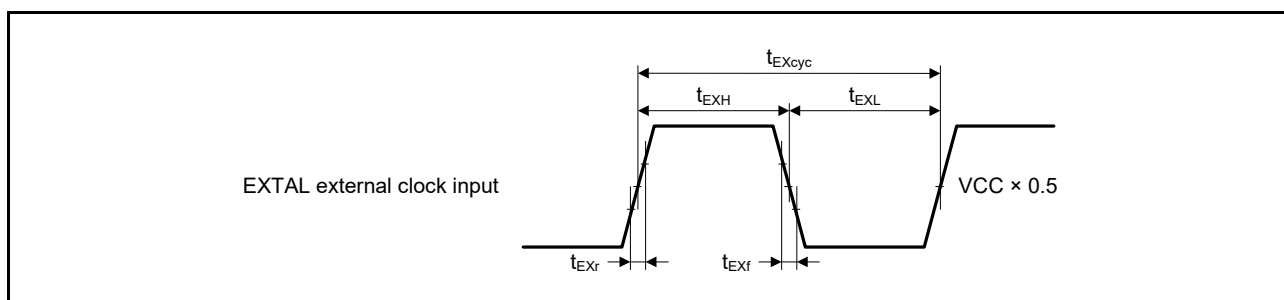


Figure 2.6 EXTAL external clock input timing

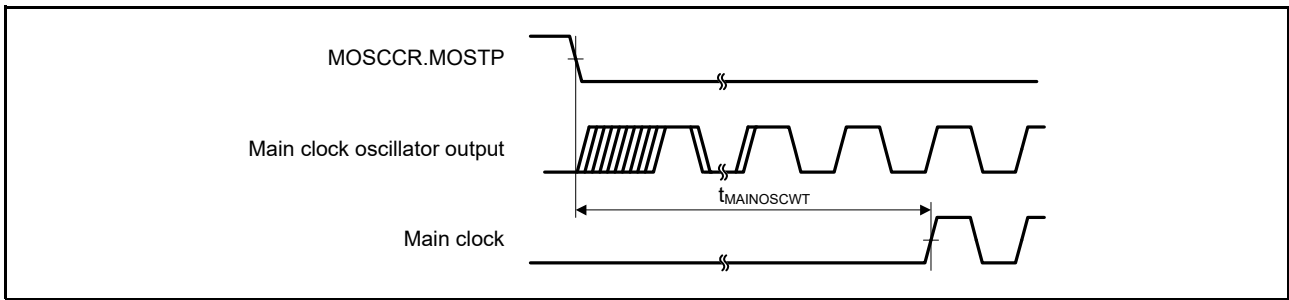


Figure 2.7 Main clock oscillation start timing

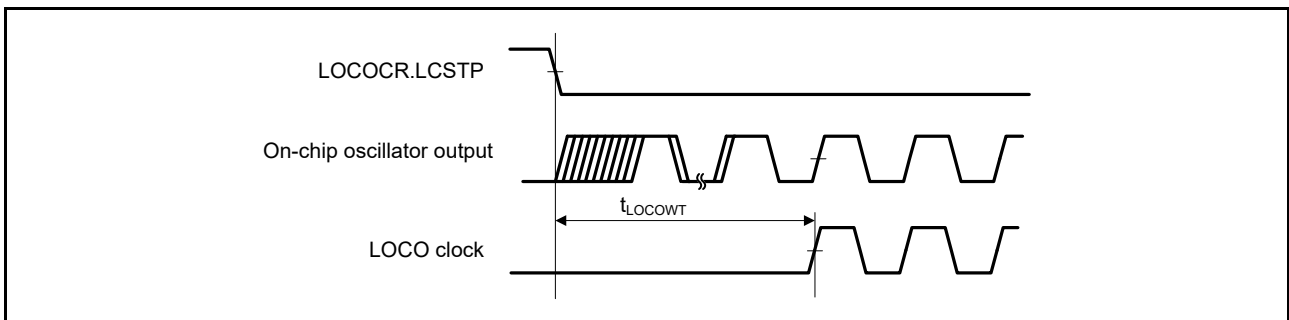


Figure 2.8 LOCO clock oscillation start timing

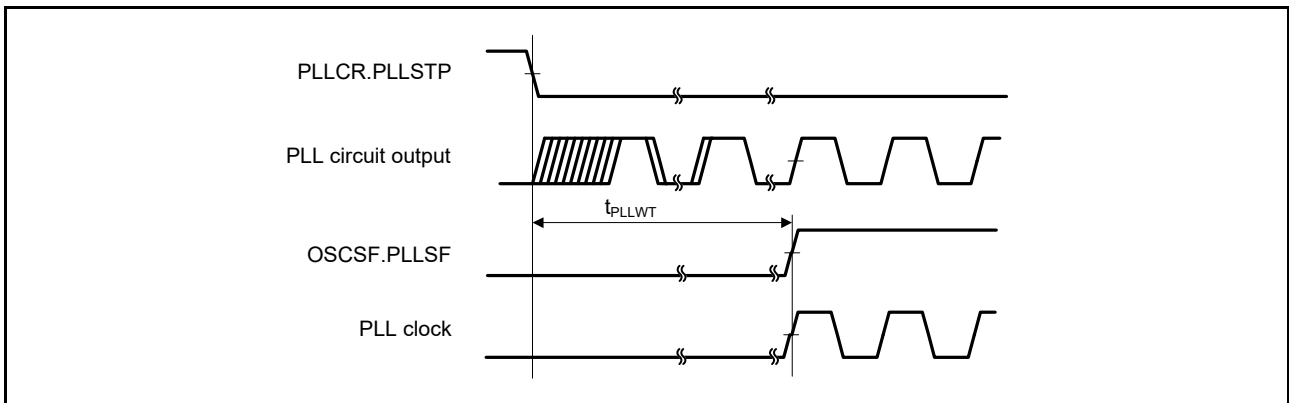


Figure 2.9 PLL clock oscillation start timing

Note: Only operate the PLL after the main clock oscillation has stabilized.

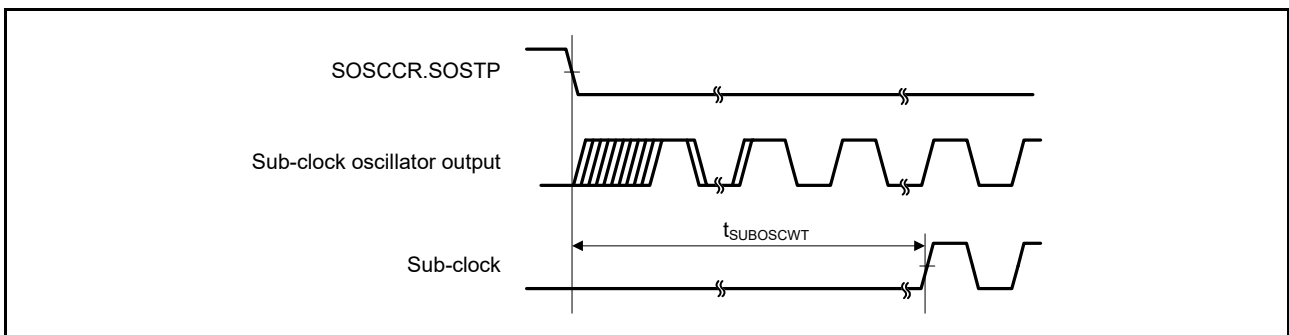


Figure 2.10 Sub-clock oscillation start timing

2.3.3 Reset Timing

Table 2.15 Reset timing

Parameter		Symbol	Min	Typ	Max	Unit	Test conditions
RES pulse width	Power-on	t_{RESWP}	1	-	-	ms	Figure 2.11
	Deep Software Standby mode	t_{RESWD}	0.6	-	-	ms	Figure 2.12
	Software Standby mode, Subosc-speed mode	t_{RESWS}	0.3	-	-	ms	
	All other	t_{RESW}	200	-	-	μ s	
Wait time after RES cancellation		t_{RESWT}	-	29	32	μ s	Figure 2.11
Wait time after internal reset cancellation (IWDT reset, WDT reset, software reset, SRAM parity error reset, bus master MPU error reset, bus slave MPU error reset, stack pointer error reset)		t_{RESW2}	-	320	390	μ s	-

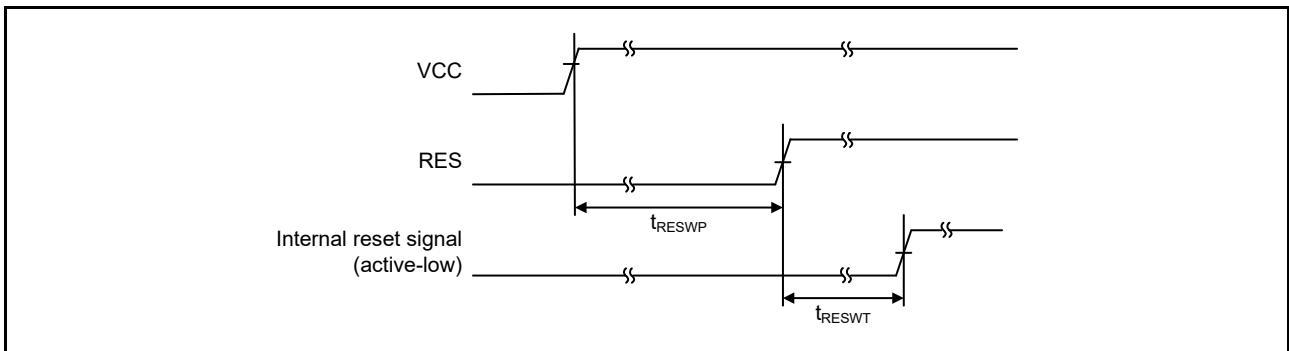


Figure 2.11 Power-on reset timing

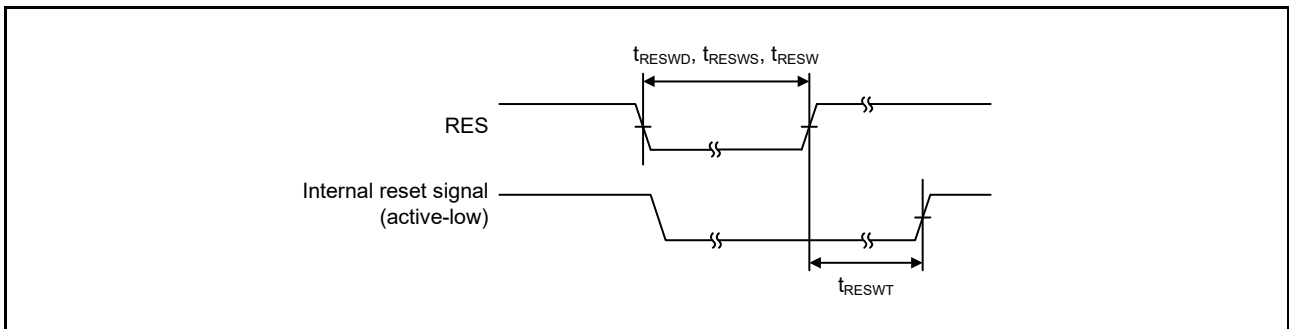


Figure 2.12 Reset input timing

2.3.4 Wakeup Timing

Table 2.16 Timing of recovery from low power modes

Parameter			Symbol	Min	Typ	Max	Unit	Test conditions
Recovery time from Software Standby mode*1	Crystal resonator connected to main clock oscillator	System clock source is main clock oscillator*2	t_{SBYMC}	-	2.4*9	2.8*9	ms	Figure 2.13 The division ratio of all oscillators is 1.
		System clock source is PLL with main clock oscillator*3	t_{SBYPC}	-	2.7*9	3.2*9	ms	
	External clock input to main clock oscillator	System clock source is main clock oscillator*4	t_{SBYEX}	-	230*9	280*9	μ s	
		System clock source is PLL with main clock oscillator*5	t_{SBYPE}	-	570*9	700*9	μ s	
	System clock source is sub-clock oscillator*8		t_{SBYSC}	-	1.2*9	1.3*9	ms	
	System clock source is LOCO*8		t_{SBYLO}	-	1.2*9	1.4*9	ms	
	System clock source is HOCO*6		t_{SBYHO}	-	240*9, *10	300*9, *10	μ s	
	System clock source is MOCO*7		t_{SBYMO}	-	220*9	300*9	μ s	
Recovery time from Deep Software Standby mode			t_{DSBY}	-	0.65	1.0	ms	Figure 2.14
Wait time after cancellation of Deep Software Standby mode			t_{DSBYWT}	34	-	35	t_{cyc}	
Recovery time from Software Standby mode to Snooze mode	High-speed mode when system clock source is HOCO (20 MHz)		t_{SNZ}	-	35*9, *10	70*9, *10	μ s	Figure 2.15
	High-speed mode when system clock source is MOCO (8 MHz)		t_{SNZ}	-	11*9	14*9	μ s	

- Note 1. The recovery time is determined by the system clock source. When multiple oscillators are active, the recovery time can be determined with the following equation:
Total recovery time = recovery time for an oscillator as the system clock source + the longest oscillation stabilization time of any oscillators requiring longer stabilization times than the system clock source + 2 LOCO cycles (when LOCO is operating) + 3 SOSC cycles (when Subosc is oscillating and MSTPC0 = 0 (CAC module stop)).
- Note 2. When the frequency of the crystal is 24 MHz (Main Clock Oscillator Wait Control Register (MOSCWTCR) is set to 05h). For other settings (MOSCWTCR is set to Xh), the recovery time can be determined with the following equation:
 $t_{SBYMC} (\text{MOSCWTCR} = Xh) = t_{SBYMC} (\text{MOSCWTCR} = 05h) + (t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = Xh) - t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = 05h))$
- Note 3. When the frequency of PLL is 240 MHz (Main Clock Oscillator Wait Control Register (MOSCWTCR) is set to 05h). For other settings (MOSCWTCR is set to Xh), the recovery time can be determined with the following equation:
 $t_{SBYMC} (\text{MOSCWTCR} = Xh) = t_{SBYMC} (\text{MOSCWTCR} = 05h) + (t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = Xh) - t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = 05h))$
- Note 4. When the frequency of the external clock is 24 MHz (Main Clock Oscillator Wait Control Register (MOSCWTCR) is set to 01h). For other settings (MOSCWTCR is set to Xh), the recovery time can be determined with the following equation:
 $t_{SBYMC} (\text{MOSCWTCR} = Xh) = t_{SBYMC} (\text{MOSCWTCR} = 01h) + (t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = Xh) - t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = 01h))$
- Note 5. When the frequency of PLL is 240 MHz (Main Clock Oscillator Wait Control Register (MOSCWTCR) is set to 01h). For other settings (MOSCWTCR is set to Xh), the recovery time can be determined with the following equation:
 $t_{SBYMC} (\text{MOSCWTCR} = Xh) = t_{SBYMC} (\text{MOSCWTCR} = 01h) + (t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = Xh) - t_{\text{MAINOSCWT}} (\text{MOSCWTCR} = 01h))$
- Note 6. The HOCO frequency is 20 MHz.
- Note 7. The MOCO frequency is 8 MHz.
- Note 8. In Subosc-speed mode, the sub-clock oscillator or LOCO continues oscillating in Software Standby mode.
- Note 9. When the SNZCR.RXDREQEN bit is set to 0, the following time is added as the power supply recovery time:
STCONR.STCON[1:0] = 00b:16 μ s (typical), 34 μ s (maximum)
STCONR.STCON[1:0] = 11b:16 μ s (typical), 104 μ s (maximum).
- Note 10. When the SNZCR.RXDREQEN bit is set to 0, 16 μ s (typical) or 18 μ s (maximum) is added as the HOCO wait time.

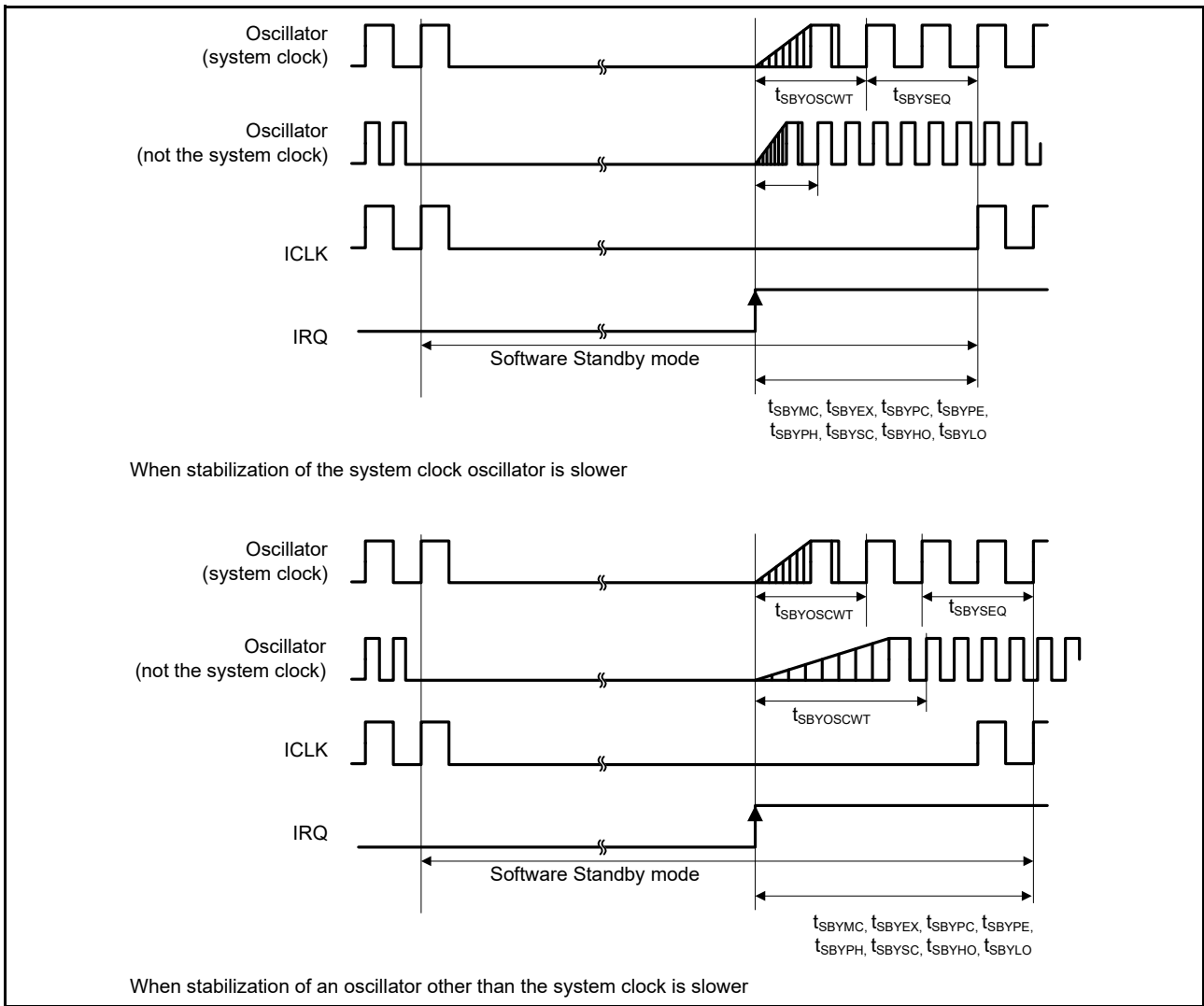


Figure 2.13 Software Standby mode cancellation timing

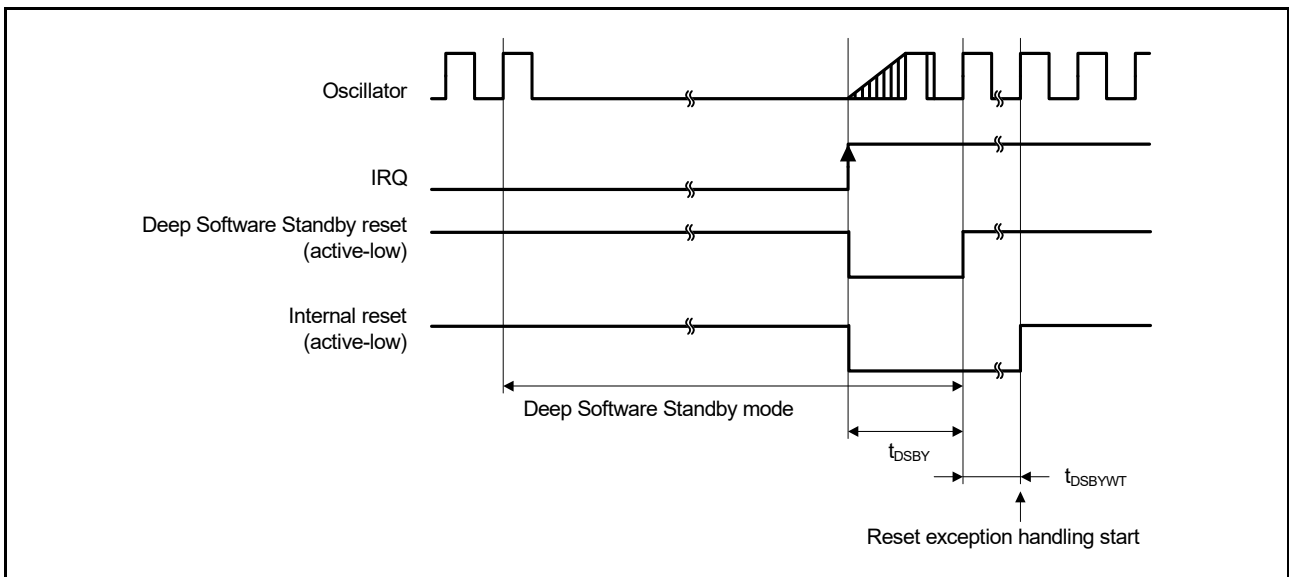


Figure 2.14 Deep Software Standby mode cancellation timing

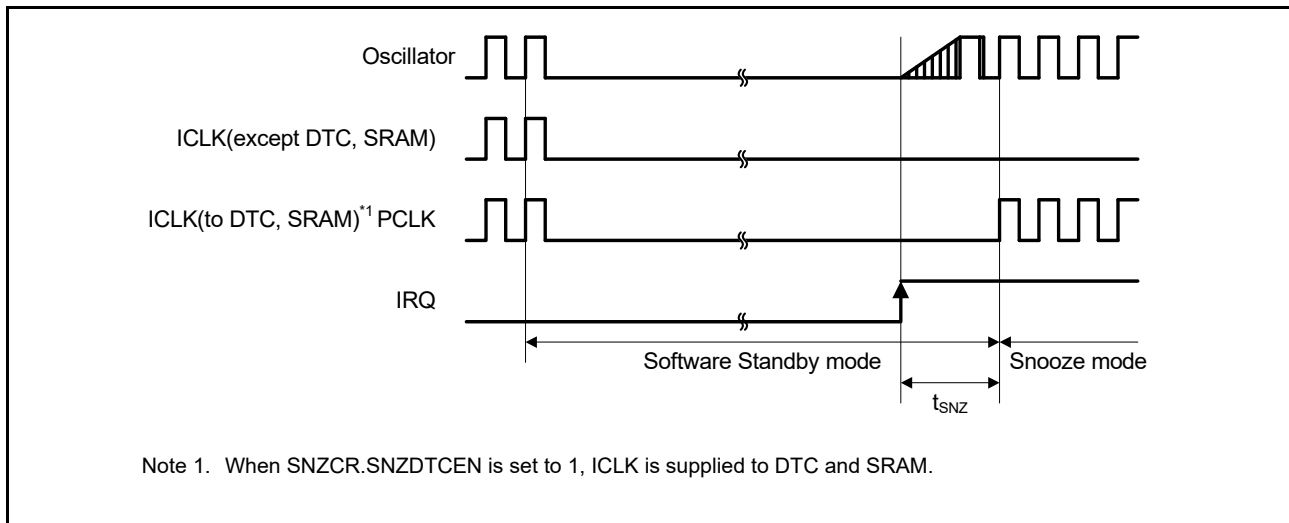


Figure 2.15 Recovery timing from Software Standby mode to Snooze mode

2.3.5 NMI and IRQ Noise Filter

Table 2.17 NMI and IRQ noise filter

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions	
NMI pulse width	t_{NMIW}	200	-	-	ns	NMI digital filter disabled	
		$t_{Pcyc} \times 2^{*1}$	-	-			$t_{Pcyc} \times 2 \leq 200$ ns
		200	-	-		NMI digital filter enabled	$t_{NMICK} \times 3 \leq 200$ ns
		$t_{NMICK} \times 3.5^{*2}$	-	-			$t_{NMICK} \times 3 > 200$ ns
IRQ pulse width	t_{IRQW}	200	-	-	ns	IRQ digital filter disabled	
		$t_{Pcyc} \times 2^{*1}$	-	-			$t_{Pcyc} \times 2 \leq 200$ ns
		200	-	-		IRQ digital filter enabled	$t_{IRQCK} \times 3 \leq 200$ ns
		$t_{IRQCK} \times 3.5^{*3}$	-	-			$t_{IRQCK} \times 3 > 200$ ns

- Note: 200 ns minimum in Software Standby mode.
- Note: If the clock source is switched, add 4 clock cycles of the switched source.
- Note 1. t_{Pcyc} indicates the PCLKB cycle.
- Note 2. t_{NMICK} indicates the cycle of the NMI digital filter sampling clock.
- Note 3. t_{IRQCK} indicates the cycle of the IRQi digital filter sampling clock.

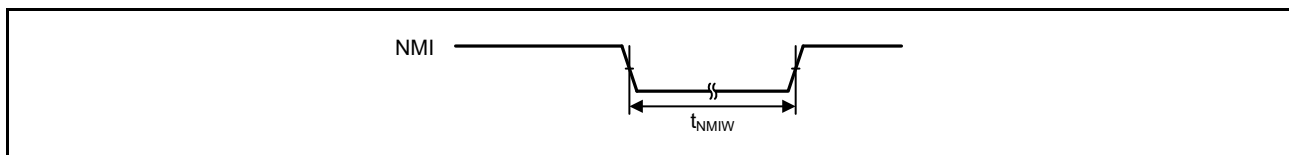


Figure 2.16 NMI interrupt input timing

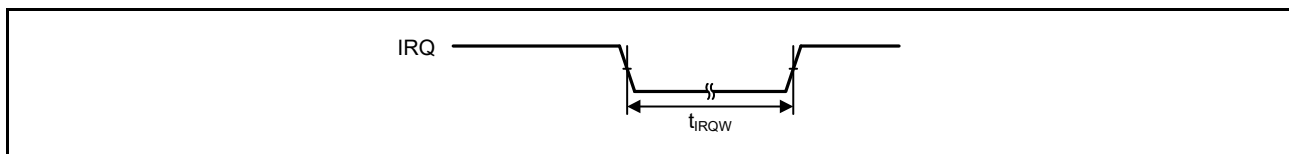


Figure 2.17 IRQ interrupt input timing

2.3.6 I/O Ports, POEG, GPT32, AGT, KINT, and ADC12 Trigger Timing

Table 2.18 I/O ports, POEG, GPT32, AGT, KINT, and ADC12 trigger timing

GPT32 conditions: High drive output is selected in the Port Drive Capability bit in the PmnPFS register.
 AGT conditions: Middle drive output is selected in the Port Drive Capability bit in the PmnPFS register.

Parameter		Symbol	Min	Max	Unit	Test conditions	
I/O ports	Input data pulse width	t_{PRW}	1.5	-	t_{Pcyc}	Figure 2.18	
POEG	POEG input trigger pulse width	t_{POEW}	3	-	t_{Pcyc}	Figure 2.19	
GPT32	Input capture pulse width	Single edge	t_{GTICW}	1.5	-	t_{PDcyc}	Figure 2.20
		Dual edge		2.5	-		
	GTIOCxY output skew (x = 0 to 7, Y = A or B)	Middle drive buffer	t_{GTISK}^{*1}	-	4	ns	Figure 2.21
		High drive buffer		-	4		
	GTIOCxY output skew (x = 8 to 12, Y = A or B)	Middle drive buffer		-	4		
		High drive buffer		-	4		
GTIOCxY output skew (x = 0 to 12, Y = A or B)	Middle drive buffer	-		6			
	High drive buffer	-		6			
OPS output skew GTOUUP, GTOULO, GTOVUP, GTOVLO, GTOWUP, GTOWLO		t_{GTOSK}	-	5	ns	Figure 2.22	
GPT (PWM Delay Generation Circuit)	GTIOCxY_Z output skew (x = 0 to 3, Y = A or B, Z = A)	t_{HRSK}^{*2}	-	2.0	ns	Figure 2.23	
AGT	AGTIO, AGTEE input cycle	t_{ACYC}^{*3}	100	-	ns	Figure 2.24	
	AGTIO, AGTEE input high width, low width	t_{ACKWH} , t_{ACKWL}	40	-	ns		
	AGTIO, AGTO, AGTOA, AGTOB output cycle	t_{ACYC2}	62.5	-	ns		
ADC12	ADC12 trigger input pulse width	t_{TRGW}	1.5	-	t_{Pcyc}	Figure 2.25	
KINT	KRn(n = 00 to 07) pulse width	t_{KR}	250	-	ns	Figure 2.26	

Note: t_{Pcyc} : PCLKB cycle, t_{PDcyc} : PCLKD cycle.

Note 1. This skew applies when the same driver I/O is used. If the I/O of the middle and high drivers is mixed, operation is not guaranteed.

Note 2. The load is 30 pF.

Note 3. Constraints on input cycle:

When not switching the source clock: $t_{Pcyc} \times 2 < t_{ACYC}$ should be satisfied.

When switching the source clock: $t_{Pcyc} \times 6 < t_{ACYC}$ should be satisfied.

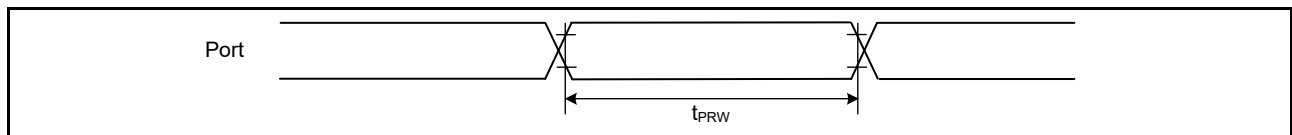


Figure 2.18 I/O ports input timing

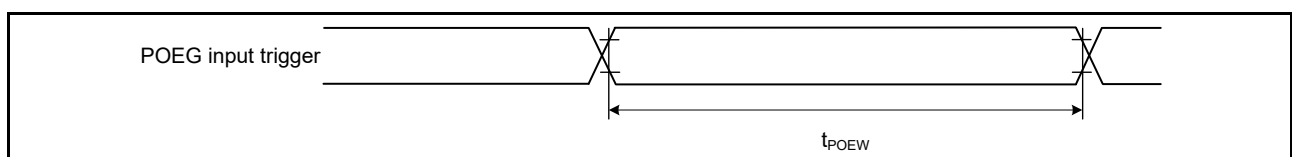


Figure 2.19 POEG input trigger timing

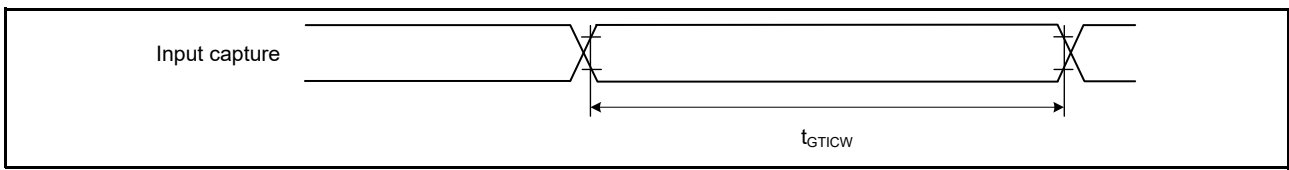


Figure 2.20 GPT32 input capture timing

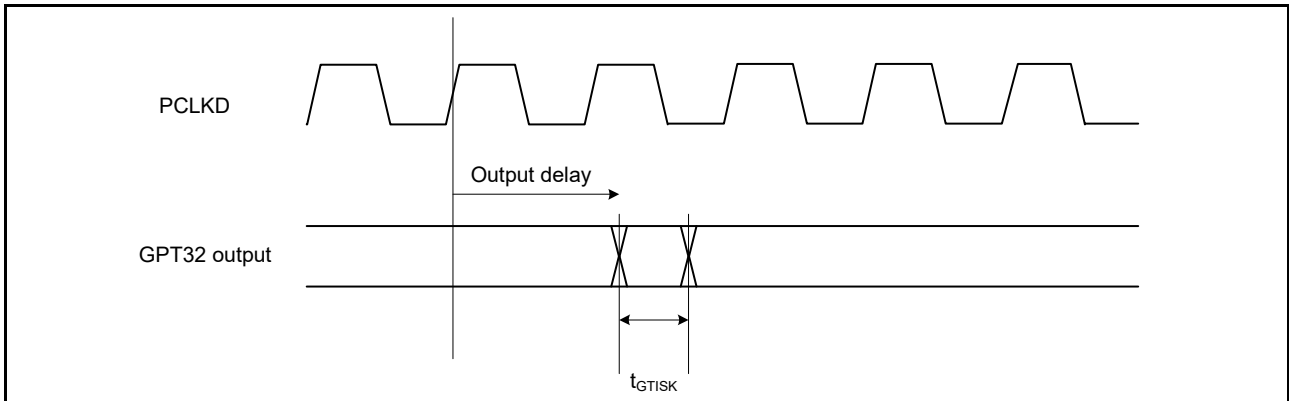


Figure 2.21 GPT32 output delay skew

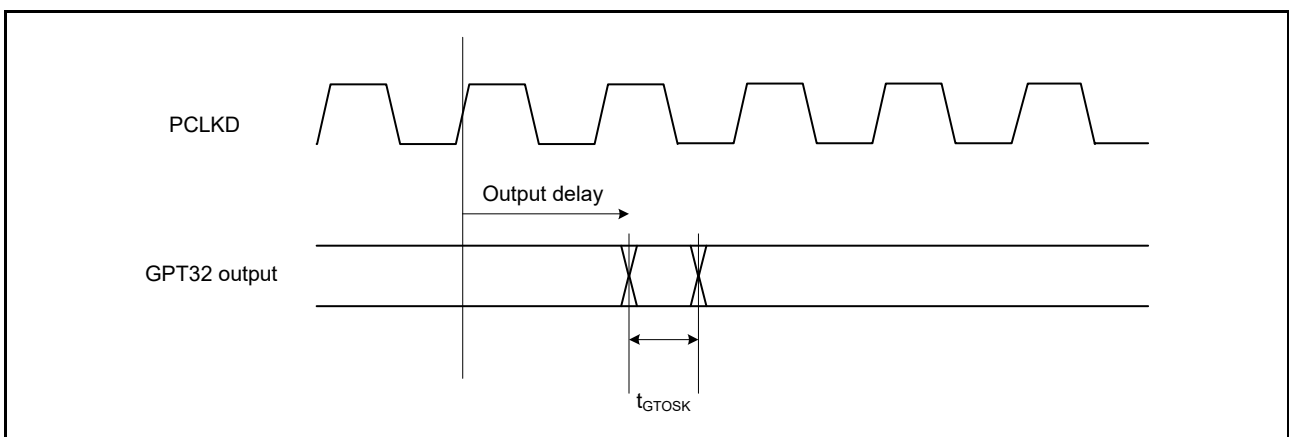


Figure 2.22 GPT32 output delay skew for OPS

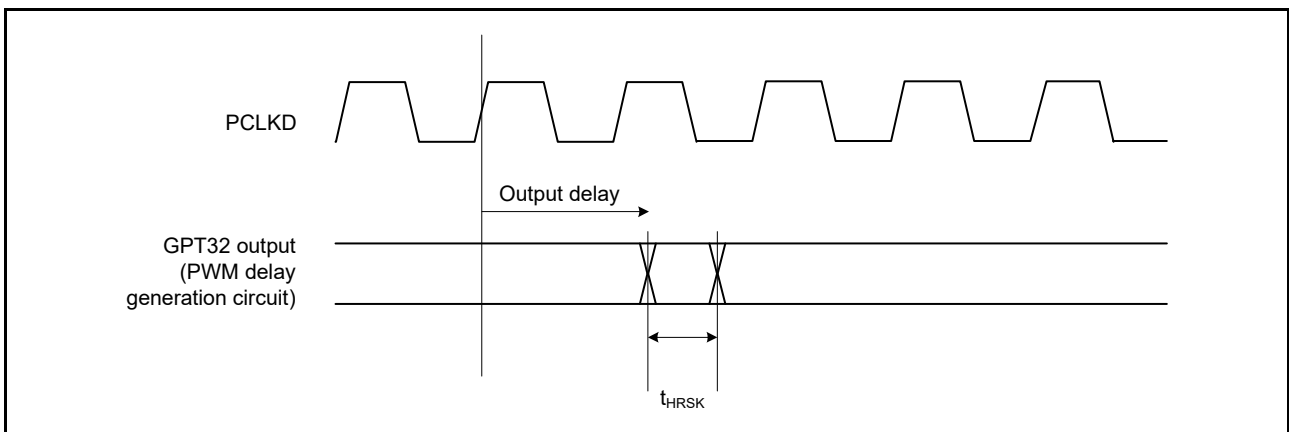


Figure 2.23 GPT32 (PWM delay generation circuit) output delay skew

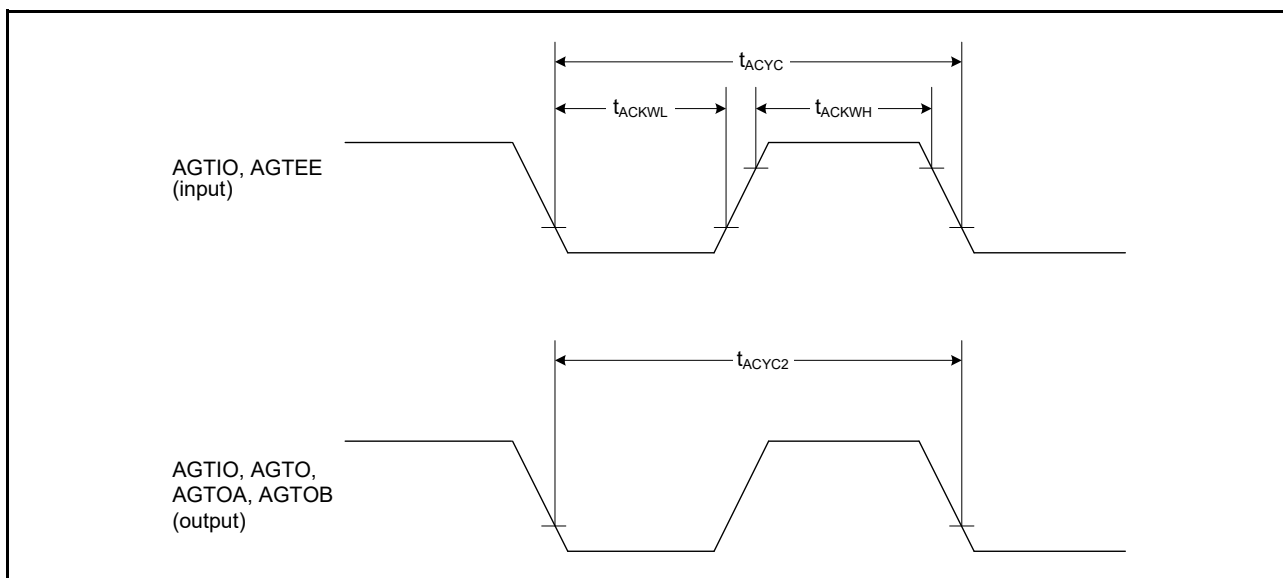


Figure 2.24 AGT input/output timing

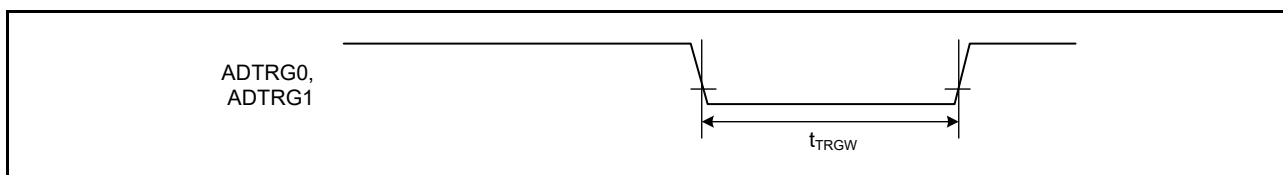


Figure 2.25 ADC12 trigger input timing

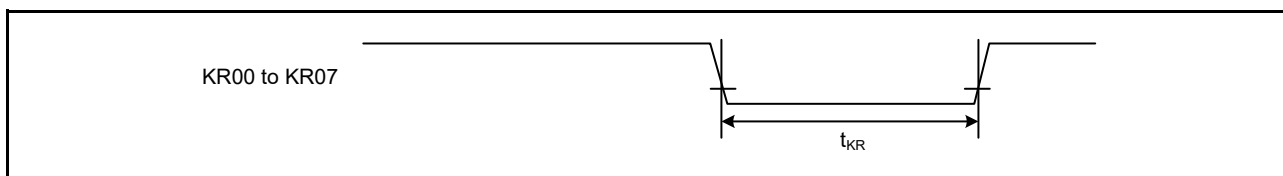


Figure 2.26 Key interrupt input timing

2.3.7 PWM Delay Generation Circuit Timing

Table 2.19 PWM Delay Generation Circuit timing

Parameter	Min	Typ	Max	Unit	Test conditions
Operation frequency	80	-	120	MHz	-
Resolution	-	260	-	ps	PCLKD = 120 MHz
DNL*1	-	±2.0	-	LSB	-

Note 1. This value normalizes the differences between lines in 1-LSB resolution.

2.3.8 CAC Timing

Table 2.20 CAC timing

Parameter		Symbol	Min	Typ	Max	Unit	Test conditions
CAC	CACREF input pulse width	t_{CACREF}	$t_{PBcyc} \leq t_{cac}^{*2}$	$4.5 \times t_{cac} + 3 \times t_{PBcyc}$	-	-	ns
			$t_{PBcyc} > t_{cac}^{*2}$	$5 \times t_{cac} + 6.5 \times t_{PBcyc}$	-	-	ns

Note 1. t_{pBcyc} : PCLKB cycle.

Note 2. t_{cac} : CAC count clock source cycle.

2.3.9 SCI Timing

Table 2.21 SCI timing (1)

Conditions: High drive output is selected in the Port Drive Capability bit in the PmnPFS register for the following pins: SCK0 to SCK4, SCK8, SCK9.

For other pins, middle drive output is selected in the Port Drive Capability bit in the PmnPFS register.

Parameter		Symbol	Min	Max	Unit*1	Test conditions	
SCI	Input clock cycle	Asynchronous	t_{Scyc}	4	-	t_{Pcyc}	Figure 2.27
		Clock synchronous		6	-		
Input clock pulse width		t_{SCKW}	0.4	0.6	t_{Scyc}		
Input clock rise time		t_{SCKr}	-	5	ns		
Input clock fall time		t_{SCKf}	-	5	ns		
Output clock cycle	Asynchronous	t_{Scyc}	6	-	t_{Pcyc}		
		Clock synchronous		4		-	
Output clock pulse width		t_{SCKW}	0.4	0.6	t_{Scyc}		
Output clock rise time		t_{SCKr}	-	5	ns		
Output clock fall time		t_{SCKf}	-	5	ns		
Transmit data delay	Clock synchronous	t_{TXD}	-	25	ns	Figure 2.28	
Receive data setup time	Clock synchronous	t_{RXS}	15	-	ns		
Receive data hold time	Clock synchronous	t_{RXH}	5	-	ns		

Note 1. t_{pcyc} : PCLKA cycle.

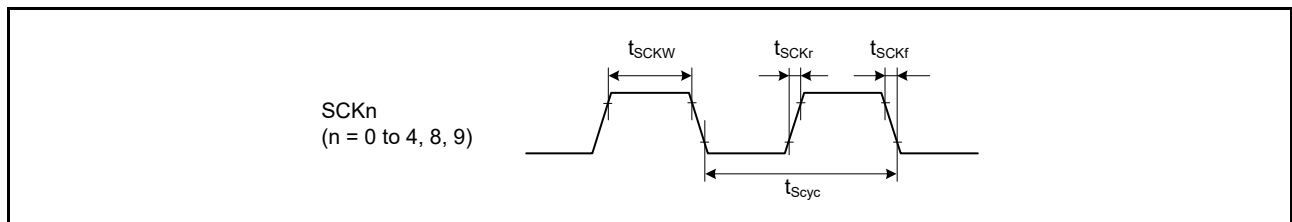


Figure 2.27 SCK clock input/output timing

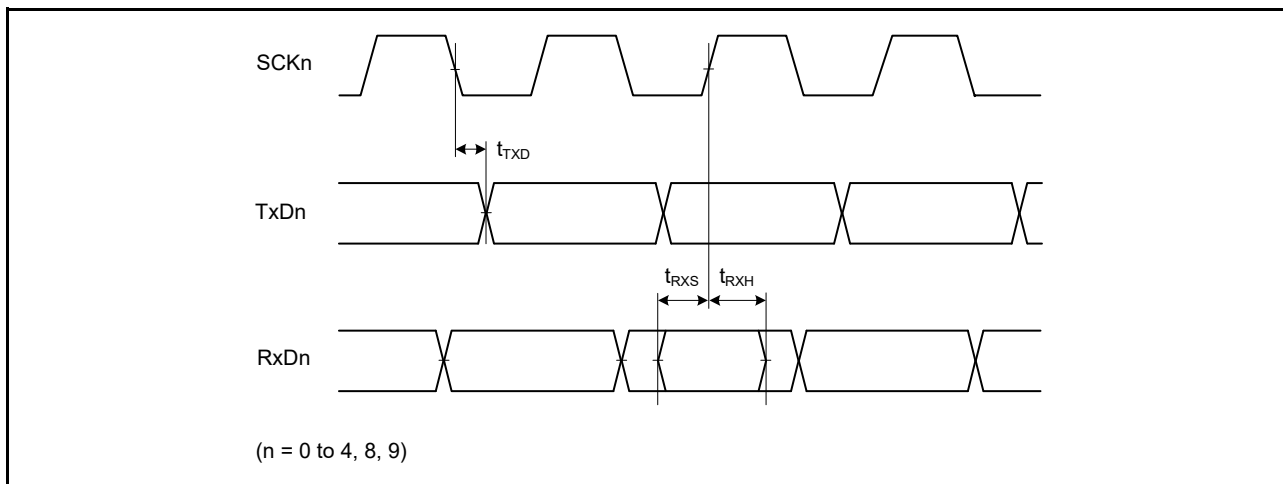


Figure 2.28 SCI input/output timing in clock synchronous mode

Table 2.22 SCI timing (2)

Conditions: High drive output is selected in the Port Drive Capability bit in the PmnPFS register for the following pins: SCK0 to SCK4, SCK8, SCK9.

For other pins, middle drive output is selected in the Port Drive Capability bit in the PmnPFS register.

Parameter		Symbol	Min	Max	Unit	Test conditions
Simple SPI	SCK clock cycle output (master)	t_{SPCyc}	4 (PCLKA ≤ 60 MHz) 8 (PCLKA > 60 MHz)	65536	t_{PCyc}	Figure 2.29 Figure 2.30 to Figure 2.33 Figure 2.33
	SCK clock cycle input (slave)	-	6 (PCLKA ≤ 60 MHz) 12 (PCLKA > 60 MHz)	65536		
	SCK clock high pulse width	t_{SPCKWH}	0.4	0.6	t_{SPCyc}	
	SCK clock low pulse width	t_{SPCKWL}	0.4	0.6	t_{SPCyc}	
	SCK clock rise and fall time	t_{SPCKr} , t_{SPCKf}	-	20	ns	
	Data input setup time	t_{SU}	33.3	-	ns	
	Data input hold time	t_{H}	33.3	-	ns	
	SS input setup time	t_{LEAD}	1	-	t_{SPCyc}	
	SS input hold time	t_{LAG}	1	-	t_{SPCyc}	
	Data output delay	t_{OD}	-	33.3	ns	
	Data output hold time	t_{OH}	-10	-	ns	
	Data rise and fall time	t_{Dr} , t_{Df}	-	16.6	ns	
	SS input rise and fall time	t_{SSLr} , t_{SSLf}	-	16.6	ns	
	Slave access time	t_{SA}	-	4 (PCLKA ≤ 60 MHz) 8 (PCLKA > 60 MHz)	t_{PCyc}	
	Slave output release time	t_{REL}	-	5 (PCLKA ≤ 60 MHz) 10 (PCLKA > 60 MHz)	t_{PCyc}	

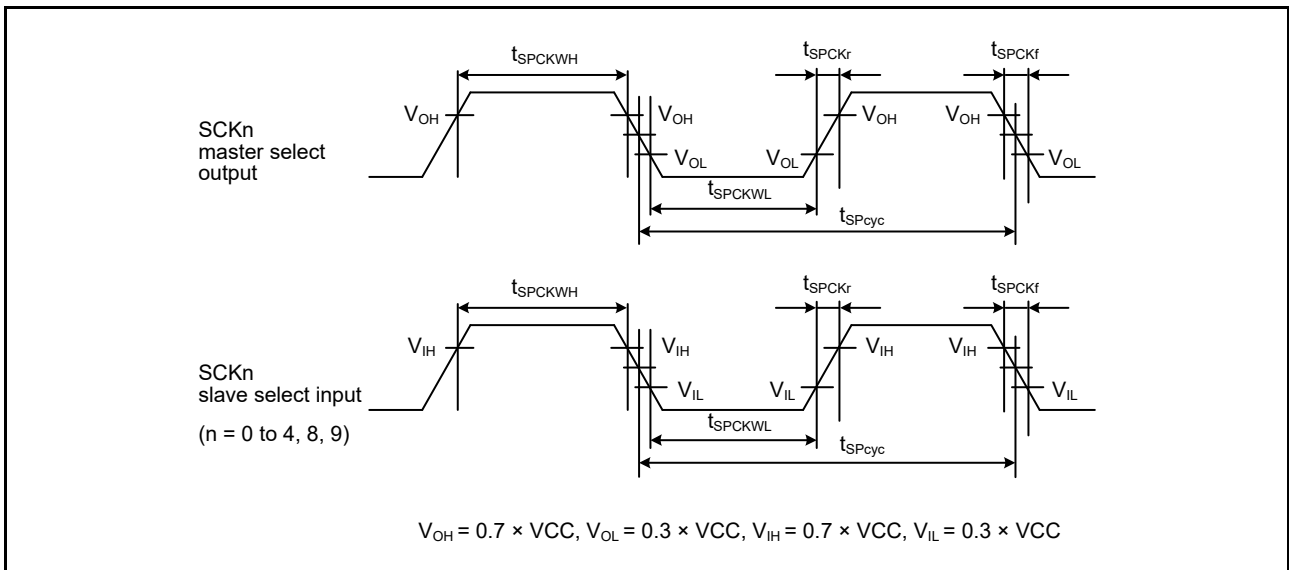


Figure 2.29 SCI simple SPI mode clock timing

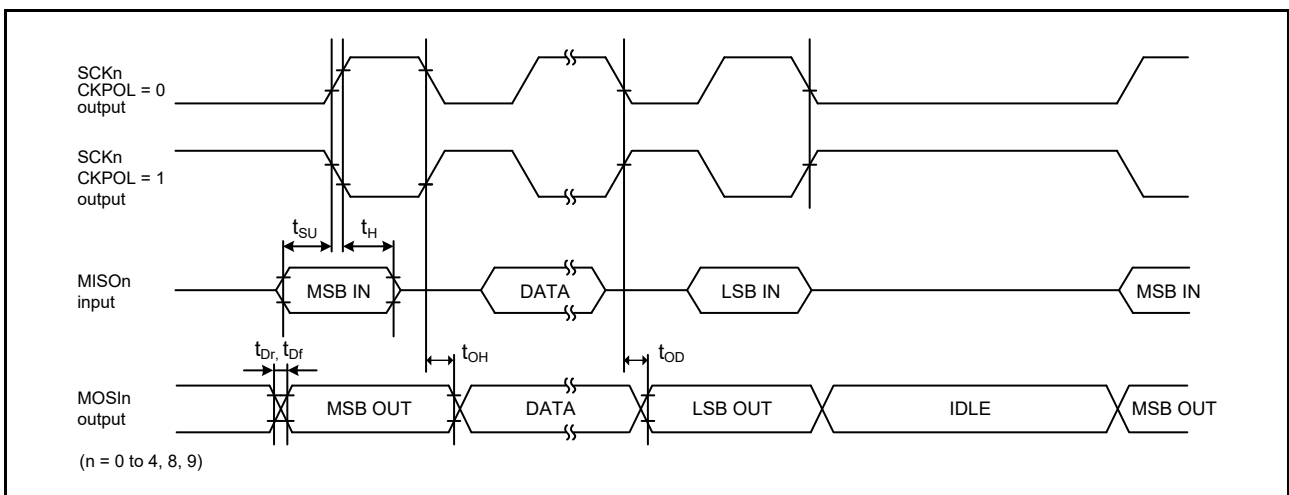


Figure 2.30 SCI simple SPI mode timing for master when CKPH = 1

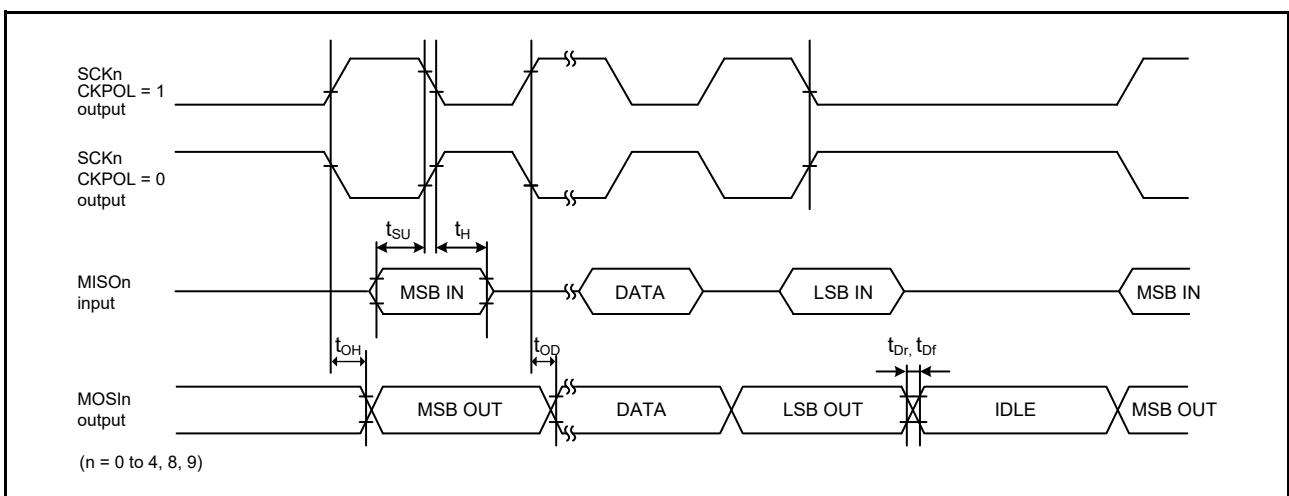


Figure 2.31 SCI simple SPI mode timing for master when CKPH = 0

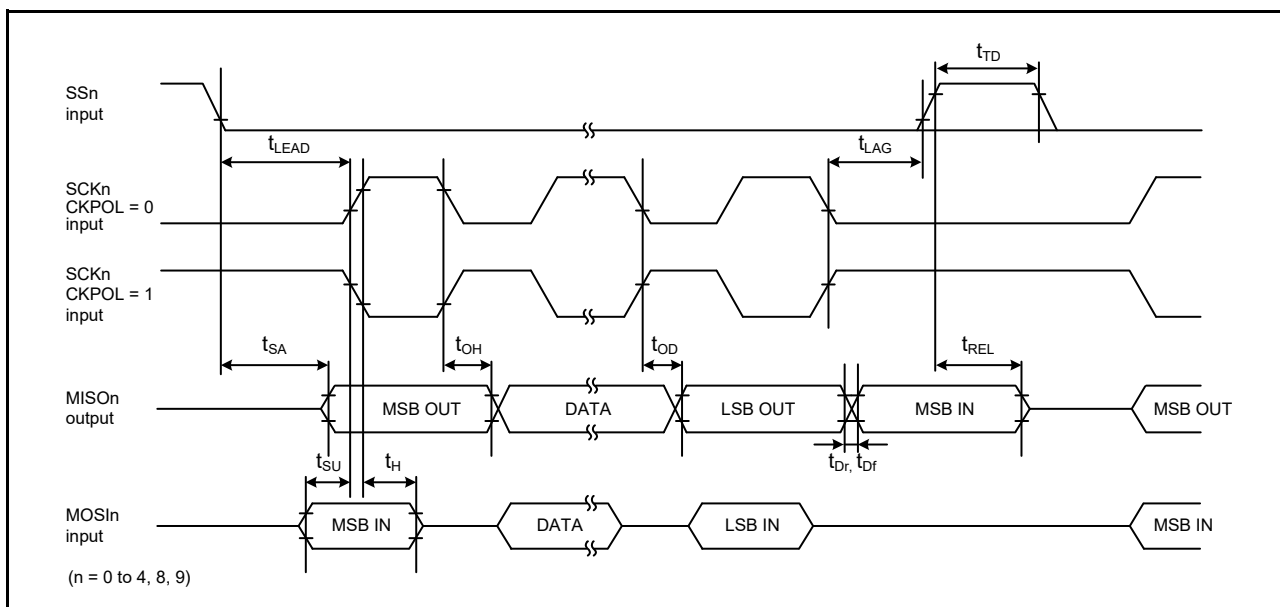


Figure 2.32 SCI simple SPI mode timing for slave when CKPH = 1

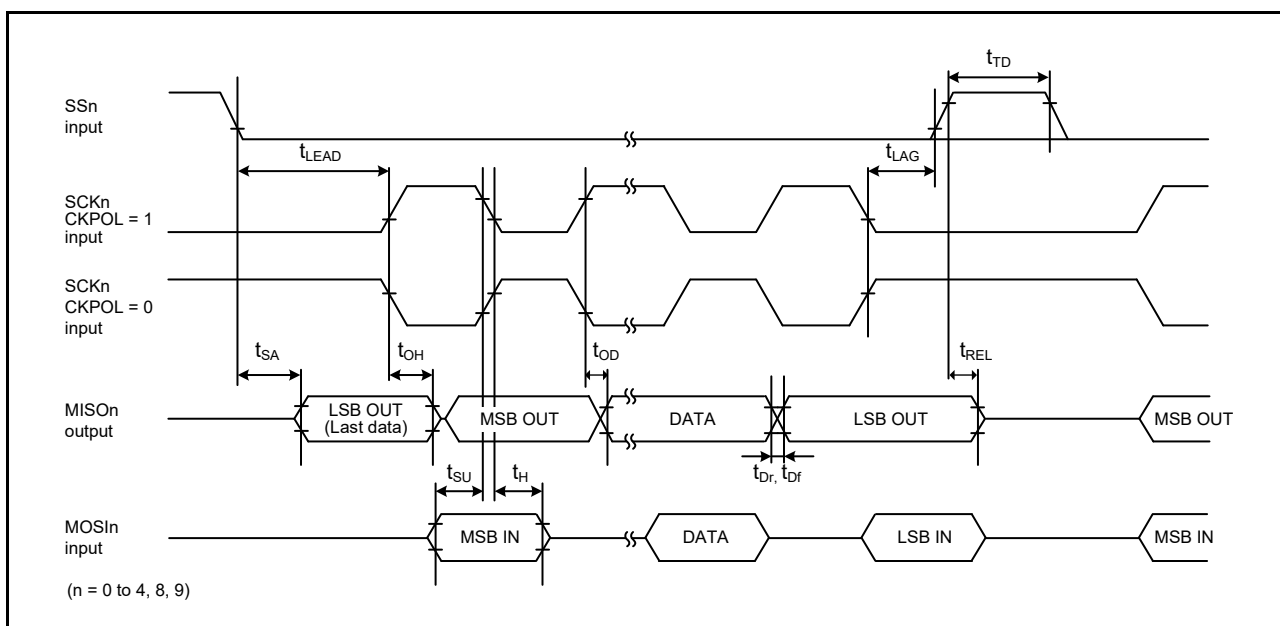


Figure 2.33 SCI simple SPI mode timing for slave when CKPH = 0

Table 2.23 SCI timing (3) (1 of 2)

Conditions: Middle drive output is selected in the Port Drive Capability bit in the PmnPFS register.

Parameter	Symbol	Min	Max	Unit	Test conditions	
Simple IIC (Standard mode)	SDA input rise time	t_{Sr}	-	1000	ns	Figure 2.34
	SDA input fall time	t_{Sf}	-	300	ns	
	SDA input spike pulse removal time	t_{SP}	0	$4 \times t_{IIcCyc}$	ns	
	Data input setup time	t_{SDAS}	250	-	ns	
	Data input hold time	t_{SDAH}	0	-	ns	
	SCL, SDA capacitive load	C_b^{*1}	-	400	pF	

Table 2.23 SCI timing (3) (2 of 2)

Conditions: Middle drive output is selected in the Port Drive Capability bit in the PmnPFS register.

Parameter	Symbol	Min	Max	Unit	Test conditions	
Simple IIC (Fast mode)	SDA input rise time	t_{Sr}	-	300	ns	Figure 2.34
	SDA input fall time	t_{Sf}	-	300	ns	
	SDA input spike pulse removal time	t_{SP}	0	$4 \times t_{IICcyc}$	ns	
	Data input setup time	t_{SDAS}	100	-	ns	
	Data input hold time	t_{SDAH}	0	-	ns	
	SCL, SDA capacitive load	C_b^{*1}	-	400	pF	

Note: t_{IICcyc} : IIC internal reference clock (IIC ϕ) cycle.

Note 1. C_b indicates the total capacity of the bus line.

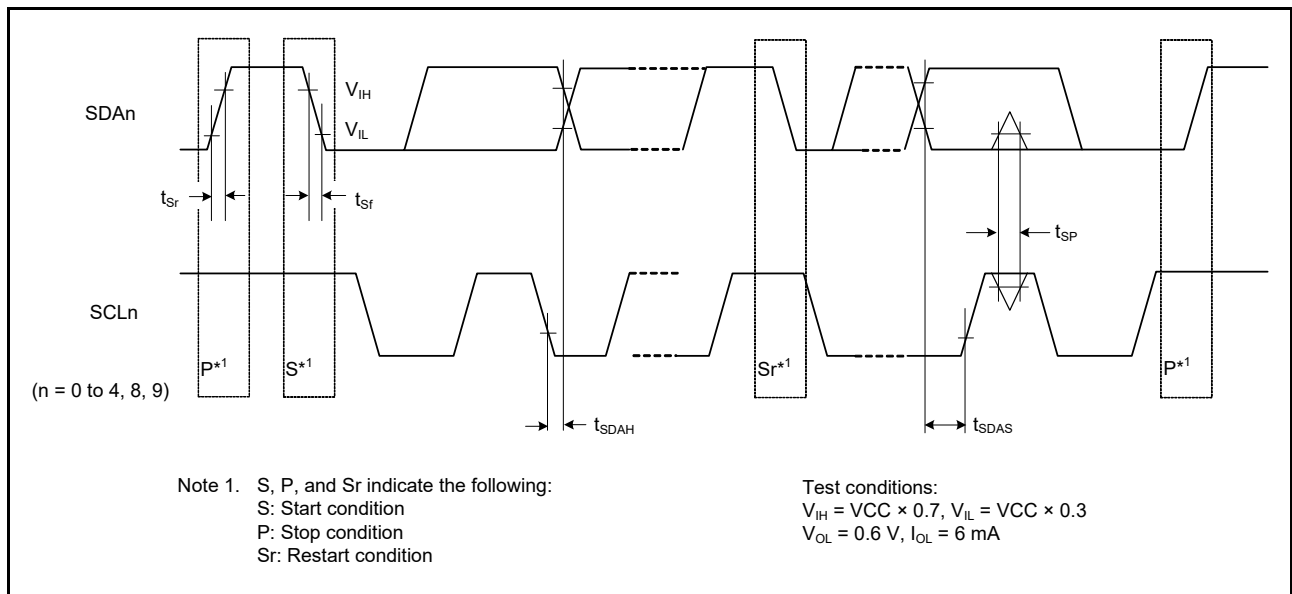


Figure 2.34 SCI simple IIC mode timing

2.3.10 SPI Timing

Table 2.24 SPI timing

Conditions:

For RSPCKA and RSPCKB pins, high drive output is selected with the Port Drive Capability bit in the PmnPFS register.

For other pins, middle drive output is selected in the Port Drive Capability bit in the PmnPFS register.

Parameter		Symbol	Min	Max	Unit*1	Test conditions*2		
SPI	RSPCK clock cycle	Master	t_{SPCyc}	2 (PCLKA ≤ 60 MHz) 4 (PCLKA > 60 MHz)	4096	t_{Pcyc}	Figure 2.35 C = 30 pF	
		Slave		4	4096			
RSPCK clock high pulse width	Master	t_{SPCKWH}	$(t_{SPCyc} - t_{SPCKr} - t_{SPCKf}) / 2 - 3$	-	ns			
	Slave			$2 \times t_{Pcyc}$	-			
RSPCK clock low pulse width	Master	t_{SPCKWL}	$(t_{SPCyc} - t_{SPCKr} - t_{SPCKf}) / 2 - 3$	-	ns			
	Slave			$2 \times t_{Pcyc}$	-			
RSPCK clock rise and fall time	Master	t_{SPCKr}	-	5	ns			
	Slave	t_{SPCKf}	-	1	μs			
Data input setup time	Master	t_{SU}		4	-	ns		Figure 2.36 to Figure 2.41 C = 30 pF
	Slave			5	-			
Data input hold time	Master (PCLKA division ratio set to 1/2)	t_{HF}		0	-	ns		
	Master (PCLKA division ratio set to a value other than 1/2)	t_H		t_{Pcyc}	-			
	Slave	t_H		20	-			
SSL setup time	Master	t_{LEAD}		$N \times t_{SPCyc} - 10^3$	$N \times t_{SPCyc} + 100^3$	ns		
	Slave			$6 \times t_{Pcyc}$	-	ns		
SSL hold time	Master	t_{LAG}		$N \times t_{SPCyc} - 10^4$	$N \times t_{SPCyc} + 100^4$	ns		
	Slave			$6 \times t_{Pcyc}$	-	ns		
Data output delay	Master	t_{OD}		-	6.3	ns		
	Slave			-	20			
Data output hold time	Master	t_{OH}		0	-	ns		
	Slave			0	-			
Successive transmission delay	Master	t_{TD}		$t_{SPCyc} + 2 \times t_{Pcyc}$	$8 \times t_{SPCyc} + 2 \times t_{Pcyc}$	ns		
	Slave			$6 \times t_{Pcyc}$				
MOSI and MISO rise and fall time	Output	t_{Dr}, t_{Df}		-	5	ns		
	Input			-	1	μs		
SSL rise and fall time	Output	t_{SSLr}		-	5	ns		
	Input	t_{SSLf}		-	1	μs		
Slave access time		t_{SA}		-	$2 \times t_{Pcyc} + 28$	ns	Figure 2.40 and Figure 2.41 C = 30 pF	
Slave output release time		t_{REL}		-	$2 \times t_{Pcyc} + 28$			

Note 1. t_{Pcyc} : PCLKA cycle.

- Note 2. Must use pins that have a letter appended to their name, for instance “_A”, “_B”, to indicate group membership. For the SPI interface, the AC portion of the electrical characteristics is measured for each group.
- Note 3. N is set to an integer from 1 to 8 by the SPCKD register.
- Note 4. N is set to an integer from 1 to 8 by the SSLND register.

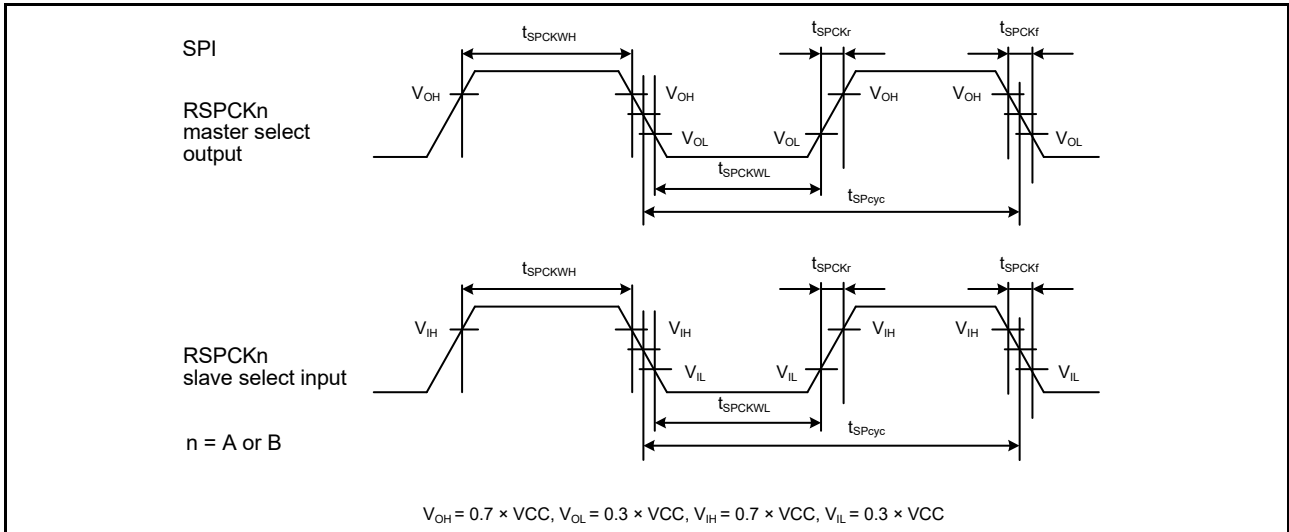


Figure 2.35 SPI clock timing

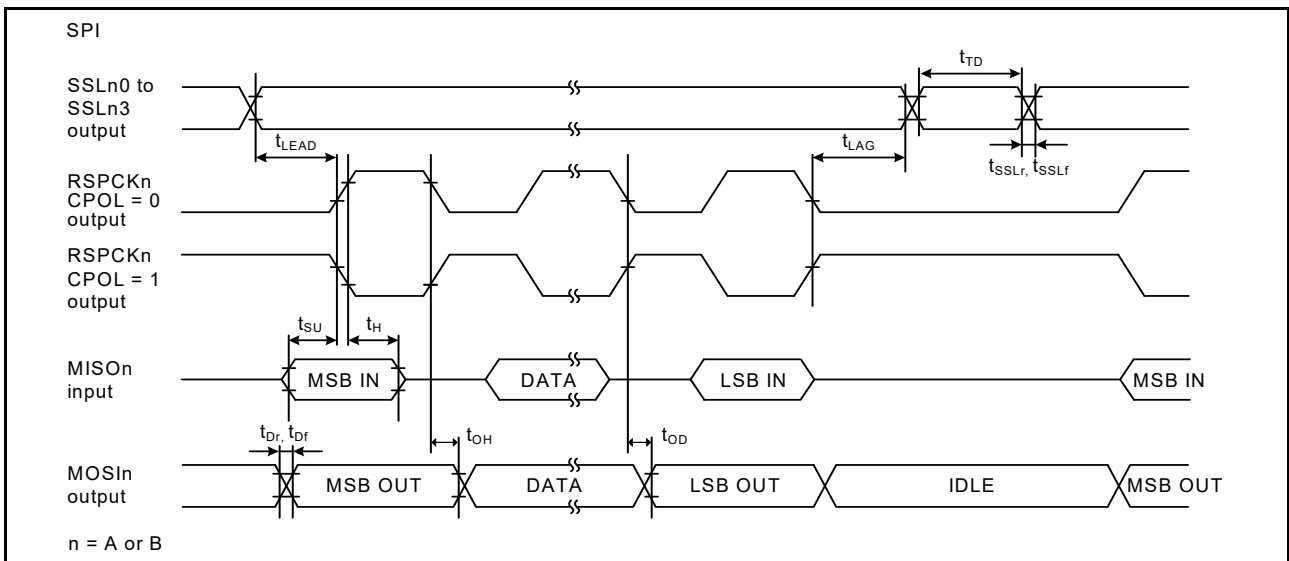


Figure 2.36 SPI timing for master when CPHA = 0

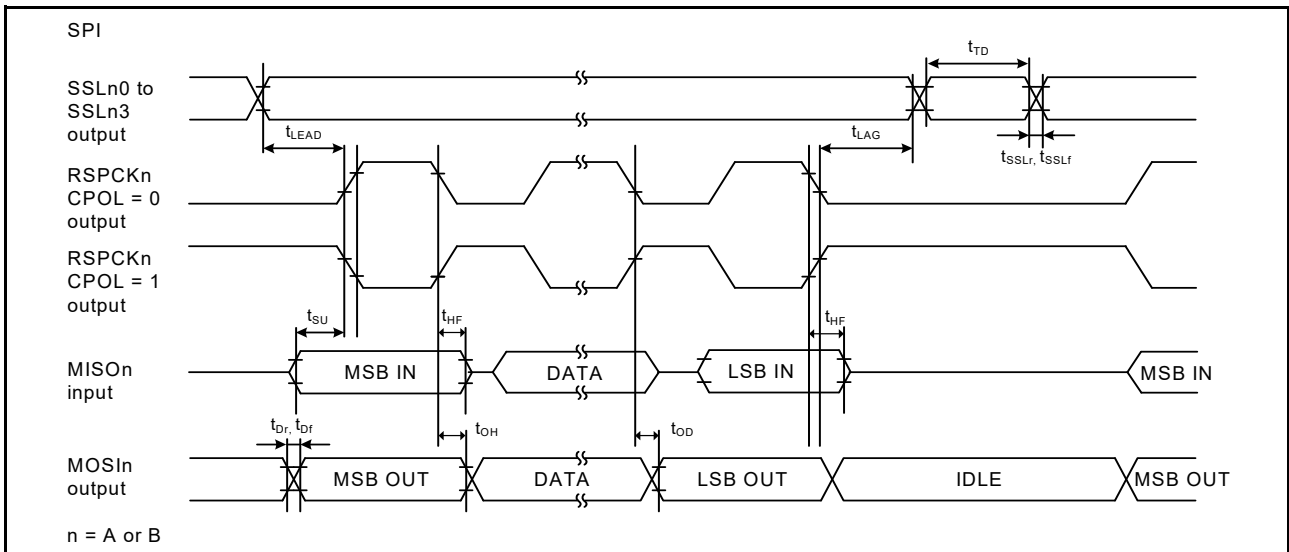


Figure 2.37 SPI timing for master when CPHA = 0 and the bit rate is set to PCLKA/2

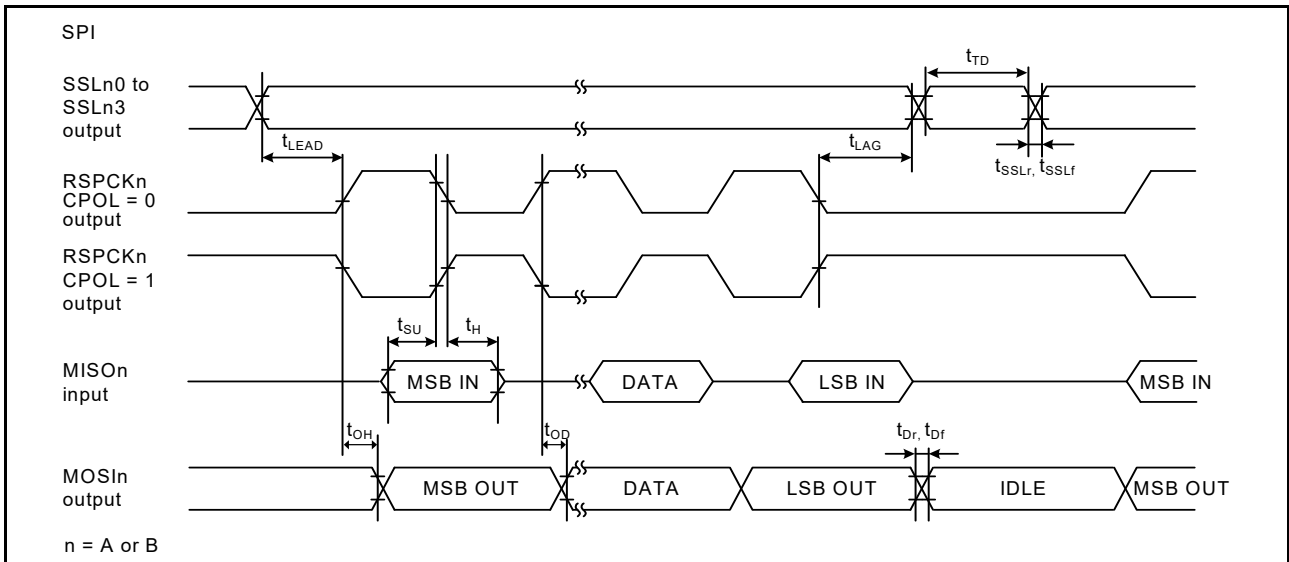


Figure 2.38 SPI timing for master when CPHA = 1

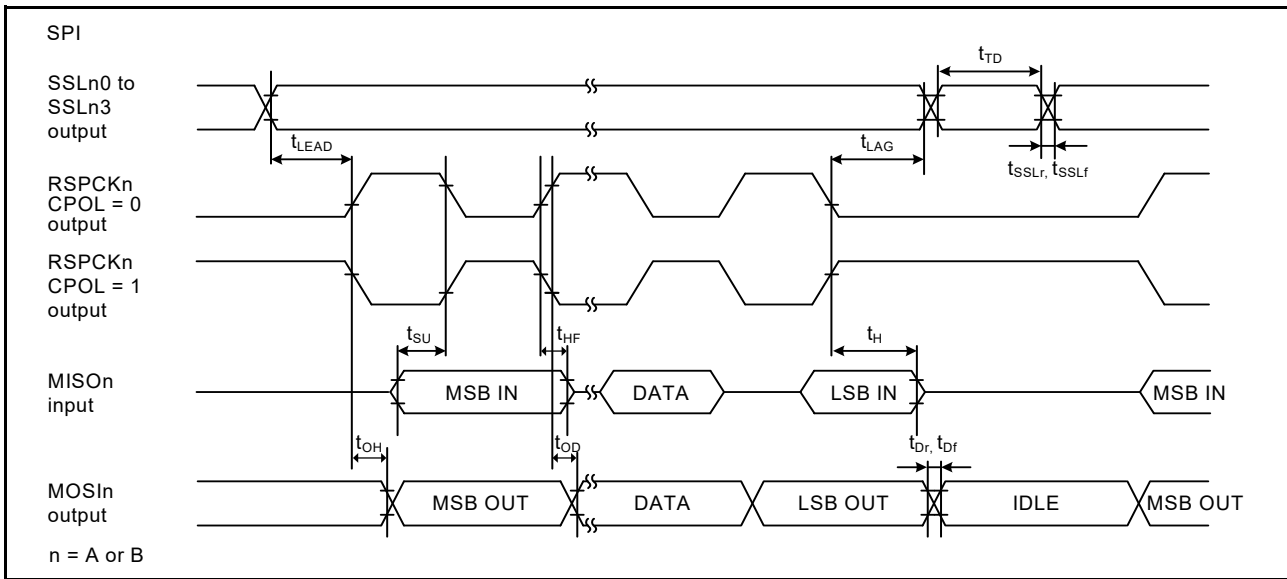


Figure 2.39 RSPI timing for master when CPHA = 1 and the bit rate is set to PCLKA/2

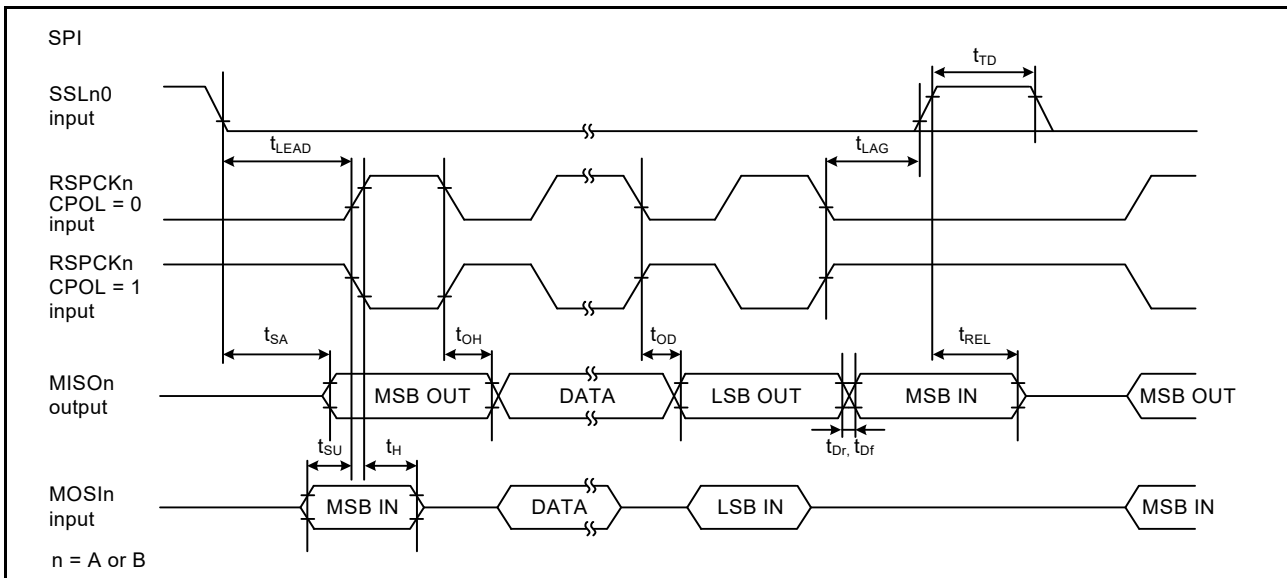


Figure 2.40 SPI timing for slave when CPHA = 0

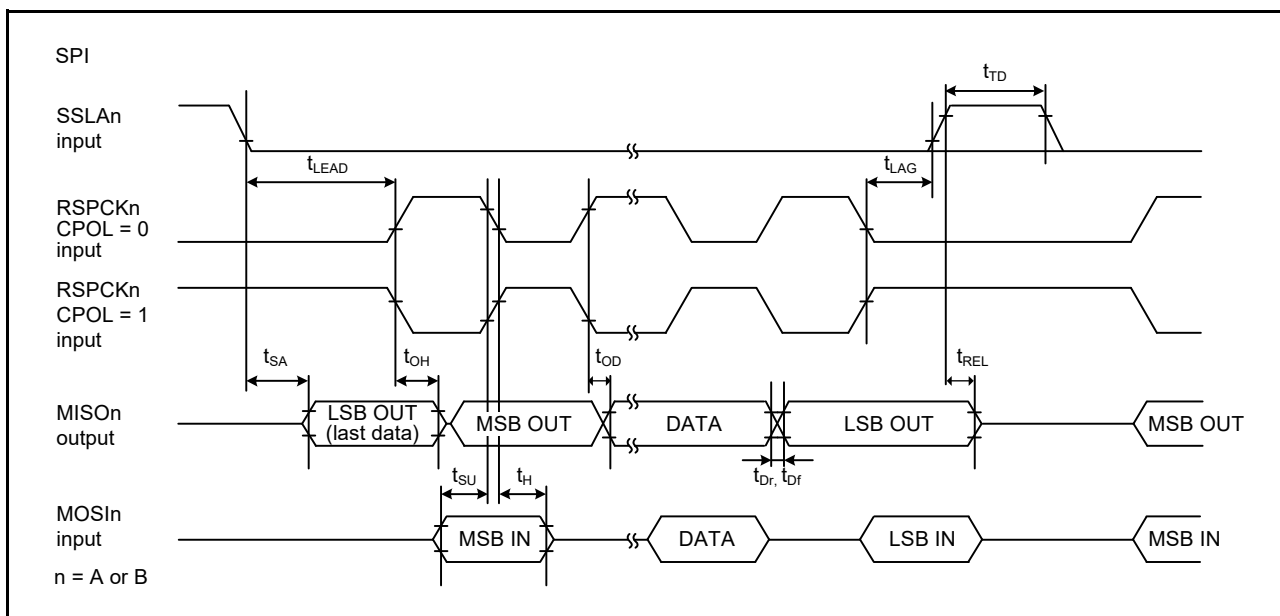


Figure 2.41 SPI timing for slave when CPHA = 1

2.3.11 IIC Timing

Table 2.25 IIC timing (1) (1 of 2)

- (1) Conditions: Middle drive output is selected in the Port Drive Capability bit in the PmnPFS register for the following pins: SDA0_B, SCL0_B, SDA1_A, SCL1_A, SDA1_B, SCL1_B.
- (2) The following pins do not require setting: SCL0_A, SDA0_A.
- (3) Use pins that have a letter appended to their names, for instance “_A” or “_B”, to indicate group membership. For the IIC interface, the AC portion of the electrical characteristics is measured for each group.

Parameter	Symbol	Min*1	Max	Unit	Test conditions*3	
IIC (Standard mode, SMBus) ICFER.FMPE = 0	SCL input cycle time	t_{SCL}	$6 (12) \times t_{IICcyc} + 1300$	-	ns	Figure 2.42
	SCL input high pulse width	t_{SCLH}	$3 (6) \times t_{IICcyc} + 300$	-	ns	
	SCL input low pulse width	t_{SCLL}	$3 (6) \times t_{IICcyc} + 300$	-	ns	
	SCL, SDA input rise time	t_{Sr}	-	1000	ns	
	SCL, SDA input fall time	t_{Sf}	-	300	ns	
	SCL, SDA input spike pulse removal time	t_{SP}	0	$1 (4) \times t_{IICcyc}$	ns	
	SDA input bus free time when wakeup function is disabled	t_{BUF}	$3 (6) \times t_{IICcyc} + 300$	-	ns	
	SDA input bus free time when wakeup function is enabled	t_{BUF}	$3 (6) \times t_{IICcyc} + 4 \times t_{Pcyc} + 300$	-	ns	
	START condition input hold time when wakeup function is disabled	t_{STAH}	$t_{IICcyc} + 300$	-	ns	
	START condition input hold time when wakeup function is enabled	t_{STAH}	$1 (5) \times t_{IICcyc} + t_{Pcyc} + 300$	-	ns	
	Repeated START condition input setup time	t_{STAS}	1000	-	ns	
	STOP condition input setup time	t_{STOS}	1000	-	ns	
	Data input setup time	t_{SDAS}	$t_{IICcyc} + 50$	-	ns	
	Data input hold time	t_{SDAH}	0	-	ns	
SCL, SDA capacitive load	C_b	-	400	pF		

Table 2.25 IIC timing (1) (2 of 2)

- (1) Conditions: Middle drive output is selected in the Port Drive Capability bit in the PmnPFS register for the following pins: SDA0_B, SCL0_B, SDA1_A, SCL1_A, SDA1_B, SCL1_B.
 (2) The following pins do not require setting: SCL0_A, SDA0_A.
 (3) Use pins that have a letter appended to their names, for instance “_A” or “_B”, to indicate group membership. For the IIC interface, the AC portion of the electrical characteristics is measured for each group.

Parameter	Symbol	Min*1	Max	Unit	Test conditions*3	
IIC (Fast mode)	SCL input cycle time	t_{SCL}	$6 (12) \times t_{IICcyc} + 600$	-	ns	Figure 2.42
	SCL input high pulse width	t_{SCLH}	$3 (6) \times t_{IICcyc} + 300$	-	ns	
	SCL input low pulse width	t_{SCLL}	$3 (6) \times t_{IICcyc} + 300$	-	ns	
	SCL, SDA input rise time	t_{Sr}	$20 \times (\text{external pullup voltage}/5.5V)^2$	300	ns	
	SCL, SDA input fall time	t_{Sf}	$20 \times (\text{external pullup voltage}/5.5V)^2$	300	ns	
	SCL, SDA input spike pulse removal time	t_{SP}	0	$1 (4) \times t_{IICcyc}$	ns	
	SDA input bus free time when wakeup function is disabled	t_{BUF}	$3 (6) \times t_{IICcyc} + 300$	-	ns	
	SDA input bus free time when wakeup function is enabled	t_{BUF}	$3 (6) \times t_{IICcyc} + 4 \times t_{Pcyc} + 300$	-	ns	
	START condition input hold time when wakeup function is disabled	t_{STAH}	$t_{IICcyc} + 300$	-	ns	
	START condition input hold time when wakeup function is enabled	t_{STAH}	$1 (5) \times t_{IICcyc} + t_{Pcyc} + 300$	-	ns	
	Repeated START condition input setup time	t_{STAS}	300	-	ns	
	STOP condition input setup time	t_{STOS}	300	-	ns	
	Data input setup time	t_{SDAS}	$t_{IICcyc} + 50$	-	ns	
	Data input hold time	t_{SDAH}	0	-	ns	
SCL, SDA capacitive load	C_b	-	400	pF		

Note: t_{IICcyc} : IIC internal reference clock (IIC ϕ) cycle, t_{Pcyc} : PCLKB cycle.

Note 1. Values in parentheses apply when ICMR3.NF[1:0] is set to 11b while the digital filter is enabled with ICFER.NFE set to 1.

Note 2. Only supported for SCL0_A, SDA0_A.

Note 3. Must use pins that have a letter appended to their name, for instance “_A”, “_B”, to indicate group membership. For the IIC interface, the AC portion of the electrical characteristics is measured for each group.

Table 2.26 IIC timing (2)

Setting of the SCL0_A, SDA0_A pins is not required with the Port Drive Capability bit in the PmnPFS register.

Parameter	Symbol	Min*1,*2	Max	Unit	Test conditions	
IIC (Fast mode+) ICFER.FMPE = 1	SCL input cycle time	t_{SCL}	$6 (12) \times t_{IICcyc} + 240$	-	ns	Figure 2.42
	SCL input high pulse width	t_{SCLH}	$3 (6) \times t_{IICcyc} + 120$	-	ns	
	SCL input low pulse width	t_{SCLL}	$3 (6) \times t_{IICcyc} + 120$	-	ns	
	SCL, SDA input rise time	t_{Sr}	-	120	ns	
	SCL, SDA input fall time	t_{Sf}	-	120	ns	
	SCL, SDA input spike pulse removal time	t_{SP}	0	$1 (4) \times t_{IICcyc}$	ns	
	SDA input bus free time when wakeup function is disabled	t_{BUF}	$3 (6) \times t_{IICcyc} + 120$	-	ns	
	SDA input bus free time when wakeup function is enabled	t_{BUF}	$3 (6) \times t_{IICcyc} + 4 \times t_{Pcyc} + 120$	-	ns	
	Start condition input hold time when wakeup function is disabled	t_{STAH}	$t_{IICcyc} + 120$	-	ns	
	START condition input hold time when wakeup function is enabled	t_{STAH}	$1 (5) \times t_{IICcyc} + t_{Pcyc} + 120$	-	ns	
	Restart condition input setup time	t_{STAS}	120	-	ns	
	Stop condition input setup time	t_{STOS}	120	-	ns	
	Data input setup time	t_{SDAS}	$t_{IICcyc} + 30$	-	ns	
	Data input hold time	t_{SDAH}	0	-	ns	
	SCL, SDA capacitive load	C_b	-	550	pF	

Note: t_{IICcyc} : IIC internal reference clock (IIC ϕ) cycle, t_{Pcyc} : PCLKB cycle.

Note 1. Values in parentheses apply when ICMR3.NF[1:0] is set to 11b while the digital filter is enabled with ICFER.NFE set to 1.

Note 2. C_b indicates the total capacity of the bus line.

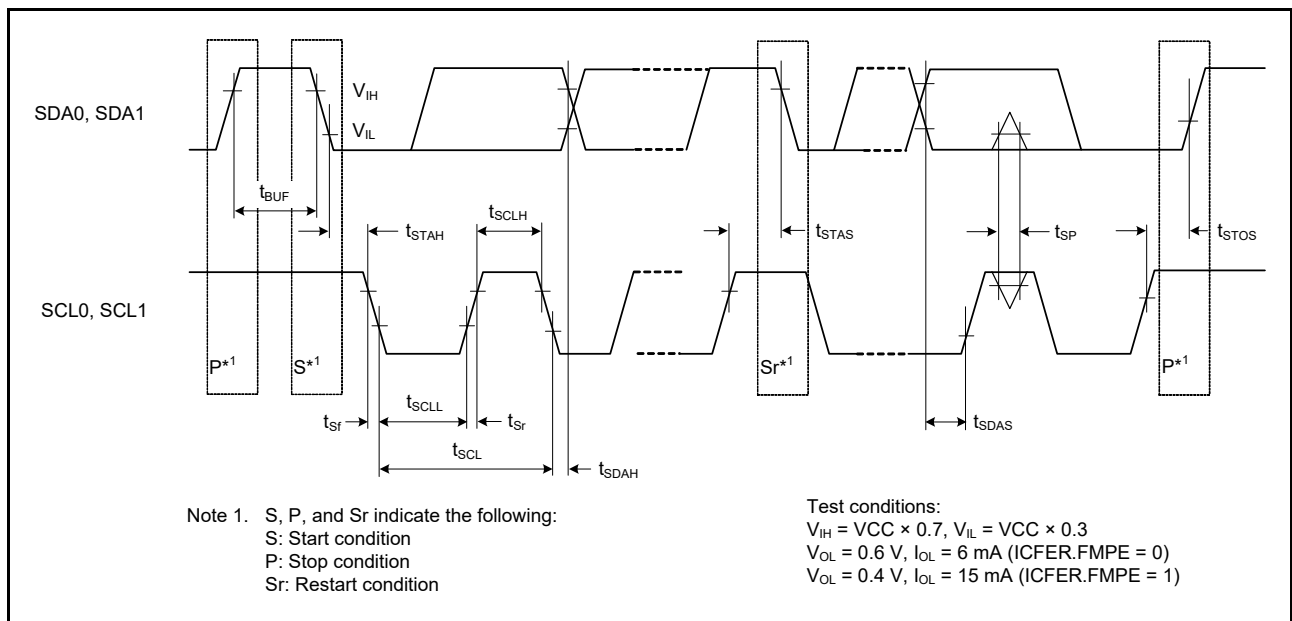


Figure 2.42 I2C bus interface input/output timing

2.4 ADC12 Characteristics

Table 2.27 A/D conversion characteristics for unit 0 (1 of 2)

Conditions: PCLKC = 1 to 60 MHz

Parameter			Min	Typ	Max	Unit	Test conditions
Frequency			1	-	60	MHz	-
Analog input capacitance			-	-	30	pF	-
Quantization error			-	±0.5	-	LSB	-
Resolution			-	-	12	Bits	-
Channel-dedicated sample-and-hold circuits in use*3 (AN000 to AN002)	Conversion time*1 (operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	1.06 (0.4 + 0.25)*2	-	-	μs	<ul style="list-style-type: none"> Sampling of channel-dedicated sample-and-hold circuits in 24 states Sampling in 15 states
	Offset error		-	±1.5	±3.5	LSB	AN000 to AN002 = 0.25 V
	Full-scale error		-	±1.5	±3.5	LSB	AN000 to AN002 = VREFH0 - 0.25 V
	Absolute accuracy		-	±2.5	±5.5	LSB	-
	DNL pseudo-differential nonlinearity error		-	±1.0	±2.0	LSB	-
	INL integral nonlinearity error		-	±1.5	±3.0	LSB	-
	Holding characteristics of sample-and hold circuits		-	-	20	μs	-
Dynamic range			0.25	-	VREFH0 - 0.25	V	-
Channel-dedicated sample-and-hold circuits not in use (AN000 to AN002)	Conversion time*1 (operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.48 (0.267)*2	-	-	μs	Sampling in 16 states
	Offset error		-	±1.0	±2.5	LSB	-
	Full-scale error		-	±1.0	±2.5	LSB	-
	Absolute accuracy		-	±2.0	±4.5	LSB	-
	INL integral nonlinearity error		-	±1.0	±2.5	LSB	-
High-precision channels (AN003, AN005, AN006)	Conversion time*1 (operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.48 (0.267)*2	-	-	μs	Sampling in 16 states
		Max. = 400 Ω	0.40 (0.183)*2	-	-	μs	Sampling in 11 states VCC = AVCC0 = 3.0 to 3.6 V 3.0 V ≤ VREFH0 ≤ AVCC0
	Offset error		-	±1.0	±2.5	LSB	-
	Full-scale error		-	±1.0	±2.5	LSB	-
	Absolute accuracy		-	±2.0	±4.5	LSB	-
	INL integral nonlinearity error		-	±1.0	±2.5	LSB	-
High-precision channels (AN007)	Conversion time*1 (operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.75 (0.533)*2	-	-	μs	Sampling in 32 states
	Offset error		-	±1.0	±2.5	LSB	-
	Full-scale error		-	±1.0	±2.5	LSB	-
	Absolute accuracy		-	±2.0	±4.5	LSB	-
	INL integral nonlinearity error		-	±1.0	±2.5	LSB	-

Table 2.27 A/D conversion characteristics for unit 0 (2 of 2)

Conditions: PCLKC = 1 to 60 MHz

Parameter			Min	Typ	Max	Unit	Test conditions
Normal-precision channels (AN016 to AN018, AN020)	Conversion time*1 (Operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.88 (0.667)*2	-	-	μs	Sampling in 40 states
	Offset error		-	±1.0	±5.5	LSB	-
	Full-scale error		-	±1.0	±5.5	LSB	-
	Absolute accuracy		-	±2.0	±7.5	LSB	-
	DNL pseudo-differential nonlinearity error		-	±0.5	±4.5	LSB	-
	INL integral nonlinearity error		-	±1.0	±5.5	LSB	-

Note: These specification values apply when there is no access to the external bus during A/D conversion. If access occurs during A/D conversion, the values might not fall within the indicated ranges.

The use of ports 0 as digital outputs is not allowed when the 12-bit A/D converter is used.

The characteristics apply when AVCC0, AVSS0, VREFH0, VREFH, VREFL0, VREFL, and 12-bit A/D converter input voltage are stable.

Note 1. The conversion time includes the sampling and comparison times. The number of sampling states is indicated for the test conditions.

Note 2. Values in parentheses indicate the sampling time.

Note 3. When simultaneously using channel-dedicated sample-and-hold circuits in unit 0 and unit 1, see [Table 2.29](#).

Table 2.28 A/D conversion characteristics for unit 1 (1 of 2)

Conditions: PCLKC = 1 to 60 MHz

Parameter			Min	Typ	Max	Unit	Test conditions
Frequency			1	-	60	MHz	-
Analog input capacitance			-	-	30	pF	-
Quantization error			-	±0.5	-	LSB	-
Resolution			-	-	12	Bits	-
Channel-dedicated sample-and-hold circuits in use*3 (AN100 to AN102)	Conversion time*1 (operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	1.06 (0.4 + 0.25)*2	-	-	μs	<ul style="list-style-type: none"> Sampling of channel-dedicated sample-and-hold circuits in 24 states Sampling in 15 states
	Offset error		-	±1.5	±3.5	LSB	AN100 to AN102 = 0.25 V
	Full-scale error		-	±1.5	±3.5	LSB	AN100 to AN102 = VREFH - 0.25 V
	Absolute accuracy		-	±2.5	±5.5	LSB	-
	DNL pseudo-differential nonlinearity error		-	±1.0	±2.0	LSB	-
	INL integral nonlinearity error		-	±1.5	±3.0	LSB	-
	Holding characteristics of sample-and-hold circuits		-	-	20	μs	-
Dynamic range			0.25	-	VREFH - 0.25	V	-
Channel-dedicated sample-and-hold circuits not in use (AN100 to AN102)	Conversion time*1 (Operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.48 (0.267)*2	-	-	μs	Sampling in 16 states
	Offset error		-	±1.0	±2.5	LSB	-
	Full-scale error		-	±1.0	±2.5	LSB	-
	Absolute accuracy		-	±2.0	±4.5	LSB	-
	DNL pseudo-differential nonlinearity error		-	±0.5	±1.5	LSB	-
	INL integral nonlinearity error		-	±1.0	±2.5	LSB	-

Table 2.28 A/D conversion characteristics for unit 1 (2 of 2)

Conditions: PCLKC = 1 to 60 MHz

Parameter			Min	Typ	Max	Unit	Test conditions
High-precision channels (AN105, AN106)	Conversion time*1 (Operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.48 (0.267)*2	-	-	μs	Sampling in 16 states
		Max. = 400 Ω	0.40 (0.183)*2	-	-	μs	Sampling in 11 states VCC = AVCC0 = 3.0 to 3.6 V 3.0 V ≤ VREFH ≤ AVCC0
	Offset error		-	±1.0	±2.5	LSB	-
	Full-scale error		-	±1.0	±2.5	LSB	-
	Absolute accuracy		-	±2.0	±4.5	LSB	-
	DNL pseudo-differential nonlinearity error		-	±0.5	±1.5	LSB	-
	INL integral nonlinearity error		-	±1.0	±2.5	LSB	-
High-precision channels (AN107)	Conversion time*1 (Operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.75 (0.533)*2	-	-	μs	Sampling in 32 states
		Max. = 400 Ω	0.40 (0.183)*2	-	-	μs	Sampling in 11 states VCC = AVCC0 = 3.0 to 3.6 V 3.0 V ≤ VREFH ≤ AVCC0
	Offset error		-	±1.0	±2.5	LSB	-
	Full-scale error		-	±1.0	±2.5	LSB	-
	Absolute accuracy		-	±2.0	±4.5	LSB	-
	DNL pseudo-differential nonlinearity error		-	±0.5	±1.5	LSB	-
Normal-precision channels (AN116, AN117)	Conversion time*1 (Operation at PCLKC = 60 MHz)	Permissible signal source impedance Max. = 1 kΩ	0.88 (0.667)*2	-	-	μs	Sampling in 40 states
		Max. = 400 Ω	0.40 (0.183)*2	-	-	μs	Sampling in 11 states VCC = AVCC0 = 3.0 to 3.6 V 3.0 V ≤ VREFH ≤ AVCC0
	Offset error		-	±1.0	±5.5	LSB	-
	Full-scale error		-	±1.0	±5.5	LSB	-
	Absolute accuracy		-	±2.0	±7.5	LSB	-
	DNL pseudo-differential nonlinearity error		-	±0.5	±4.5	LSB	-
INL integral nonlinearity error		-	±1.0	±5.5	LSB	-	

Note: These specification values apply when there is no access to the external bus during A/D conversion. If access occurs during A/D conversion, the values might not fall within the indicated ranges.

The use of ports 0 as digital outputs is not allowed when the 12-bit A/D converter is used.

The characteristics apply when AVCC0, AVSS0, VREFH0, VREFH, VREFL0, VREFL, and 12-bit A/D converter input voltage are stable.

Note 1. The conversion time includes the sampling and comparison times. The number of sampling states is indicated for the test conditions.

Note 2. Values in parentheses indicate the sampling time.

Note 3. When simultaneously using channel-dedicated sample-and-hold circuits in unit 0 and unit 1, see [Table 2.29](#).

Table 2.29 A/D conversion characteristics for simultaneous use of channel-dedicated sample-and-hold circuits in unit 0 and unit 1

Conditions: PCLKC = 30/60 MHz

Parameter		Min	Typ	Max	Test conditions
Channel-dedicated sample-and-hold circuits in use with continuous sampling function enabled (AN000 to AN002)	Offset error	-	±1.5	±5.0	<ul style="list-style-type: none"> PCLKC = 60 MHz Sampling in 15 states
	Full-scale error	-	±2.5	±5.0	
	Absolute accuracy	-	±4.0	±8.0	
Channel-dedicated sample-and-hold circuits in use with continuous sampling function enabled (AN100 to AN102)	Offset error	-	±1.5	±5.0	<ul style="list-style-type: none"> PCLKC = 30 MHz Sampling in 7 states
	Full-scale error	-	±2.5	±5.0	
	Absolute accuracy	-	±4.0	±8.0	
Channel-dedicated sample-and-hold circuits in use with continuous sampling function enabled (AN000 to AN002)	Offset error	-	±1.5	±3.5	<ul style="list-style-type: none"> PCLKC = 30 MHz Sampling in 7 states
	Full-scale error	-	±1.5	±3.5	
	Absolute accuracy	-	±3.0	+4.5/-6.5	
Channel-dedicated sample-and-hold circuits in use with continuous sampling function enabled (AN100 to AN102)	Offset error	-	±1.5	±3.5	<ul style="list-style-type: none"> PCLKC = 30 MHz Sampling in 7 states
	Full-scale error	-	±1.5	±3.5	
	Absolute accuracy	-	±3.0	+4.5/-6.5	

Note: When simultaneously using channel-dedicated sample-and-hold circuits in unit 0 and unit 1, setting the ADHMSR.SHMD bit to 1 is recommended.

Table 2.30 A/D internal reference voltage characteristics

Parameter	Min	Typ	Max	Unit	Test conditions
A/D internal reference voltage	1.13	1.18	1.23	V	-
Sampling time	4.15	-	-	μs	-

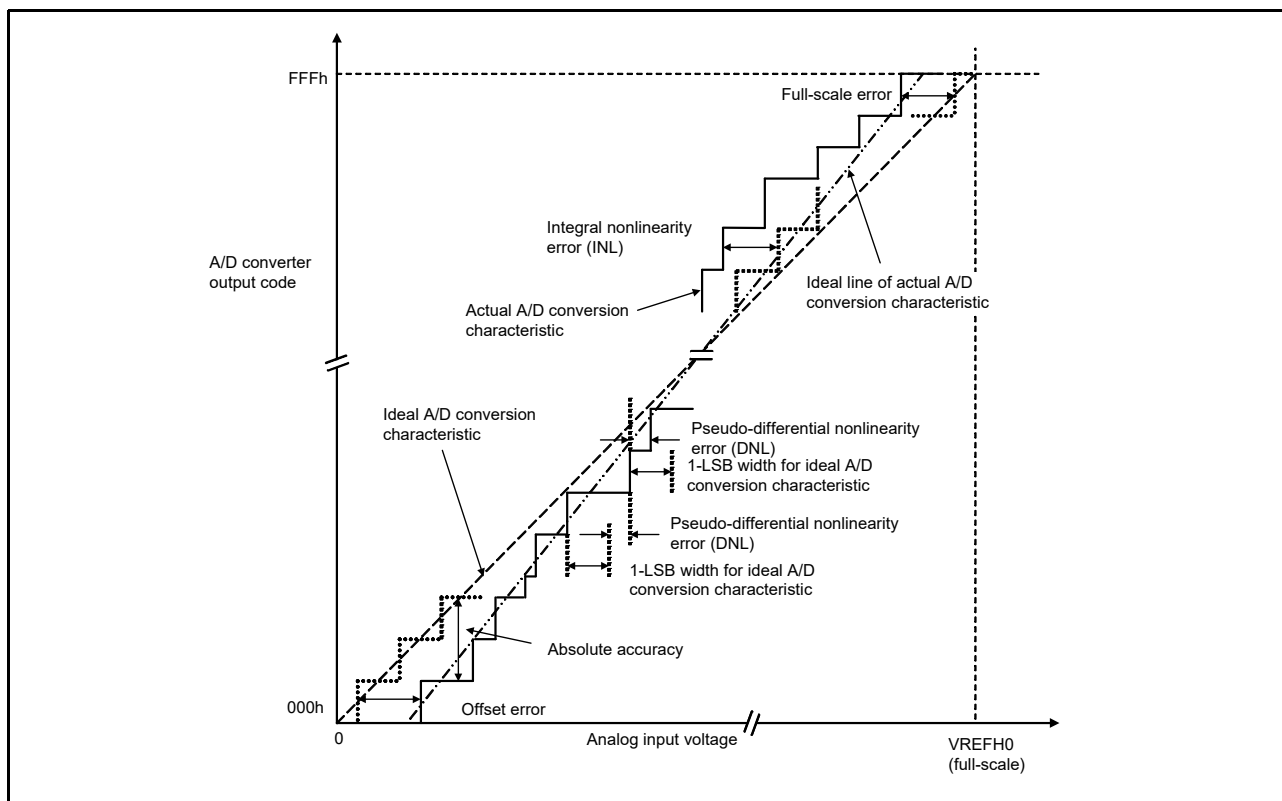


Figure 2.43 Illustration of ADC12 characteristic terms

Absolute accuracy

Absolute accuracy is the difference between output code based on the theoretical A/D conversion characteristics, and the actual A/D conversion result. When measuring absolute accuracy, the voltage at the midpoint of the width of the analog input voltage (1-LSB width), which can meet the expectation of outputting an equal code based on the theoretical A/D conversion characteristics, is used as the analog input voltage. For example, if 12-bit resolution is used and the reference voltage VREFH0 is 3.072 V, then the 1-LSB width becomes 0.75 mV, and 0 mV, 0.75 mV, and 1.5 mV are used as the analog input voltages. If the analog input voltage is 6 mV, an absolute accuracy of ±5 LSB means that the actual A/D conversion result is in the range of 003h to 00Dh, though an output code of 008h can be expected from the theoretical A/D conversion characteristics.

Integral nonlinearity error (INL)

Integral nonlinearity error is the maximum deviation between the ideal line when the measured offset and full-scale errors are zeroed, and the actual output code.

Pseudo-differential nonlinearity error (DNL)

Pseudo-differential nonlinearity error is the difference between the 1-LSB width based on the ideal A/D conversion characteristics and the width of the actual output code.

Offset error

Offset error is the difference between the transition point of the ideal first output code and the actual first output code.

Full-scale error

Full-scale error is the difference between the transition point of the ideal last output code and the actual last output code.

2.5 DAC12 Characteristics

Table 2.31 D/A conversion characteristics

Parameter	Min	Typ	Max	Unit	Test conditions
Resolution	-	-	12	Bits	-
Without output amplifier					
Absolute accuracy	-	-	±24	LSB	Resistive load 2 MΩ
INL	-	±2.0	±8.0	LSB	Resistive load 2 MΩ
DNL	-	±1.0	±2.0	LSB	-
Output impedance	-	8.5	-	kΩ	-
Conversion time	-	-	3.0	μs	Resistive load 2 MΩ, Capacitive load 20 pF
Output voltage range	0	-	VREFH	V	-
With output amplifier					
INL	-	±2.0	±4.0	LSB	-
DNL	-	±1.0	±2.0	LSB	-
Conversion time	-	-	4.0	μs	-
Resistive load	5	-	-	kΩ	-
Capacitive load	-	-	50	pF	-
Output voltage range	0.2	-	VREFH - 0.2	V	-

2.6 TSN Characteristics

Table 2.32 TSN characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
Relative accuracy	-	-	±1.0	-	°C	-
Temperature slope	-	-	4.0	-	mV/°C	-
Output voltage (at 25°C)	-	-	1.24	-	V	-
Temperature sensor start time	t _{START}	-	-	30	μs	-
Sampling time	-	4.15	-	-	μs	-

2.7 OSC Stop Detect Characteristics

Table 2.33 Oscillation stop detection circuit characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
Detection time	t _{dr}	-	-	1	ms	Figure 2.44

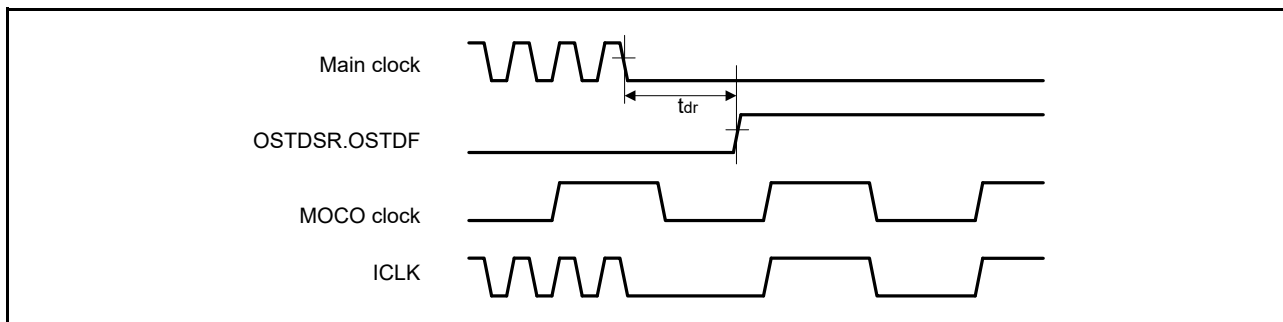


Figure 2.44 Oscillation stop detection timing

2.8 POR and LVD Characteristics

Table 2.34 Power-on reset circuit and voltage detection circuit characteristics

Parameter			Symbol	Min	Typ	Max	Unit	Test conditions	
Voltage detection level	Power-on reset (POR)	DPSBYCR.DEEPCUT[1:0] = 00b or 01b	V_{POR}	2.5	2.6	2.7	V	Figure 2.45	
		DPSBYCR.DEEPCUT[1:0] = 11b		1.8	2.25	2.7			
	Voltage detection circuit (LVD0)		V_{det0_1}	2.84	2.94	3.04		Figure 2.46	
				V_{det0_2}	2.77	2.87			2.97
				V_{det0_3}	2.70	2.80			2.90
	Voltage detection circuit (LVD1)		V_{det1_1}	2.89	2.99	3.09		Figure 2.47	
				V_{det1_2}	2.82	2.92			3.02
				V_{det1_3}	2.75	2.85			2.95
	Voltage detection circuit (LVD2)		V_{det2_1}	2.89	2.99	3.09		Figure 2.48	
				V_{det2_2}	2.82	2.92			3.02
				V_{det2_3}	2.75	2.85			2.95
	Internal reset time	Power-on reset time		t_{POR}	-	4.5		-	ms
LVD0 reset time		t_{LVD0}	-	0.51	-	Figure 2.46			
LVD1 reset time		t_{LVD1}	-	0.38	-	Figure 2.47			
LVD2 reset time		t_{LVD2}	-	0.38	-	Figure 2.48			
Minimum VCC down time*1			t_{VOFF}	200	-	-	μ s	Figure 2.45, Figure 2.46	
Response delay			t_{det}	-	-	200	μ s	Figure 2.45 to Figure 2.48	
LVD operation stabilization time (after LVD is enabled)			$t_{d(E-A)}$	-	-	10	μ s	Figure 2.47, Figure 2.48	
Hysteresis width (LVD1 and LVD2)			V_{LVH}	-	70	-	mV		

Note 1. The minimum VCC down time indicates the time when VCC is below the minimum value of voltage detection levels V_{POR} , V_{det1} , and V_{det2} for POR and LVD.

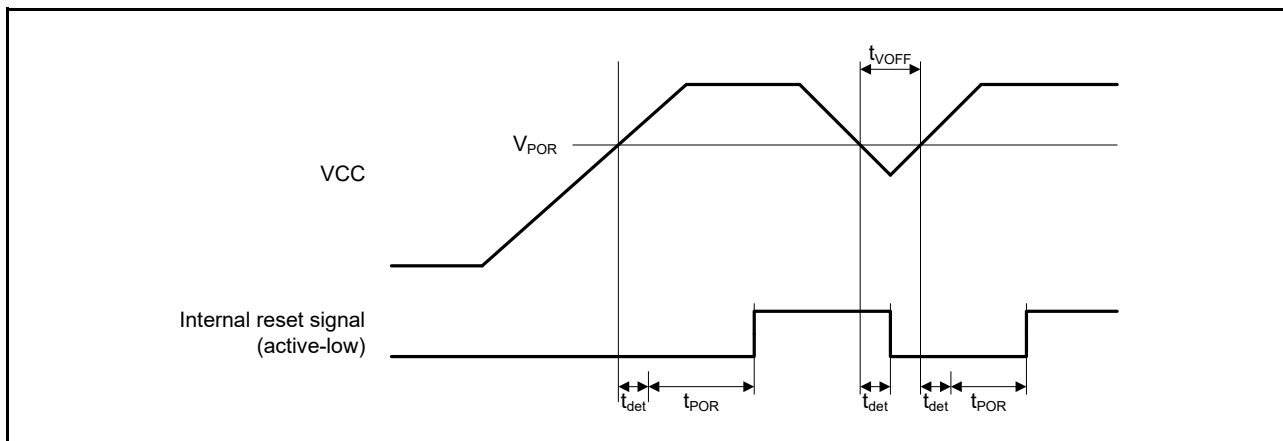


Figure 2.45 Power-on reset timing

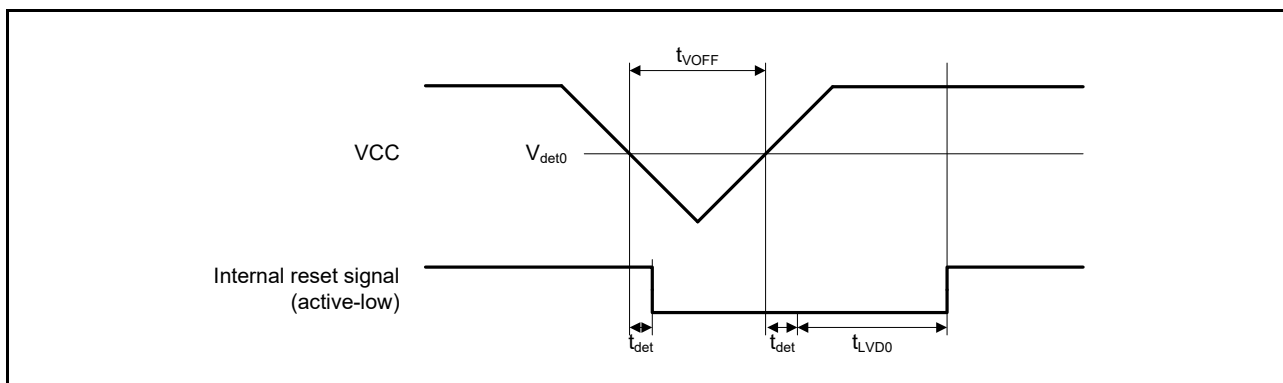


Figure 2.46 Voltage detection circuit timing (V_{det0})

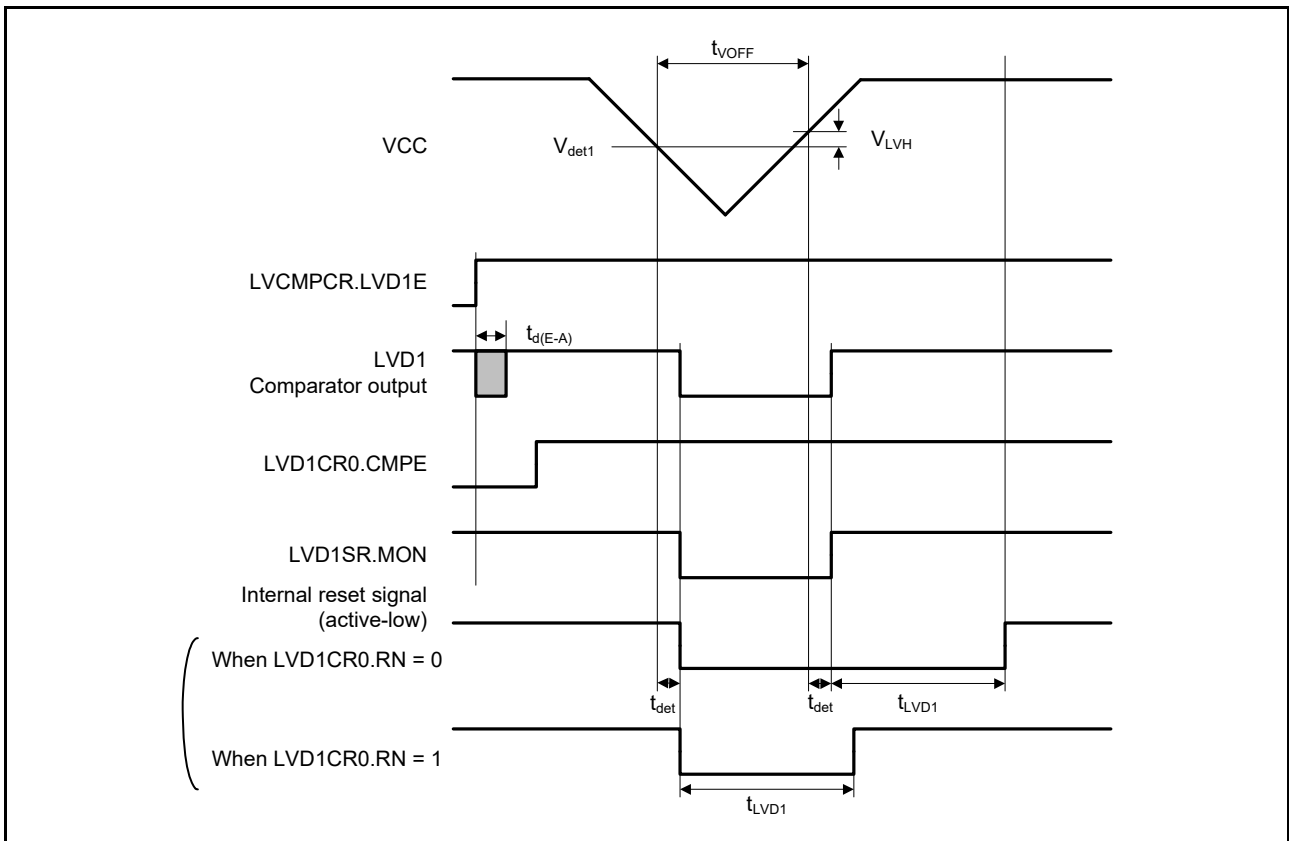


Figure 2.47 Voltage detection circuit timing (V_{det1})

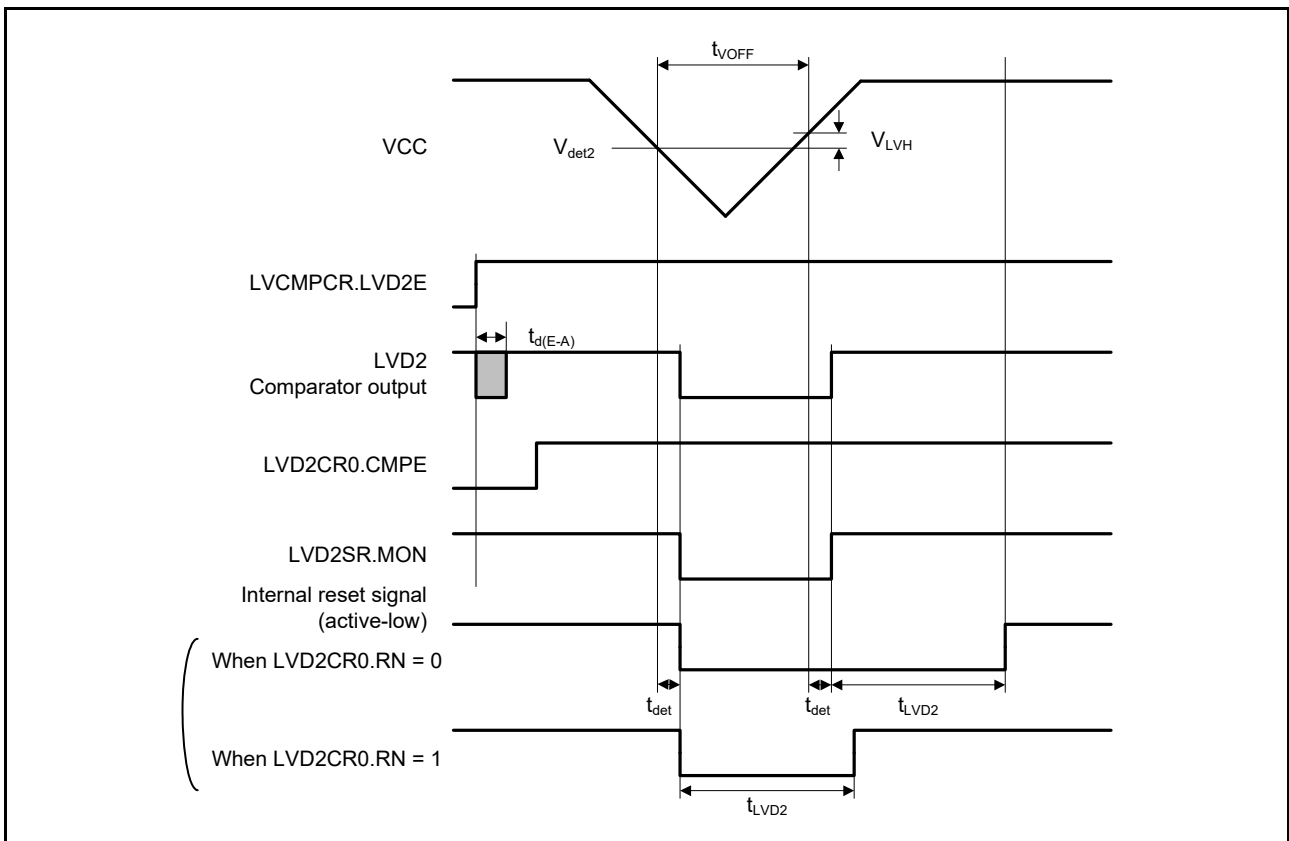


Figure 2.48 Voltage detection circuit timing (V_{det2})

2.9 ACMPHS Characteristics

Table 2.35 ACMPHS characteristics

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
Reference voltage range	VREF	0	-	AVCC0	V	-
Input voltage range	VI	0	-	AVCC0	V	-
Output delay*1	Td	-	50	100	ns	VI = VREF ± 100 mV
Internal reference voltage	Vref	1.13	1.18	1.23	V	-

Note 1. This value is the internal propagation delay.

2.10 PGA Characteristics

Table 2.36 PGA characteristics in single mode

Parameter	Symbol	Min	Typ	Max	Unit
PGAVSS input voltage range	PGAVSS	0	-	0	V
	AIN0 (G = 2.000)	$0.050 \times AVCC0$	-	$0.45 \times AVCC0$	V
	AIN1 (G = 2.500)	$0.047 \times AVCC0$	-	$0.360 \times AVCC0$	V
	AIN2 (G = 2.667)	$0.046 \times AVCC0$	-	$0.337 \times AVCC0$	V
	AIN3 (G = 2.857)	$0.046 \times AVCC0$	-	$0.32 \times AVCC0$	V
	AIN4 (G = 3.077)	$0.045 \times AVCC0$	-	$0.292 \times AVCC0$	V
	AIN5 (G = 3.333)	$0.044 \times AVCC0$	-	$0.265 \times AVCC0$	V
	AIN6 (G = 3.636)	$0.042 \times AVCC0$	-	$0.247 \times AVCC0$	V
	AIN7 (G = 4.000)	$0.040 \times AVCC0$	-	$0.212 \times AVCC0$	V
	AIN8 (G = 4.444)	$0.036 \times AVCC0$	-	$0.191 \times AVCC0$	V
	AIN9 (G = 5.000)	$0.033 \times AVCC0$	-	$0.17 \times AVCC0$	V
	AIN10 (G = 5.714)	$0.031 \times AVCC0$	-	$0.148 \times AVCC0$	V
	AIN11 (G = 6.667)	$0.029 \times AVCC0$	-	$0.127 \times AVCC0$	V
	AIN12 (G = 8.000)	$0.027 \times AVCC0$	-	$0.09 \times AVCC0$	V
	AIN13 (G = 10.000)	$0.025 \times AVCC0$	-	$0.08 \times AVCC0$	V
AIN14 (G = 13.333)	$0.023 \times AVCC0$	-	$0.06 \times AVCC0$	V	
Gain error	Gerr0 (G = 2.000)	-1.0	-	1.0	%
	Gerr1 (G = 2.500)	-1.0	-	1.0	%
	Gerr2 (G = 2.667)	-1.0	-	1.0	%
	Gerr3 (G = 2.857)	-1.0	-	1.0	%
	Gerr4 (G = 3.077)	-1.0	-	1.0	%
	Gerr5 (G = 3.333)	-1.5	-	1.5	%
	Gerr6 (G = 3.636)	-1.5	-	1.5	%
	Gerr7 (G = 4.000)	-1.5	-	1.5	%
	Gerr8 (G = 4.444)	-2.0	-	2.0	%
	Gerr9 (G = 5.000)	-2.0	-	2.0	%
	Gerr10 (G = 5.714)	-2.0	-	2.0	%
	Gerr11 (G = 6.667)	-2.0	-	2.0	%
	Gerr12 (G = 8.000)	-2.0	-	2.0	%
	Gerr13 (G = 10.000)	-2.0	-	2.0	%
	Gerr14 (G = 13.333)	-2.0	-	2.0	%
Offset error	Voff	-8	-	8	mV

Table 2.37 PGA characteristics in pseudo-differential mode

Parameter		Symbol	Min	Typ	Max	Unit
PGAVSS input voltage range		PGAVSS	-0.5	-	0.3	V
Pseudo-differential input voltage range	G = 1.500	AIN-PGAVSS	-0.5	-	0.5	V
	G = 2.333		-0.4	-	0.4	V
	G = 4.000		-0.2	-	0.2	V
	G = 5.667		-0.15	-	0.15	V
Gain error	G = 1.500	Gerr	-1.0	-	1.0	%
	G = 2.333		-1.0	-	1.0	
	G = 4.000		-1.0	-	1.0	
	G = 5.667		-1.0	-	1.0	

2.11 Flash Memory Characteristics

2.11.1 Code Flash Memory Characteristics

Table 2.38 Code flash memory characteristics

Conditions: Program or erase: FCLK = 4 to 60 MHz

Read: FCLK ≤ 60 MHz

Parameter		Symbol	FCLK = 4 MHz			20 MHz ≤ FCLK ≤ 60 MHz			Unit	Test conditions
			Min	Typ	Max	Min	Typ	Max		
Programming time $N_{PEC} \leq 100$ times	128-byte	t_{P128}	-	0.75	13.2	-	0.34	6.0	ms	
	8-KB	t_{P8K}	-	49	176	-	22	80	ms	
	32-KB	t_{P32K}	-	194	704	-	88	320	ms	
Programming time $N_{PEC} > 100$ times	128-byte	t_{P128}	-	0.91	15.8	-	0.41	7.2	ms	
	8-KB	t_{P8K}	-	60	212	-	27	96	ms	
	32-KB	t_{P32K}	-	234	848	-	106	384	ms	
Erasure time $N_{PEC} \leq 100$ times	8-KB	t_{E8K}	-	78	216	-	43	120	ms	
	32-KB	t_{E32K}	-	283	864	-	157	480	ms	
Erasure time $N_{PEC} > 100$ times	8-KB	t_{E8K}	-	94	260	-	52	144	ms	
	32-KB	t_{E32K}	-	341	1040	-	189	576	ms	
Reprogramming/erasure cycle*4		N_{PEC}	10000*1	-	-	10000*1	-	-	Times	
Suspend delay during programming		t_{SPD}	-	-	264	-	-	120	μs	
First suspend delay during erasure in suspend priority mode		t_{SESD1}	-	-	216	-	-	120	μs	
Second suspend delay during erasure in suspend priority mode		t_{SESD2}	-	-	1.7	-	-	1.7	ms	
Suspend delay during erasure in erasure priority mode		t_{SEED}	-	-	1.7	-	-	1.7	ms	
Forced stop command		t_{FD}	-	-	32	-	-	20	μs	
Data hold time*2		t_{DRP}	10*2, *3	-	-	10*2, *3	-	-	Years	Ta = +85°C
			30*2, *3	-	-	30*2, *3	-	-		

Note 1. This is the minimum number of times to guarantee all the characteristics after reprogramming. The guaranteed range is from 1 to the minimum value.

Note 2. This indicates the minimum value of the characteristic when reprogramming is performed within the specified range.

Note 3. This result is obtained from reliability testing.

Note 4. The reprogram/erase cycle is the number of erasures for each block. When the reprogram/erase cycle is n times ($n = 10000$), erasing can be performed n times for each block. For example, when 128-byte programming is performed 64 times for different addresses in 8-KB blocks, and then the entire block is erased, the reprogram/erase cycle is counted as one. However, programming the same address several times as one erasure is not enabled. Overwriting is prohibited.

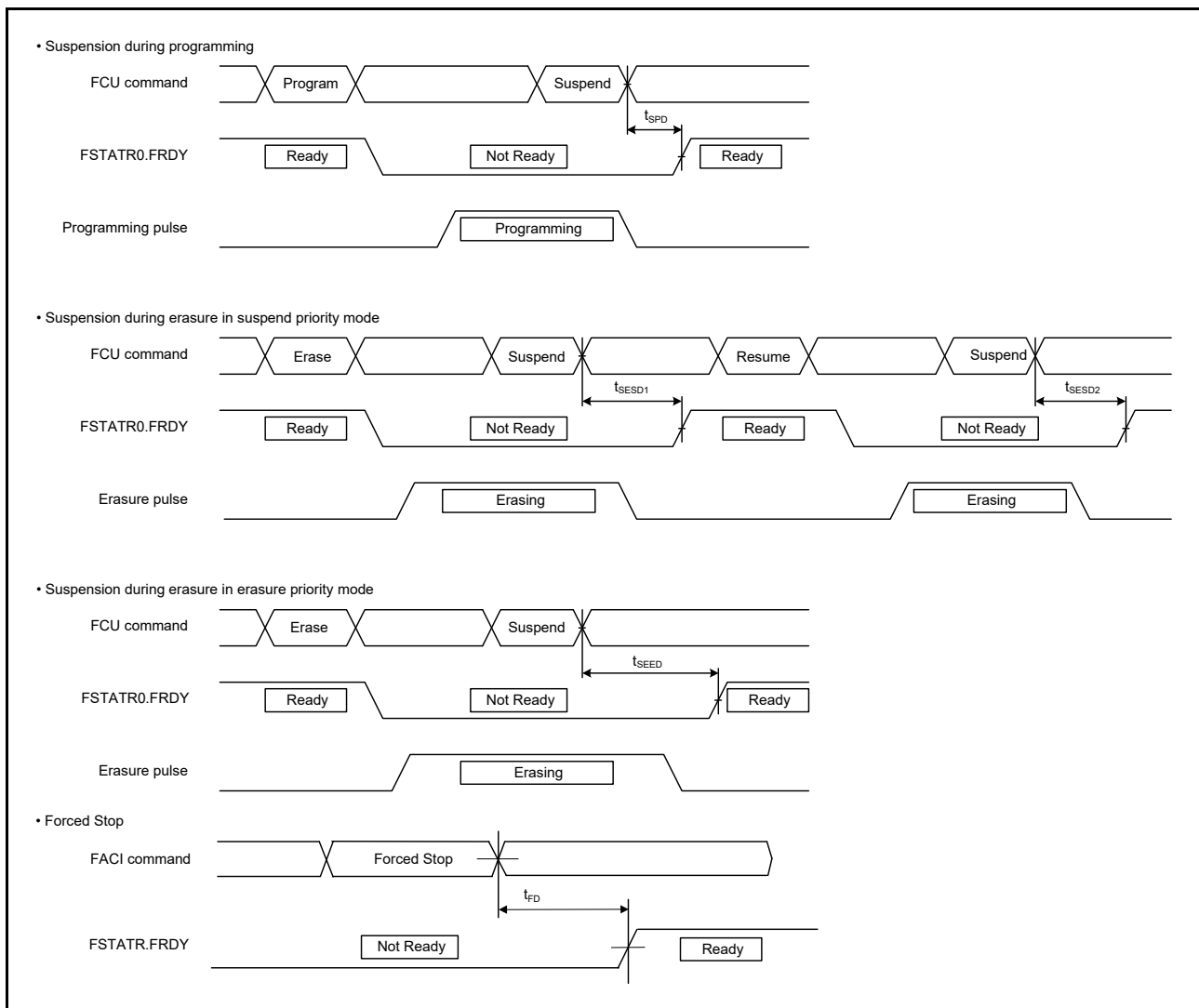


Figure 2.49 Suspension and forced stop timing for flash memory programming and erasure

2.11.2 Data Flash Memory Characteristics

Table 2.39 Data flash memory characteristics

Conditions: Program or erase: FCLK = 4 to 60 MHz

Read: FCLK ≤ 60 MHz

Parameter	Symbol	FCLK = 4 MHz			20 MHz ≤ FCLK ≤ 60 MHz			Unit	Test conditions
		Min	Typ	Max	Min	Typ	Max		
Programming time	4-byte	t _{DP4}	-	0.36	3.8	-	0.16	1.7	ms
	8-byte	t _{DP8}	-	0.38	4.0	-	0.17	1.8	
	16-byte	t _{DP16}	-	0.42	4.5	-	0.19	2.0	
Erasure time	64-byte	t _{DE64}	-	3.1	18	-	1.7	10	ms
	128-byte	t _{DE128}	-	4.7	27	-	2.6	15	
	256-byte	t _{DE256}	-	8.9	50	-	4.9	28	
Blank check time	4-byte	t _{DBC4}	-	-	84	-	-	30	μs
Reprogramming/erase cycle*1	N _{DPEC}	125000*2	-	-	125000*2	-	-	-	-
Suspend delay during programming	4-byte	t _{DSPD}	-	-	264	-	-	120	μs
	8-byte		-	-	264	-	-	120	
	16-byte		-	-	264	-	-	120	
First suspend delay during erasure in suspend priority mode	64-byte	t _{DSESD1}	-	-	216	-	-	120	μs
	128-byte		-	-	216	-	-	120	
	256-byte		-	-	216	-	-	120	
Second suspend delay during erasure in suspend priority mode	64-byte	t _{DSESD2}	-	-	300	-	-	300	μs
	128-byte		-	-	390	-	-	390	
	256-byte		-	-	570	-	-	570	
Suspend delay during erasing in erasure priority mode	64-byte	t _{DSEED}	-	-	300	-	-	300	μs
	128-byte		-	-	390	-	-	390	
	256-byte		-	-	570	-	-	570	
Forced stop command	t _{FD}	-	-	32	-	-	20	μs	
Data hold time*3	t _{DRP}	10*3,*4	-	-	10*3,*4	-	-	Year	
		30*3,*4	-	-	30*3,*4	-	-		Ta = +85°C

Note 1. The reprogram/erase cycle is the number of erasures for each block. When the reprogram/erase cycle is n times (n = 125000), erasing can be performed n times for each block. For example, when 4-byte programming is performed 16 times for different addresses in 64-byte blocks, and then the entire block is erased, the reprogram/erase cycle is counted as one. However, programming the same address several times as one erasure is not enabled. Overwriting is prohibited.

Note 2. This is the minimum number of times to guarantee all the characteristics after reprogramming. The guaranteed range is from 1 to the minimum value.

Note 3. This indicates the minimum value of the characteristic when reprogramming is performed within the specified range.

Note 4. This result is obtained from reliability testing.

2.12 Boundary Scan

Table 2.40 Boundary scan characteristics (1 of 2)

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
TCK clock cycle time	t _{TCKcyc}	100	-	-	ns	Figure 2.50
TCK clock high pulse width	t _{TCKH}	45	-	-	ns	
TCK clock low pulse width	t _{TCKL}	45	-	-	ns	
TCK clock rise time	t _{TCKr}	-	-	5	ns	
TCK clock fall time	t _{TCKf}	-	-	5	ns	

Table 2.40 Boundary scan characteristics (2 of 2)

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
TMS setup time	t_{TMSS}	20	-	-	ns	Figure 2.51
TMS hold time	t_{TMSH}	20	-	-	ns	
TDI setup time	t_{TDIS}	20	-	-	ns	
TDI hold time	t_{TDIH}	20	-	-	ns	
TDO data delay	t_{TDOD}	-	-	40	ns	
Boundary scan circuit startup time*1	T_{BSSTUP}	t_{RESWP}	-	-	-	Figure 2.52

Note 1. Boundary scan does not function until the power-on reset becomes negative.

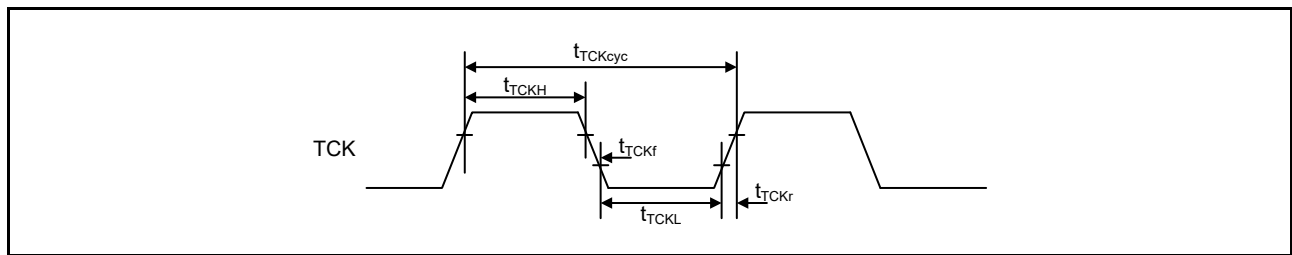


Figure 2.50 Boundary scan TCK timing

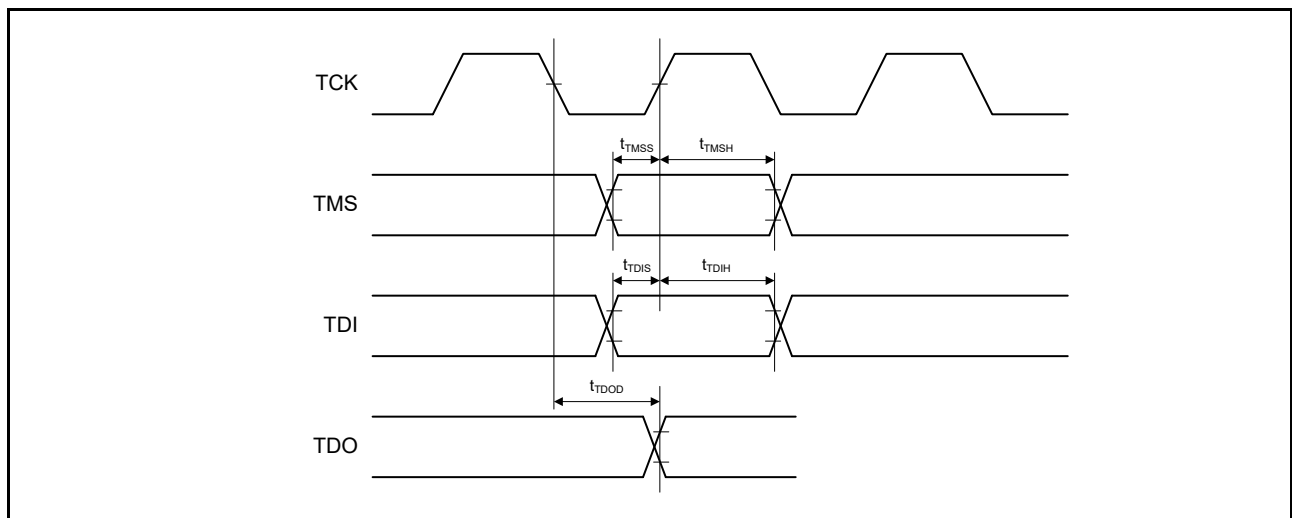


Figure 2.51 Boundary scan input/output timing

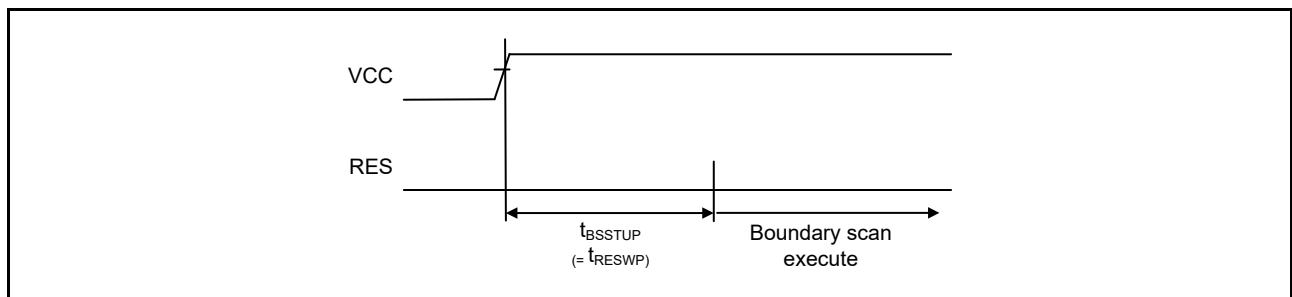


Figure 2.52 Boundary scan circuit startup timing

2.13 Joint Test Action Group (JTAG)

Table 2.41 JTAG

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
TCK clock cycle time	t_{TCKcyc}	40	-	-	ns	Figure 2.50
TCK clock high pulse width	t_{TCKH}	15	-	-	ns	
TCK clock low pulse width	t_{TCKL}	15	-	-	ns	
TCK clock rise time	t_{TCKr}	-	-	5	ns	
TCK clock fall time	t_{TCKf}	-	-	5	ns	
TMS setup time	t_{TMSS}	8	-	-	ns	Figure 2.51
TMS hold time	t_{TMSh}	8	-	-	ns	
TDI setup time	t_{TDis}	8	-	-	ns	
TDI hold time	t_{TDIH}	8	-	-	ns	
TDO data delay time	t_{TDOD}	-	-	20	ns	

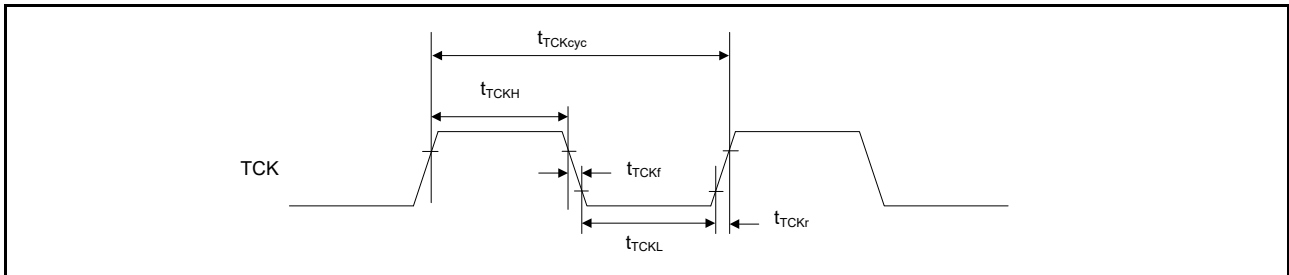


Figure 2.53 JTAG TCK timing

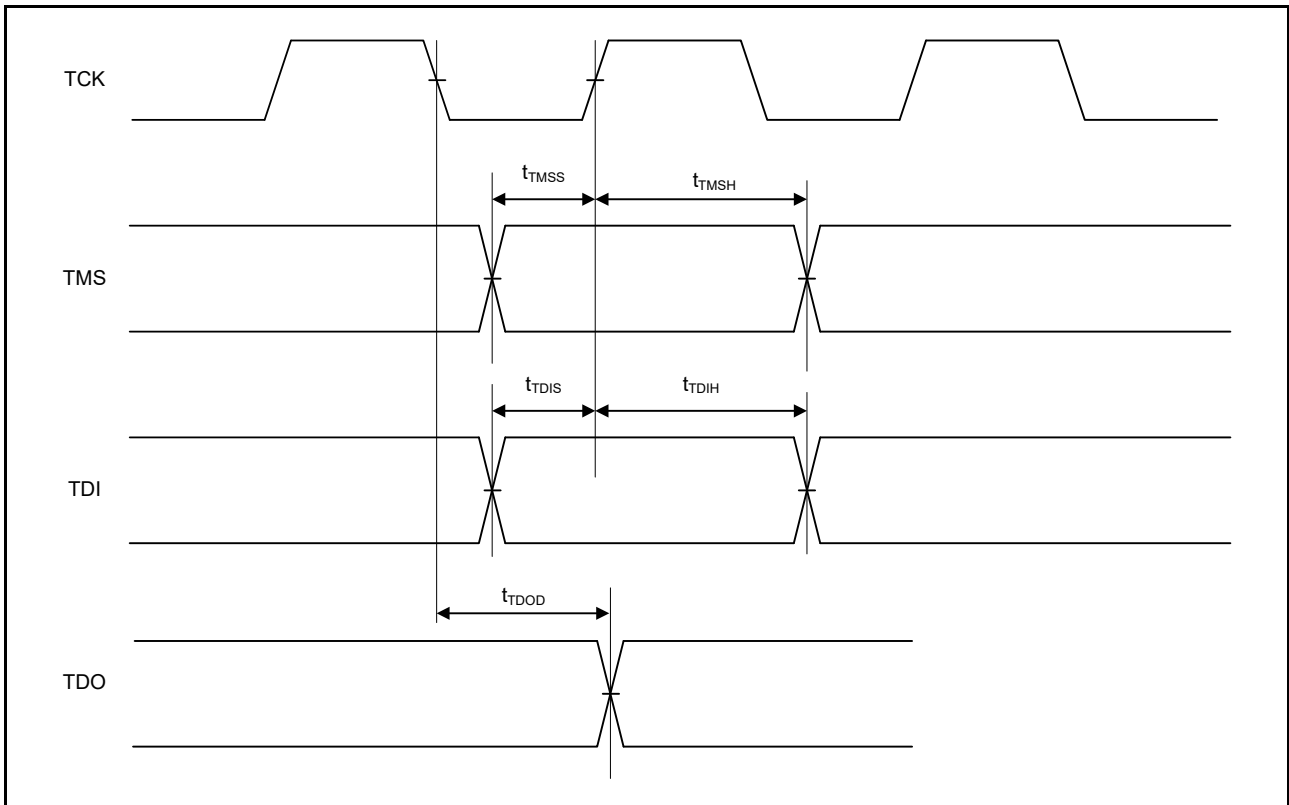


Figure 2.54 JTAG input/output timing

2.14 Serial Wire Debug (SWD)

Table 2.42 SWD

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
SWCLK clock cycle time	$t_{SWCLKcyc}$	40	-	-	ns	Figure 2.55
SWCLK clock high pulse width	t_{SWCKH}	15	-	-	ns	
SWCLK clock low pulse width	t_{SWCKL}	15	-	-	ns	
SWCLK clock rise time	t_{SWCKr}	-	-	5	ns	
SWCLK clock fall time	t_{SWCKf}	-	-	5	ns	
SWDIO setup time	t_{SWDS}	8	-	-	ns	Figure 2.56
SWDIO hold time	t_{SWDH}	8	-	-	ns	
SWDIO data delay time	t_{SWDD}	2	-	28	ns	

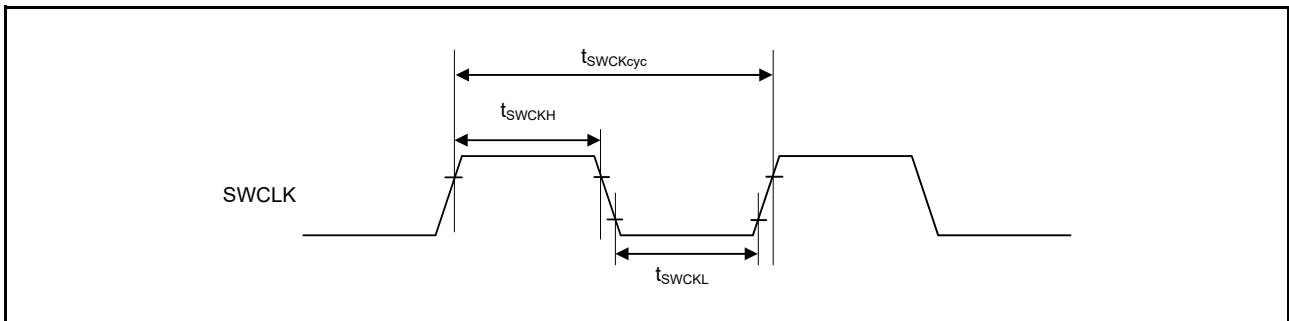


Figure 2.55 SWD SWCLK timing

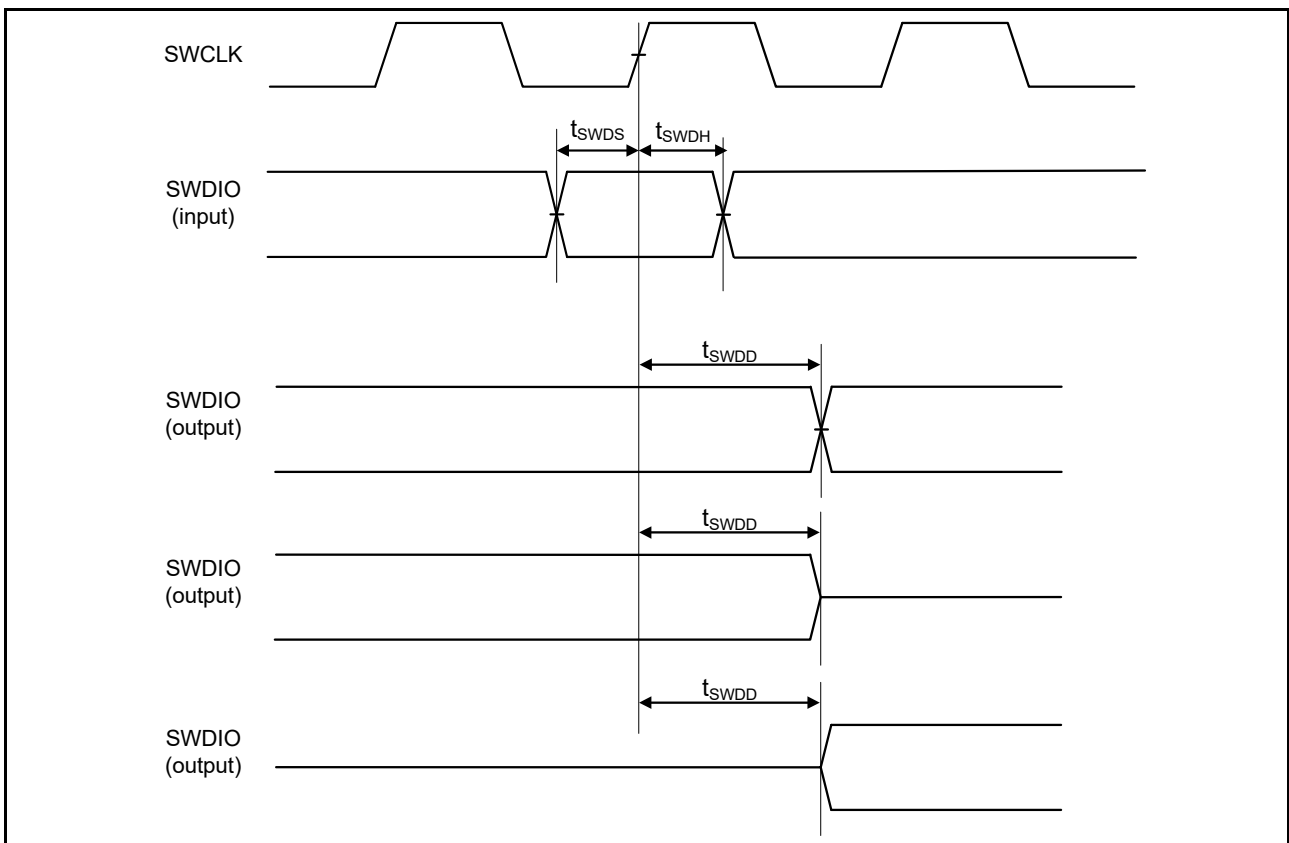


Figure 2.56 SWD input/output timing

2.15 Embedded Trace Macro Interface (ETM)

Table 2.43 ETM

Conditions: High drive output is selected in the Port Drive Capability bit in the PmnPFS register.

Parameter	Symbol	Min	Typ	Max	Unit	Test conditions
TCLK clock cycle time	$t_{TCLK_{cyc}}$	33.3	-	-	ns	Figure 2.57
TCLK clock high pulse width	t_{TCLKH}	13.6	-	-	ns	
TCLK clock low pulse width	t_{TCLKL}	13.6	-	-	ns	
TCLK clock rise time	t_{TCLKr}	-	-	3	ns	
TCLK clock fall time	t_{TCLKf}	-	-	3	ns	
TDATA[3:0] output setup time	t_{TRDS}	3.5	-	-	ns	Figure 2.58
TDATA[3:0] output hold time	t_{TRDH}	2.5	-	-	ns	

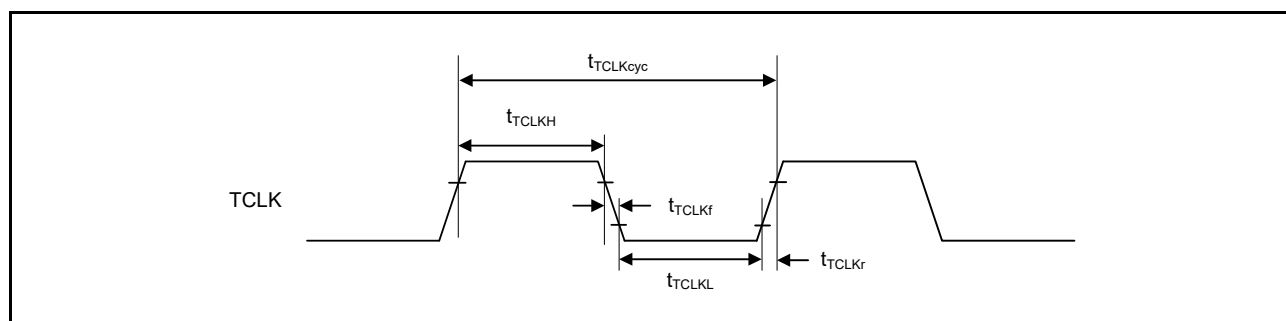


Figure 2.57 ETM TCLK timing

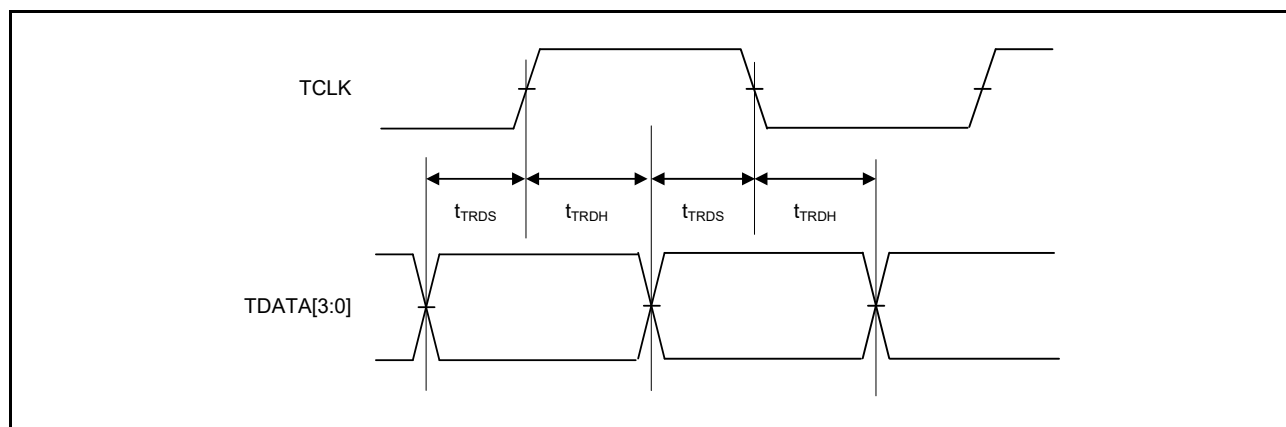


Figure 2.58 ETM output timing

Appendix 1. Package Dimensions

Information on the latest version of the package dimensions or mountings is shown in “Packages” on the Renesas Electronics Corporation website.

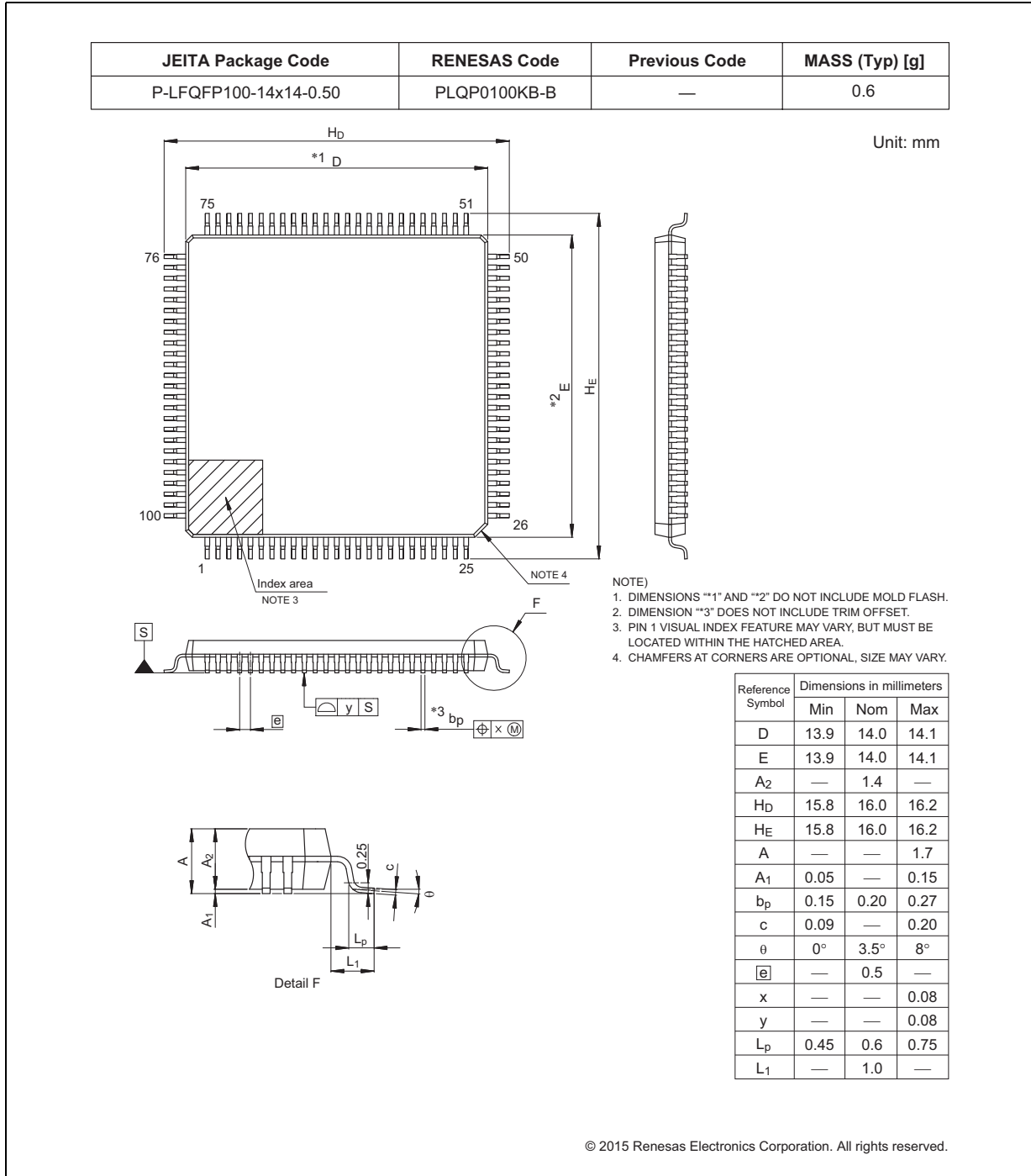
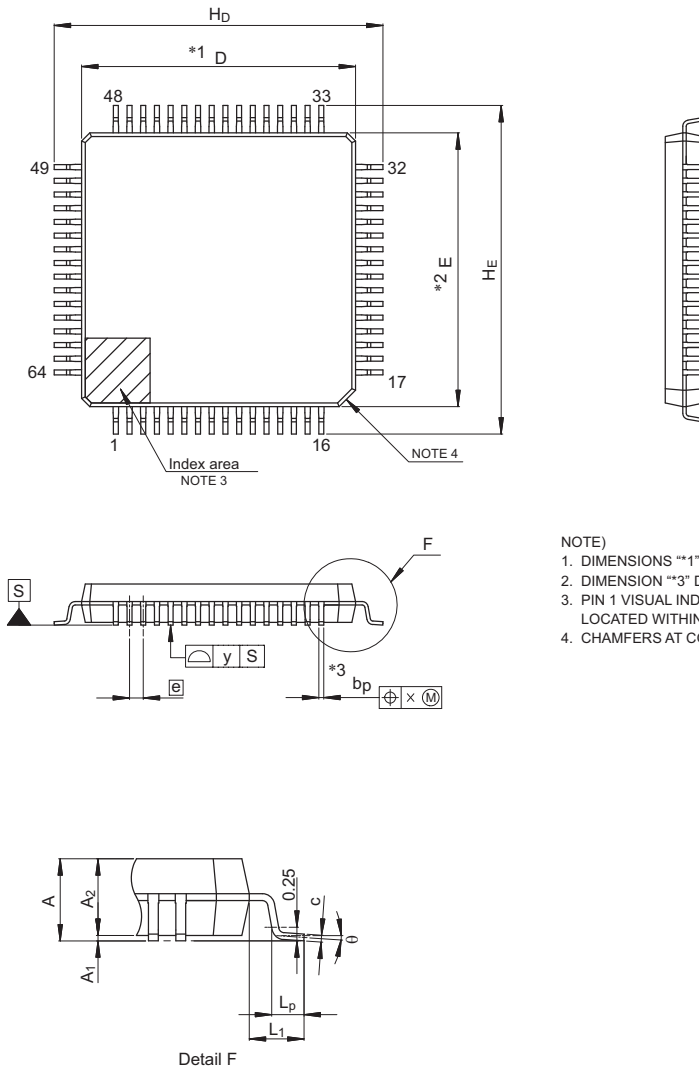


Figure 1.1 100-pin LQFP

JEITA Package Code	RENESAS Code	Previous Code	MASS (Typ) [g]
P-LFQFP64-10x10-0.50	PLQP0064KB-C	—	0.3

Unit: mm



- NOTE)
1. DIMENSIONS "**1" AND "**2" DO NOT INCLUDE MOLD FLASH.
 2. DIMENSION "**3" DOES NOT INCLUDE TRIM OFFSET.
 3. PIN 1 VISUAL INDEX FEATURE MAY VARY, BUT MUST BE LOCATED WITHIN THE HATCHED AREA.
 4. CHAMFERS AT CORNERS ARE OPTIONAL, SIZE MAY VARY.

Reference Symbol	Dimensions in millimeters		
	Min	Nom	Max
D	9.9	10.0	10.1
E	9.9	10.0	10.1
A ₂	—	1.4	—
H _D	11.8	12.0	12.2
H _E	11.8	12.0	12.2
A	—	—	1.7
A ₁	0.05	—	0.15
b _p	0.15	0.20	0.27
c	0.09	—	0.20
θ	0°	3.5°	8°
e	—	0.5	—
x	—	—	0.08
y	—	—	0.08
L _p	0.45	0.6	0.75
L ₁	—	1.0	—

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Figure 1.2 64-pin LQFP

Revision History	RA6T1 Group Datasheet
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Rev.	Date	Chapter	Summary
1.00	May 29, 2020	-	First Edition issued
1.10	Feb 24, 2022	-	Second Edition issued
1.20	Mar 3, 2023	-	Third Edition issued

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1. Precaution against Electrostatic Discharge (ESD)

A strong electrical field, when exposed to a CMOS device, can cause destruction of the gate oxide and ultimately degrade the device operation. Steps must be taken to stop the generation of static electricity as much as possible, and quickly dissipate it when it occurs. Environmental control must be adequate. When it is dry, a humidifier should be used. This is recommended to avoid using insulators that can easily build up static electricity. Semiconductor devices must be stored and transported in an anti-static container, static shielding bag or conductive material. All test and measurement tools including work benches and floors must be grounded. The operator must also be grounded using a wrist strap. Semiconductor devices must not be touched with bare hands. Similar precautions must be taken for printed circuit boards with mounted semiconductor devices.

2. Processing at power-on

The state of the product is undefined at the time when power is supplied. The states of internal circuits in the LSI are indeterminate and the states of register settings and pins are undefined at the time when power is supplied. In a finished product where the reset signal is applied to the external reset pin, the states of pins are not guaranteed from the time when power is supplied until the reset process is completed. In a similar way, the states of pins in a product that is reset by an on-chip power-on reset function are not guaranteed from the time when power is supplied until the power reaches the level at which resetting is specified.

3. Input of signal during power-off state

Do not input signals or an I/O pull-up power supply while the device is powered off. The current injection that results from input of such a signal or I/O pull-up power supply may cause malfunction and the abnormal current that passes in the device at this time may cause degradation of internal elements. Follow the guideline for input signal during power-off state as described in your product documentation.

4. Handling of unused pins

Handle unused pins in accordance with the directions given under handling of unused pins in the manual. The input pins of CMOS products are generally in the high-impedance state. In operation with an unused pin in the open-circuit state, extra electromagnetic noise is induced in the vicinity of the LSI, an associated shoot-through current flows internally, and malfunctions occur due to the false recognition of the pin state as an input signal become possible.

5. Clock signals

After applying a reset, only release the reset line after the operating clock signal becomes stable. When switching the clock signal during program execution, wait until the target clock signal is stabilized. When the clock signal is generated with an external resonator or from an external oscillator during a reset, ensure that the reset line is only released after full stabilization of the clock signal. Additionally, when switching to a clock signal produced with an external resonator or by an external oscillator while program execution is in progress, wait until the target clock signal is stable.

6. Voltage application waveform at input pin

Waveform distortion due to input noise or a reflected wave may cause malfunction. If the input of the CMOS device stays in the area between V_{IL} (Max.) and V_{IH} (Min.) due to noise, for example, the device may malfunction. Take care to prevent chattering noise from entering the device when the input level is fixed, and also in the transition period when the input level passes through the area between V_{IL} (Max.) and V_{IH} (Min.).

7. Prohibition of access to reserved addresses

Access to reserved addresses is prohibited. The reserved addresses are provided for possible future expansion of functions. Do not access these addresses as the correct operation of the LSI is not guaranteed.

8. Differences between products

Before changing from one product to another, for example to a product with a different part number, confirm that the change will not lead to problems. The characteristics of a microprocessing unit or microcontroller unit products in the same group but having a different part number might differ in terms of internal memory capacity, layout pattern, and other factors, which can affect the ranges of electrical characteristics, such as characteristic values, operating margins, immunity to noise, and amount of radiated noise. When changing to a product with a different part number, implement a system-evaluation test for the given product.

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